

## Advanced Load Management Controlled Load Switch with Low R<sub>ON</sub>

### General Description

The SQ24801C load switch is an ultra-low on-resistance, compact device with inrush current limit via programmable soft-start. The device provides power good signaling for system status monitoring and downstream load control. With soft-start capabilities to reduce inrush current and low power consumption in a small footprint, the SQ24801C is ideal for power management and hot-swap applications.

The SQ24801C is available in a compact DFN3x3-12 package.

### Features

- Integrated N-Channel MOSFET with Ultra Low R<sub>ON</sub>
- Input Voltage Range 3 V to 24 V
- 3V to 5.5V Input Voltage Range for VCC
- Programmable Soft Start Time
- Fault Detection with Power Good Output
- Thermal Shutdown Protection
- Short Circuit and Adjustable Over-Current Protection and Latch off
- Output Capacitor Discharge during EN OFF
- Extremely Low Standby Current
- RoHS Compliant and Halogen Free

### Applications

- USB Type C Power Delivery
- Portable Electronics and Systems
- Notebook and Tablet Computers
- Telecom, Networking, Medical, and Industrial Equipment
- Set-Top Boxes, Servers, and Gateways
- Hot-Swap Devices and Peripheral Ports

### Typical Application

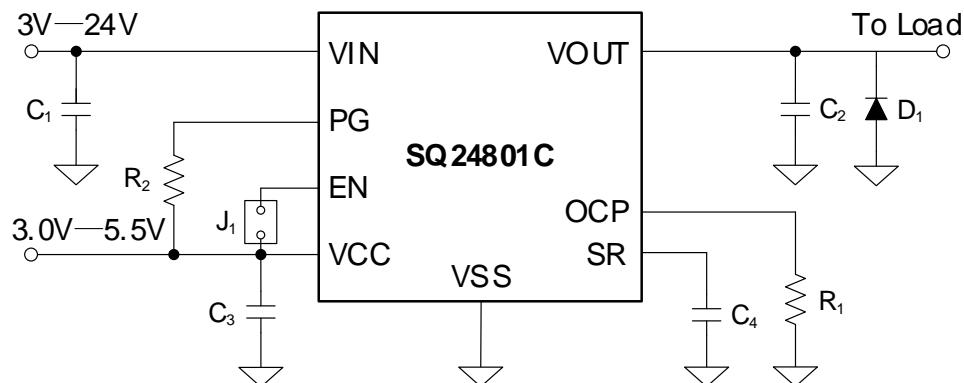


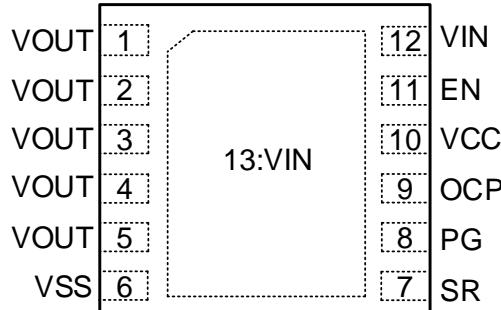
Figure 1. Schematic Diagram

**Ordering Information**

Ordering Part Number	Package Type	Top Mark
SQ24801CDCD	DFN3x3-12 RoHS Compliant and Halogen Free	EHTxyz

Device code: EHT

x=year code, y=week code, z= lot number code

**Pinout (top view)**

**Pin Description**

Pin Name	Pin number	Pin Description
VOUT	1, 2, 3, 4, 5	Source of MOSFET connected to load. Includes an internal bleed resistor to GND. – All pins must be connected to provide correct $R_{DS}$ , OCP, and current capability.
VSS	6	Driver ground.
SR	7	Slew Rate control pin. Slew rate adjustment made with an external capacitor to GND. Leave it floating if not used.
PG	8	Active-high, open-drain output that indicates when the gate of the MOSFET is fully charged, external pull-up resistor $\geq 100\text{k}\Omega$ to an external voltage source required; ties to GND if not used.
OCP	9	Over-current protection trip point adjustment made with a resistor to ground; short to ground if over-current protection is not needed.
VCC	10	Driver supply voltage (3.0 V – 5.5 V).
EN	11	Active-high digital input used to turn on the MOSFET driver, pin has an internal pull-down resistor to GND.
VIN	12, 13	Input voltage (3 V – 24 V) – Pin 13 should be used for high current ( $>0.5\text{ A}$ ).

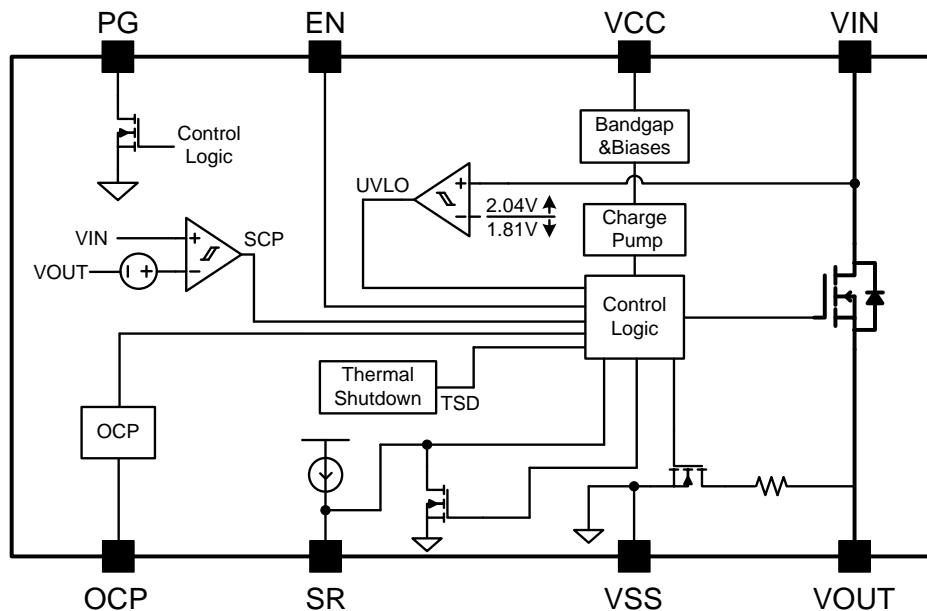
**Block Diagram**


Figure 2 Block Diagram



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## Absolute Maximum Ratings

Parameter (Note 1)	Min	Max	Unit
VCC, EN, PG, OCP	-0.3	6	V
VIN, VOUT	-0.3	30	
I <sub>MAX</sub> (Note 2)		20	A
Lead Temperature (Soldering, 10s)		260	
Junction Temperature, Operating	-40	150	
Storage Temperature	-65	150	°C

## Thermal Information

Parameter (Note 3)	Typ	Unit
θ <sub>JA</sub> Junction-to-Ambient Thermal Resistance	35	°C/W
θ <sub>JC</sub> Junction-to-Case Thermal Resistance	1.7	
P <sub>D</sub> Power Dissipation T <sub>A</sub> = 25°C	2.86	W

## Recommended Operating Conditions

Parameter (Note 4)	Min	Max	Unit
VCC	3	5.5	
VIN, VOUT	3	24	V
EN, PG	0	5.5	
OCP External Resistor to VSS	short	open	
Junction Temperature, Operating	-40	125	
Ambient Temperature	-40	85	°C



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## Electrical Characteristics

( $R_{PG} = 100\text{k}\Omega$ ;  $R_L = 10\Omega$ ,  $C_L = 0.1\mu\text{F}$ ,  $T_J = -40^\circ\text{C}$  to  $125^\circ\text{C}$ , typical values are  $T_J = 25^\circ\text{C}$ , unless otherwise specified. The values are guaranteed by test, design or statistical correlation)

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
<b>MOSFET</b>						
On-Resistance	$R_{ON}$	$V_{CC} = 4.5\text{ V}$ ; $V_{IN} = 3\text{ V}$		3.8	6.5	$\text{m}\Omega$
		$V_{CC} = 3.3\text{ V}$ ; $V_{IN} = 4.5\text{ V}$		3.8	6.5	
		$V_{CC} = 3.3\text{ V}$ ; $V_{IN} = 15\text{ V}$		3.8	6.5	
		$V_{CC} = 3.3\text{ V}$ ; $V_{IN} = 24\text{ V}$		3.8	6.5	
Leakage Current	$I_{LEAK}$	$V_{EN} = 0\text{ V}$ ; $V_{IN} = 24\text{ V}$		22.8	10000	nA
<b>Controller</b>						
$V_{IN}$ Control Current – $V_{IN}$ to $V_{SS}$	$I_{INCTL}$	$V_{EN} = 0\text{ V}$ ; $V_{IN} = 24\text{ V}$		6	30	$\mu\text{A}$
$V_{IN}$ Control Current – $V_{IN}$ to $V_{SS}$	$I_{INCTL\_EN}$	$V_{EN} = V_{CC}$ ; $V_{IN} = 24\text{ V}$		190	260	$\mu\text{A}$
Supply Standby Current	$I_{STBY}$	$V_{EN} = 0\text{ V}$ ; $V_{IN} = 24\text{ V}$		2.5	8	$\mu\text{A}$
Supply Dynamic Current	$I_{DYN}$	$V_{EN} = V_{CC}$ ; $V_{IN} = 24\text{ V}$		0.08	0.11	mA
EN Input High Voltage	$V_{IH}$		2.0			V
EN Input Low Voltage	$V_{IL}$				0.8	V
EN Input Leakage Current	$I_{IL}$	$V_{EN} = 0\text{ V}$	-1.0	0.02	1.0	$\mu\text{A}$
EN Pull Down Resistance	$R_{PD}$		75	100	130	$\text{k}\Omega$
PG Output Leakage Current	$I_{OH}$	$V_{PG} = 3.3\text{ V}$		3.3	100	nA
Slew Rate Control Constant	$K_{SR}$	SR pin floating (default)	60	95	130	$\mu\text{A}$
<b>Fault Protections</b>						
Thermal Shutdown Threshold (Note 5)	$T_{SDT}$			145		$^\circ\text{C}$
Thermal Shutdown Hysteresis (Note 5)	$T_{HYS}$			20		$^\circ\text{C}$
$V_{IN}$ Undervoltage Lockout Threshold	$V_{UVLO}$	$V_{IN}$ rising	1.8	2.04	2.3	V
$V_{IN}$ Undervoltage Lockout Hysteresis	$V_{HYS}$	$V_{CC} = 3\text{ V}$	150	227	300	$\text{mV}$
Over-Current Protection Trip ( $V_{CC} = 3.3\text{ V}$ )	$I_{TRIP}$	$ROCP = \text{open}$	1.9	2.65	3.4	A
		$ROCP = 100\text{ k}\Omega$		9.3		
		$ROCP = 20\text{ k}\Omega$		16		
		$ROCP = 1\text{ k}\Omega$		20		
		$ROCP = \text{short to GND}$		20		
Over-Current Protection Blanking Time	$t_{OCP}$			2.25		ms
Short-Circuit Protection Trip Current (Note 5)	$I_{SC}$	$T_J = -40^\circ\text{C}$		40		A
		$T_J = 150^\circ\text{C}$		20		
		$T_A = 25^\circ\text{C}$ , DC Current		30		

## Switching Characteristics

( $R_{PG} = 100\text{k}\Omega$ ;  $R_L = 10\Omega$ ,  $C_L = 0.1\mu\text{F}$ ,  $T_J = -40^\circ\text{C}$  to  $125^\circ\text{C}$ , typical values are  $T_J = 25^\circ\text{C}$ , unless otherwise specified. The values are guaranteed by test, design or statistical correlation)

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Output Slew Rate	SR	$V_{CC} = 4.5\text{ V}$ ; $V_{IN} = 3\text{ V}$	7	16	25	KV/s
		$V_{CC} = 5.0\text{ V}$ ; $V_{IN} = 3\text{ V}$	7	16	25	
		$V_{CC} = 3.3\text{ V}$ ; $V_{IN} = 24\text{ V}$	10	18	26	
		$V_{CC} = 5.0\text{ V}$ ; $V_{IN} = 24\text{ V}$	10	18	26	
Output Turn-on Delay	$t_{ON}$	$V_{CC} = 4.5\text{ V}$ ; $V_{IN} = 3\text{ V}$	100	350	700	$\mu\text{s}$
		$V_{CC} = 5.0\text{ V}$ ; $V_{IN} = 3\text{ V}$	100	350	700	
		$V_{CC} = 3.3\text{ V}$ ; $V_{IN} = 24\text{ V}$	100	430	700	
		$V_{CC} = 5.0\text{ V}$ ; $V_{IN} = 24\text{ V}$	100	430	700	
Output Turn-off Delay	$t_{OFF}$	$V_{CC} = 4.5\text{ V}$ ; $V_{IN} = 3\text{ V}$		48		$\mu\text{s}$
		$V_{CC} = 5.0\text{ V}$ ; $V_{IN} = 3\text{ V}$		48		
		$V_{CC} = 3.3\text{ V}$ ; $V_{IN} = 24\text{ V}$		48		
		$V_{CC} = 5.0\text{ V}$ ; $V_{IN} = 24\text{ V}$		48		
Power Good Turn-on Time	$t_{PG,ON}$	$V_{CC} = 4.5\text{ V}$ ; $V_{IN} = 3\text{ V}$	0.25	0.68	2.5	$\text{ms}$
		$V_{CC} = 5.0\text{ V}$ ; $V_{IN} = 3\text{ V}$	0.25	0.72	2.5	
		$V_{CC} = 3.3\text{ V}$ ; $V_{IN} = 24\text{ V}$	0.25	1.4	2.5	
		$V_{CC} = 5.0\text{ V}$ ; $V_{IN} = 24\text{ V}$	0.25	1.4	2.5	
Power Good Turn-off Time	$t_{PG,OFF}$	$V_{CC} = 4.5\text{ V}$ ; $V_{IN} = 3\text{ V}$		4.5		$\mu\text{s}$
		$V_{CC} = 5.0\text{ V}$ ; $V_{IN} = 3\text{ V}$		4.5		
		$V_{CC} = 3.3\text{ V}$ ; $V_{IN} = 24\text{ V}$		4.5		
		$V_{CC} = 5.0\text{ V}$ ; $V_{IN} = 24\text{ V}$		4.5		

**Note 1:** Stresses beyond the “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**Note 2:** Ensure that the expected operating MOSFET current will not cause the Short-Circuit Protection to turn the MOSFET off undesirably.

**Note 3:**  $\theta_{JA}$  is measured in the natural convection at  $T_A = 25^\circ\text{C}$  on a Silergy EVB test board.

**Note 4:** The device is not guaranteed to function outside its operating conditions.

**Note 5:** Guaranteed by design, not production test.

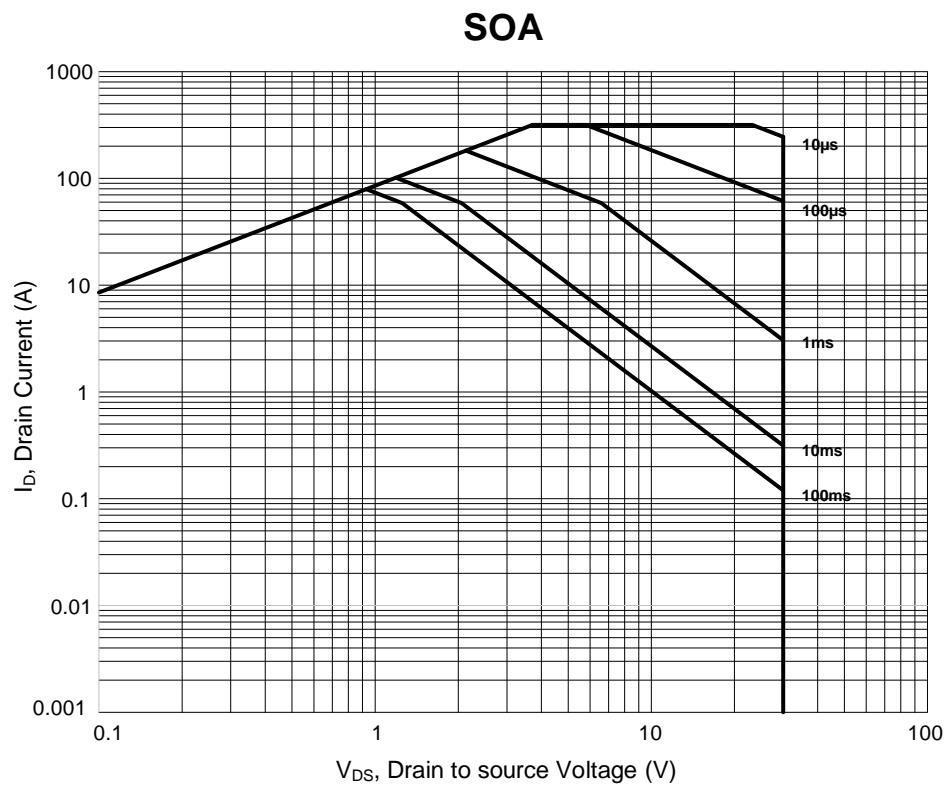
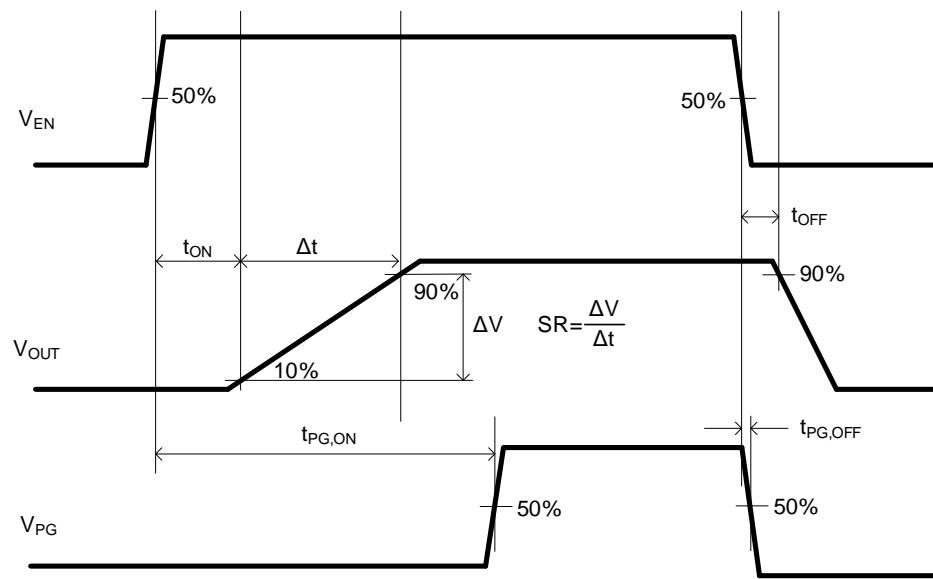
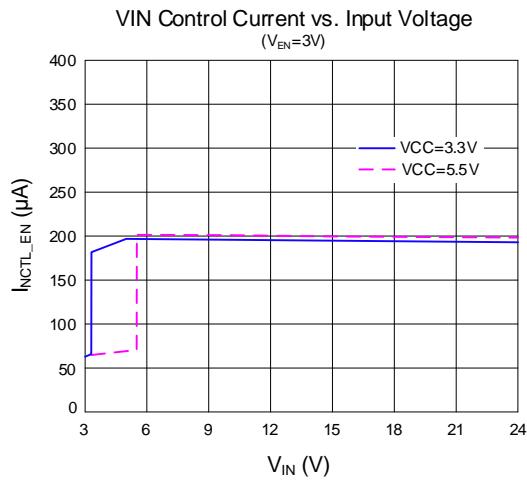
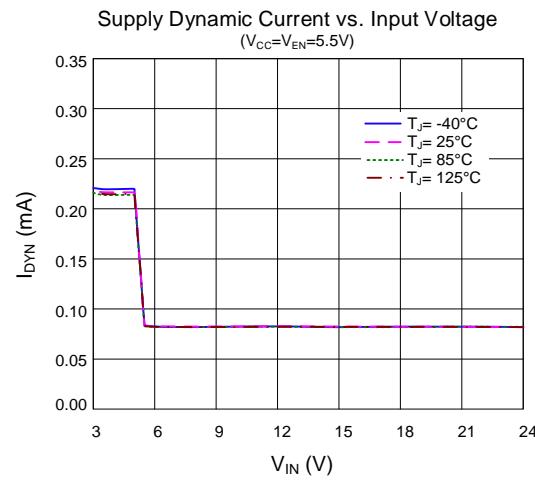
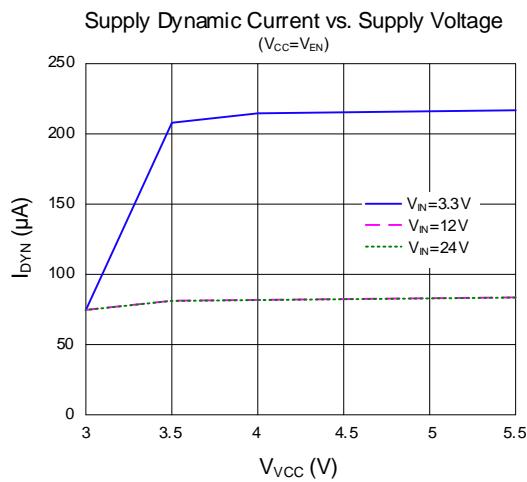
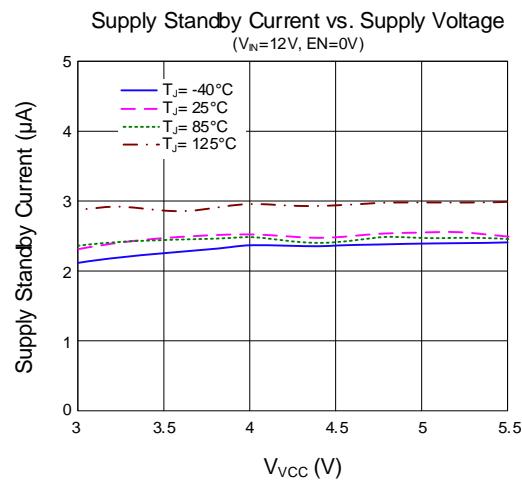
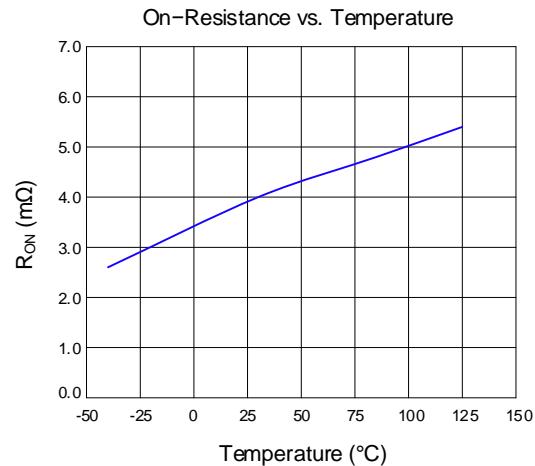
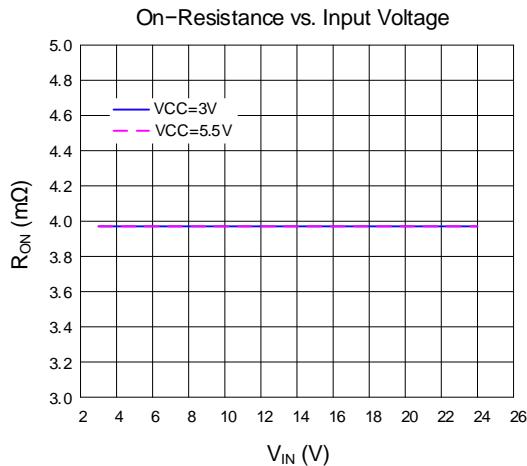


Figure 4. Safe Operating Area

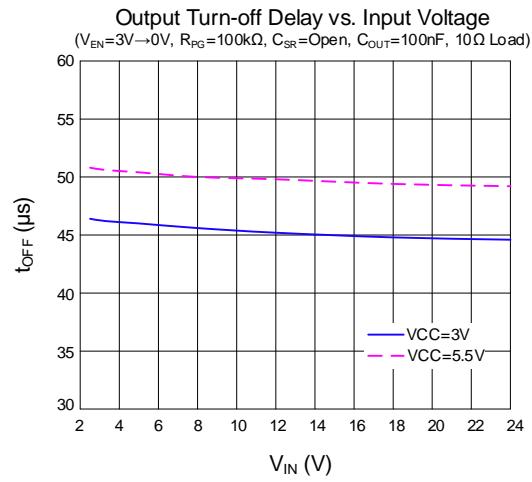
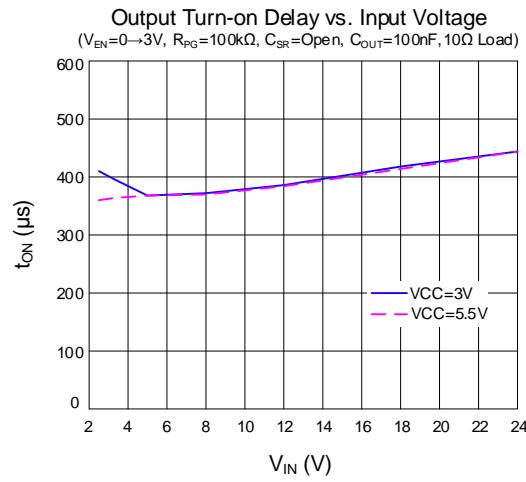
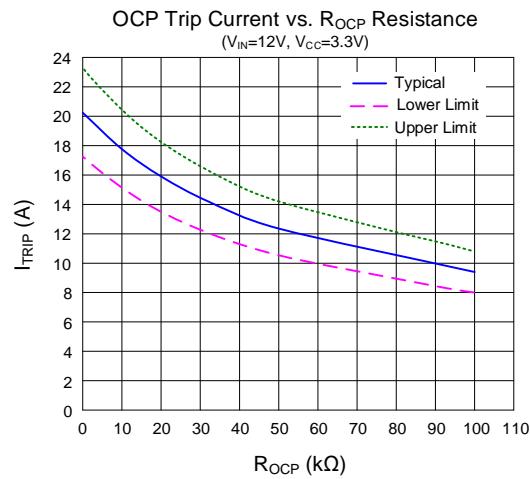
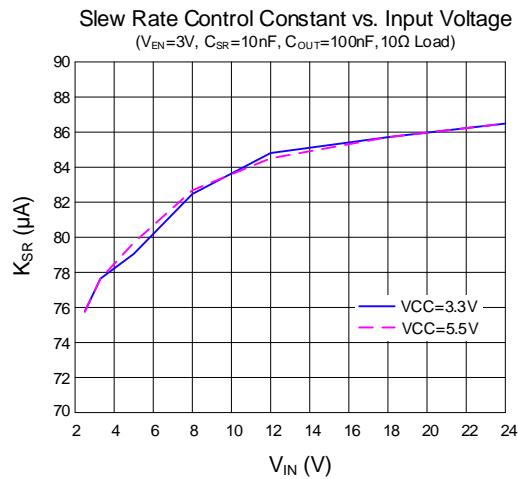
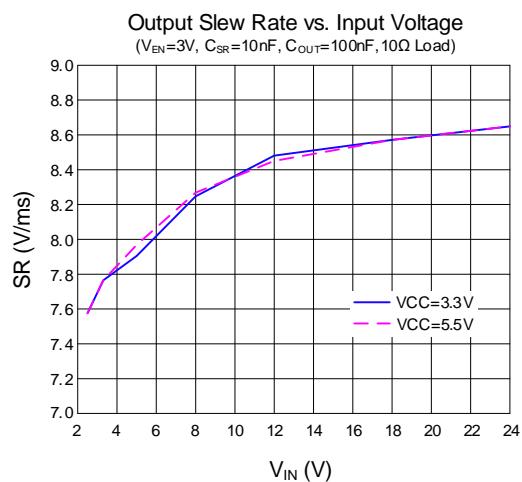
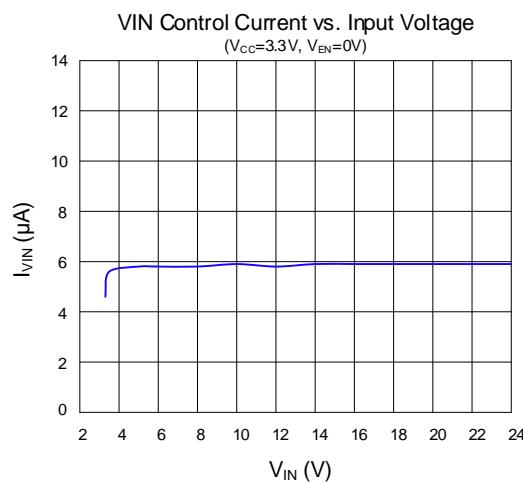
## Typical Performance Characteristics

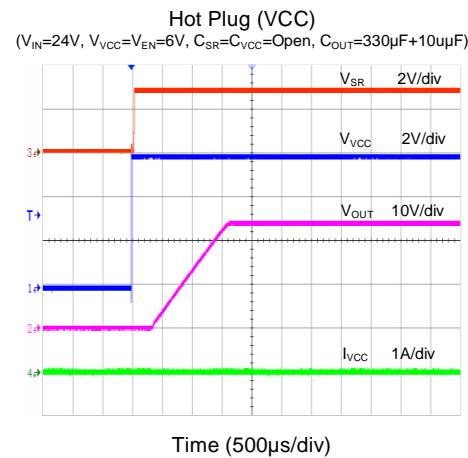
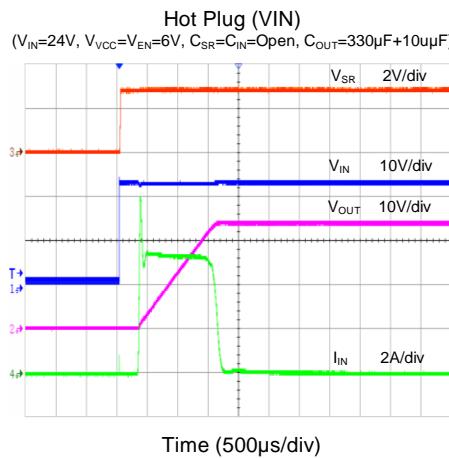
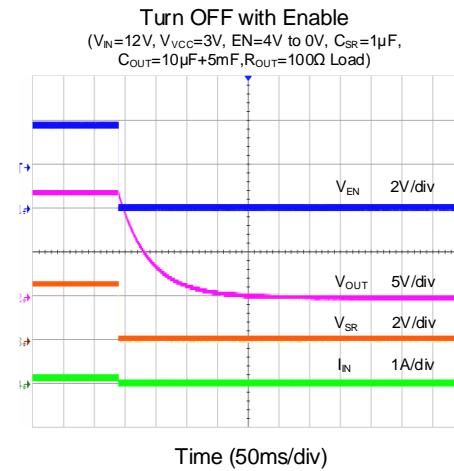
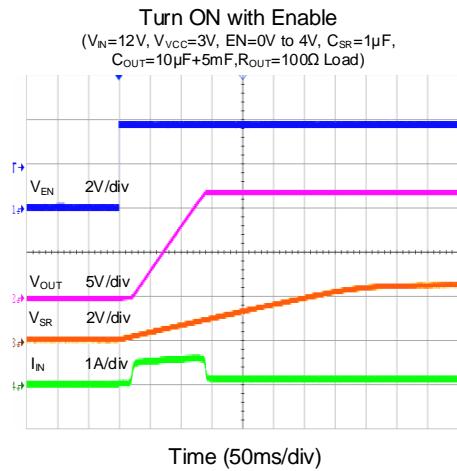
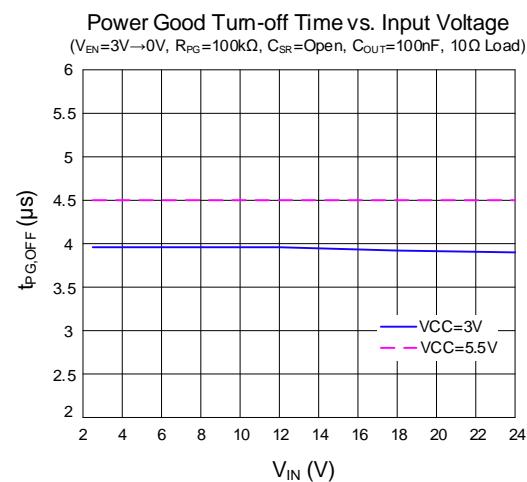
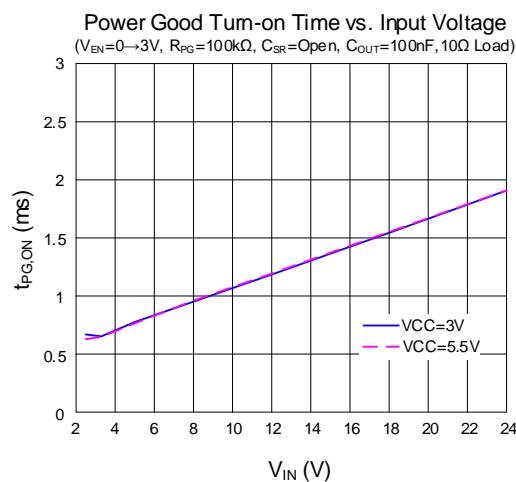




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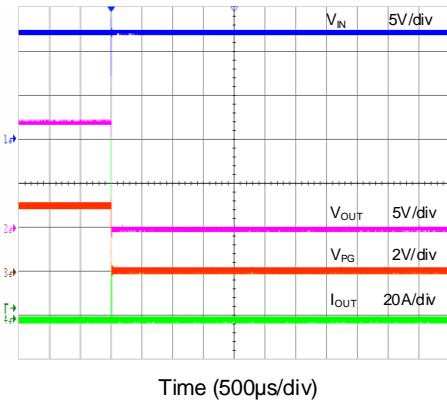


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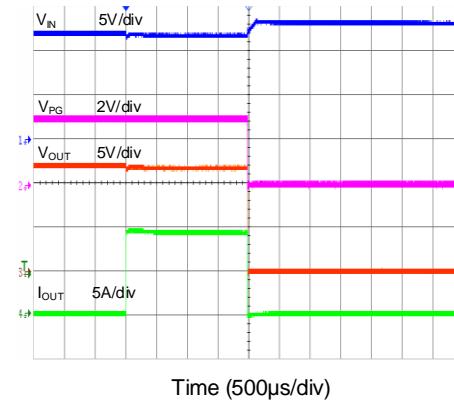
**Hard-Short**

( $V_{IN}=12V$ ,  $V_{VCC}=3V$ ,  $R_{PG}=10k\Omega$ ,  $C_{SR}=\text{open}$ ,  $C_{OUT}=10\mu F$ ,  
Short OUT)



**Over Current Protection**

( $V_{IN}=12V$ ,  $V_{VCC}=3V$ ,  $R_{PG}=100k\Omega$ ,  $C_{SR}=\text{open}$ ,  $C_{OUT}=10\mu F$ ,  
0A to 9.5A)





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## Applications Information

### Enable Control

The SQ24801CDCC allows for enabling the MOSFET in an active-high configuration. If VCC power pin has enough voltage application and the EN pin is in the logic position High level, the MOSFET will be enabled. If the EN pin is at logic low level, the MOSFET will be disabled. A sort of the internal pull-down resistor is grounded to ensure the EN pin when there is no driver, the MOSFET will be disabled.

### Power Sequencing

The SQ24801CDCC must consider the power on timing specification for  $V_{CC} \geq 3V$  before EN is enabled ( $V_{EN} \leq 1.1V$  within the full temperature range) in design. To achieve the specified performance, there are two recommended power sequences:

1. VCC → VIN → VEN
2. VIN → VCC → VEN

### Under Voltage Lockout (UVLO)

The MOSFET is turned off by the UVLO of the SQ24801CDCC when the input voltage (VIN) is lower than the UVLO threshold, and the discharge load is activated. If EN is not at logic low level, the circuit will be disabled to reduce the standby current.

Increasing VIN above UVLO upper threshold while EN is already higher than its upper threshold The SQ24801CDCC fully turns on MOSFET with the normal slew rate and output turn-on delay time.

### Slew Rate Control

The SQ24801CDCC integrates programmable output ramp control to reduce inrush current during start up. The inrush current caused by capacitor charging will be limited, so the SQ24801CDCC can be used in hot swapping applications.

The slew rate can be adjusted by using the different capacitor between SR pin and ground. The slew rate can be calculated using the following equation:

$$\text{Slew Rate} = \frac{K_{SR}}{C_{SR}} (V/s) \quad (1)$$

Where  $K_{SR}$  is a constant that is defined slew rate control (see Electrical Characteristics), and  $C_{SR}$  is the capacitor between the SR pin and ground. The slew rate of the device will always be the lower of the default slew rate and the adjusted slew rate. Therefore, if the  $C_{SR}$  is not large enough to decrease the slew rate more than the specified default value, the slew rate of the device will be the default value. The SR pin can be left floating if the slew rate does not need to be decreased. To make the power FET work within SOA limits in

output short circuit condition, the capacitance value of  $C_{SR}$  should not exceed 22nF.

### Over-Current Protection

The SQ24801CDCC device is equipped with an over-current protection (OCP) that helps protect the part and the system from a high current event which exceeds the expected operational current.

OCP trip current setting can be calculated using the following equation:

$$I_{TRIP} = \frac{1040}{R_{OCP}+60} + 2.9 \quad (2)$$

$I_{TRIP}$  is OCP trip current in Ampere.

$R_{OCP}$  is the programmable resistor in kΩ.

In the event that the current from the VIN pin to the VOUT pin exceeds the OCP threshold for longer than the blanking time, the MOSFET will shut down and the PG pin is driven low. The part remains latched in the Fault state until EN is toggled or VCC supply voltage is cycled, at which point the MOSFET will be turned on in a controlled fashion with the normal output turn-on delay and slew rate. The over-current Protection trip threshold is determined by the resistance between the OCP pin and ground. If no over-current protection is needed, then the OCP pin should be tied to GND; if the OCP protection is disabled in this way, the short-circuit protection will still remain active.

### Short-Circuit Protection

The SQ24801CDCC device is equipped with a short-circuit protection that helps protect the part and the system from a sudden high-current event, such as the output (VOUT) being hard-shorted to ground.

Once active, the circuitry monitors the voltage difference between the VIN pin and the VOUT pin. When the difference is equal to the short-circuit protection threshold voltage, the MOSFET is turned off and the load bleed is activated. The device remains off and is latched in the Fault state until EN is toggled or VCC supply voltage is cycled, at which point the MOSFET will be turned on in a controlled fashion with the normal output turn-on delay and slew rate. The short circuit protection feature protects the device from hard shorts.

### Thermal Shutdown

The thermal shutdown of the SQ24801CDCC devices protects the part from internally or externally generated excessive temperatures. This circuitry is disabled when EN is not active to reduce standby current. When an over-temperature condition is detected, the MOSFET is immediately turned off and the load bleed is activated.

The part comes out of thermal shutdown when the junction temperature decreases to a safe operating temperature as dictated by the thermal hysteresis. Upon exiting a thermal shutdown state, and if EN remains active, the MOSFET will be turned on in a controlled fashion with the normal output turn-on delay and slew rate.

### Power Good

The active-high, open-drain power good (PG) output is useful for indicating the gate of the MOSFET is fully charged. An external greater than or equal to 100 kΩ resistor pull up to the external supply voltage (VCC) enables PG goes active-high. If the power good feature is not used in the application, the PG pin should be tied to GND.

### Capacitive Load

Peak inrush current related to initial inrush current application load capacity charging should be maintained below the specified  $I_{MAX}$ . CL (capacitive load) should be less than  $C_{MAX}$  that can be determined by the following equation:

$$C_{MAX} = \frac{I_{MAX}}{SR} \quad (3)$$

Where  $I_{MAX}$  is the maximum load current, and SR is the typical default slew rate when the  $C_{SR}$  = open.

### PCB Layout Guide

1. For all applications, a 10µF or greater ceramic decoupling capacitor is recommended between IN terminal and GND. For hot-plug applications, where input power path inductance is negligible, this capacitor can be eliminated/minimized.
2. The optimum placement of decoupling capacitor is closest to the IN and GND terminals of the device. Care must be taken to minimize the loop area formed by the bypass-capacitor connection, the IN terminal, and the GND terminal of the IC.
3. Locate support components  $C_{VCC}$ ,  $C_{SR}$  and  $R_{OCP}$  close to their connection pin. Connect the other end of the component to the GND with shortest trace length.
4. Protection devices such as TVS, snubbers, capacitors, or diodes should be placed physically close to the device they are intended to protect, and routed with short traces to reduce inductance. For example, a protection Schottky diode is recommended to address negative transients due to switching of inductive loads, and it should be physically close to the OUT pins.

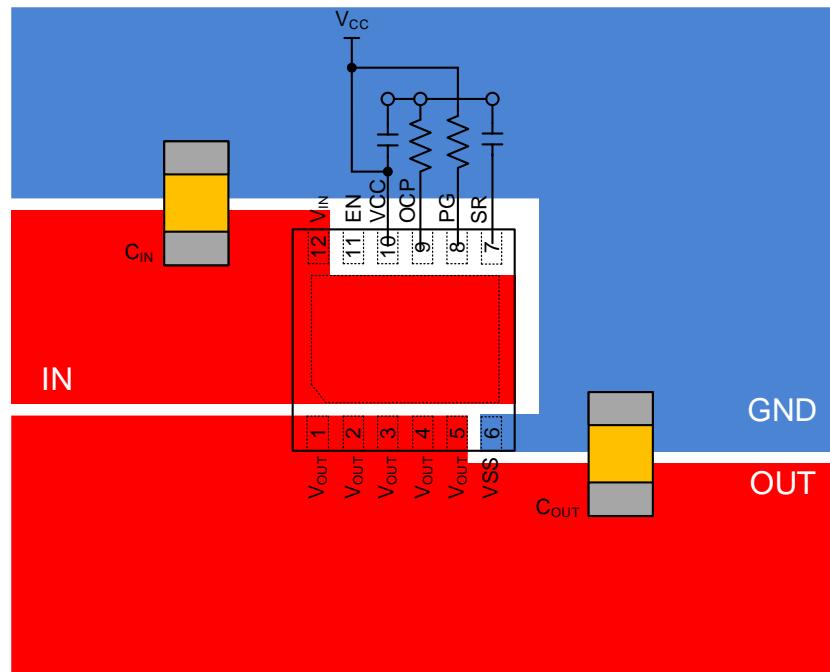


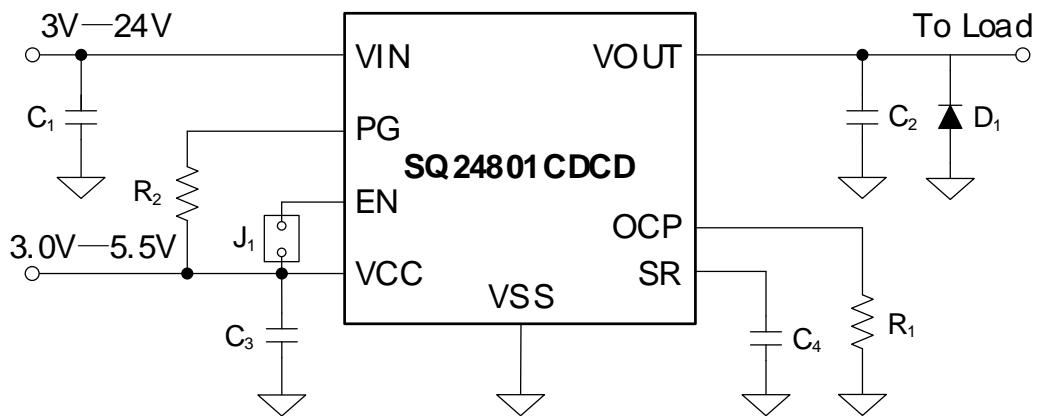
Figure 5. PCB Layout Suggestion



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Schematic

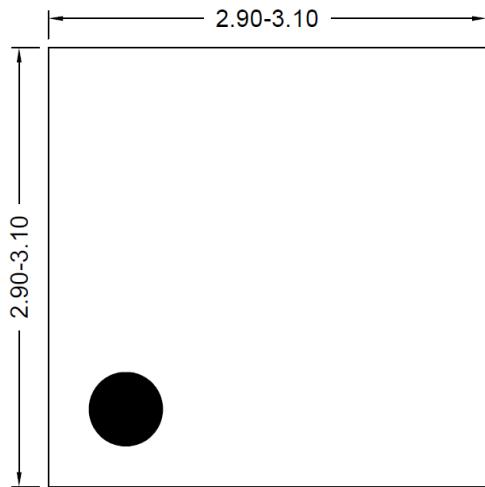
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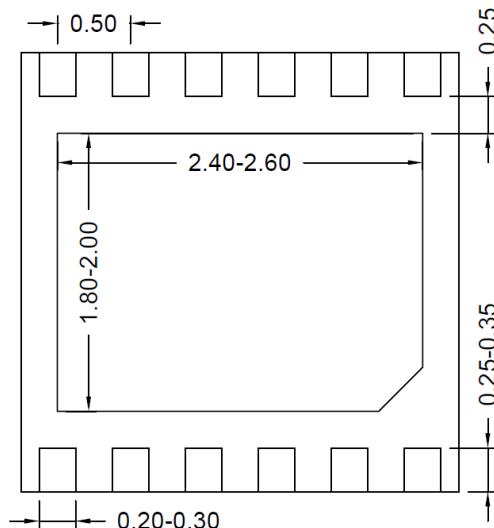
### BOM List

Reference Designator	Description	Part Number	Manufacturer
C <sub>1</sub>	10 $\mu$ F/50V, $\pm$ 10%, X5R, 1206	C3216X5R1H106KT00N	TDK
C <sub>2</sub>	10 $\mu$ F/50V, $\pm$ 10%, X5R, 1206	C3216X5R1H106KT00N	TDK
C <sub>3</sub>	1 $\mu$ F/50V, $\pm$ 10%, X5R, 0603	GRM188R61H105KAALD	Murata
C <sub>4</sub>	10nF/50V, $\pm$ 10%, X7R, 0603	GRM188R72A103KA01D	Murata
R <sub>1</sub>	0 $\Omega$ , 1%, 0.1W, 0603	RC0603FR-070RL	YAGEO
R <sub>2</sub>	100k $\Omega$ , 1%, 0.1W, 0603	RC0603FR-07100KL	YAGEO
D <sub>1</sub>	Schottky	SS54	Any
J <sub>1</sub>	Jumper, 2x1, Gold		Any

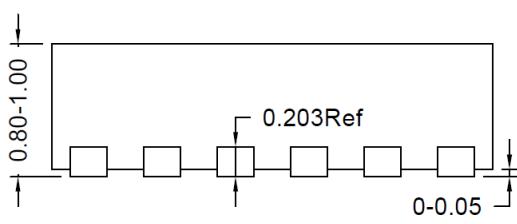
## DFN3x3-12 Package outline



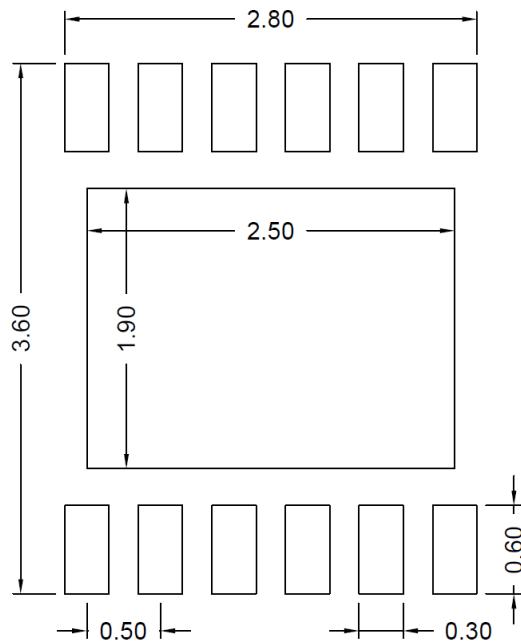
**Top View**



**Bottom view**



**Front View**

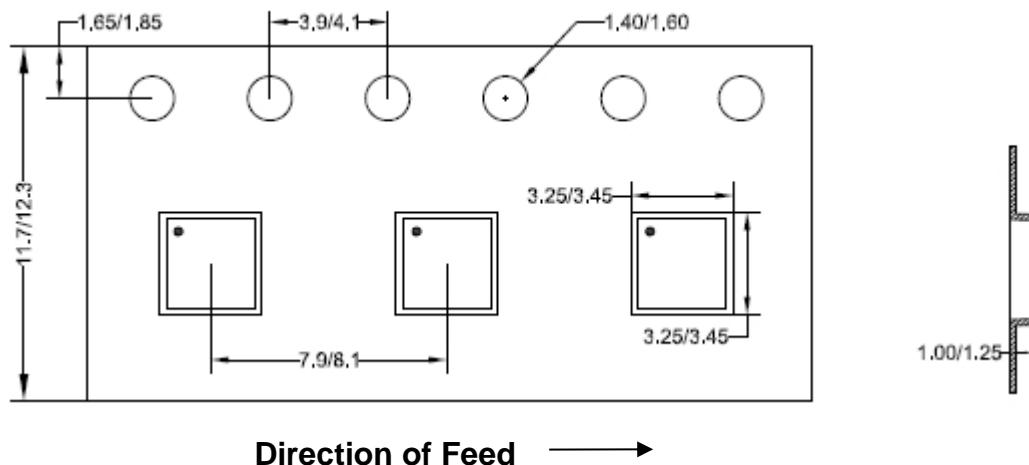


**Recommended PCB layout  
(only for reference)**

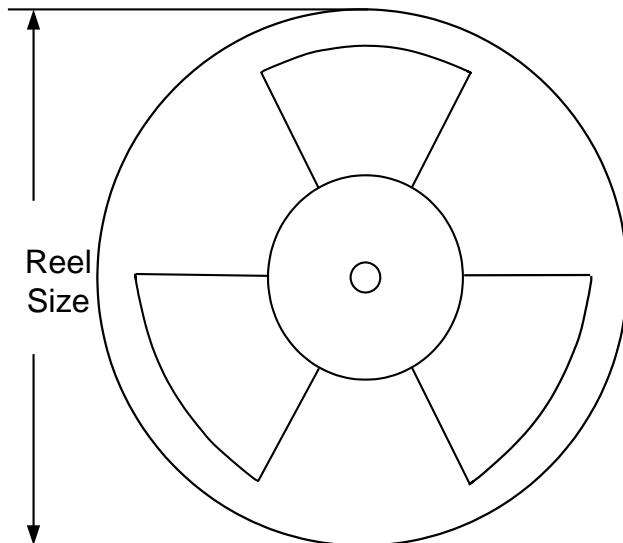
*Notes: All dimension in millimeter and exclude mold flash & metal burr.*

## Tape and Reel Information

### 1. Tape dimensions and Pin1 orientation



### 2. Reel dimensions



Package type	Tape width (mm)	Pocket pitch(mm)	Reel size (Inch)	Trailer length(mm)	Leader length (mm)	Qty per reel
DFN3x3	12	8	13"	400	400	5000

## Revision History

The revision history provided is for informational purpose only and is believed to be accurate, however, not warranted. Please make sure that you have the latest revision.

Date	Revision	Change
Aug. 22, 2024	Revision 1.0A	<ol style="list-style-type: none"><li>1. Add more detailed description for Slew Rate Control (Page 11)</li><li>2. C4 changed from 100nF to 10nF in BOM list (page 13)</li></ol>
June 14, 2024	Revision 1.0	Initial Release

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