

### High Efficiency Fast Response, 6A, 23V Input Synchronous Step Down Regulator

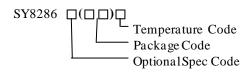
### **General Description**

The SY8286C develops a high efficiency synchronous step-down DC/DC regulator capable of delivering 6A current. The device integrates main switch and synchronous switch with very low  $R_{\rm DS(ON)}$  to minimize the conduction loss.

The SY8286C has a fixed 5V DC/DC output and integrates a 5V 100mA LDO with bypass switch and individual enable control.

The SY8286C operates over a wide input voltage range from 5.5V to 23V. The DC/DC regulator adopts the instant PWM architecture to achieve fast transient responses for high step down applications and high efficiency at light loads. In addition, it operates at pseudo-constant frequency of 600kHz under heavy load conditions to minimize the size of inductor and capacitor.

### **Ordering Information**



Ordering Number	Package type	Note
SY8286CRAC	QFN3x3-20	1

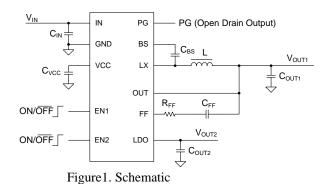
### **Features**

- Low  $R_{DS(ON)}$  for Internal Switches (Top/Bottom): 38/19 m $\Omega$
- Wide Input Voltage Range: 5.5-23V
- 5V LDO with Individual Enable Control and Bypass Switch
- Integrated Bypass Switch: 1.5Ω
- Instant PWM Architecture to Achieve Fast Transient Responses
- Internal 1.2ms Soft-start Limits the Inrush Current
- Pseudo-constant Frequency: 600kHz.
- 6A Output Current Capability
- +/-1.5% Output Voltage Accuracy
- Power Good Indicator
- Output Discharge Function
- Output Current Limit Protection
- Short Circuit Latch Off Protection
- Output Over Voltage Latch Off Protection
- Input UVLO
- Over Temperature Protection
- RoHS Compliant and Halogen Free
- Compact Package: QFN3x3-20

### **Applications**

- LCD-TV/Net-TV/3DTV
- Set Top Box
- Notebook
- High Power AP

## **Typical Applications**



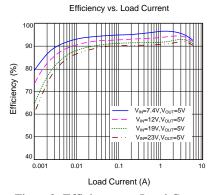
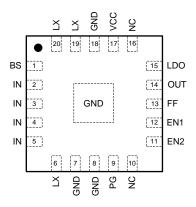


Figure 2. Efficiency vs. Load Current



## Pinout (top view)



(QFN3x3-20)

Top Mark: AWWxyz, (Device code: AWW, x=year code, y=week code, z= lot number code)

Pin Name	Pin Number	Pin Description
BS	1	Boot-strap pin. Supply high side gate driver. Decouple this pin to LX pin with a 0.1µF ceramic capacitor.
IN	2,3,4,5	Input pin. Decouple this pin to GND pin with at least a 10µF ceramic cap.
LX	6,19,20	Inductor pin. Connect this pin to the switching node of inductor.
GND	7,8,18,EP	Ground pin.
PG	9	PG is an open-drain output pin. This pin is externally pulled high when the output voltage is within 90% to 120% of regulation voltage range. Otherwise this pin is internally pull low.
NC	10,16	Not connected.
EN2	Enable control of the IC and internal LDO. Pull this pin high to turn on the internal LDO. Do not leave this pin floating.	
EN1	12	Enable control of the DC/DC regulator. Pull this pin high to turn on the regulator. Do not leave this pin floating.
FF	13	Output feed forward pin. Connect RC network from the output to this pin.
OUT	OUT  Output pin. Connect to the output of DC/DC regulator. The pin also provide bypass input for internal LDO.	
LDO	15	5V LDO output. Decouple this pin to ground with at least a 4.7μF capacitor.
VCC	17	Internal 3.3V LDO output. Power supply for internal analog circuits and driving circuit. Decouple this pin to GND with a 2.2µF ceramic capacitor.



# **Block Diagram**

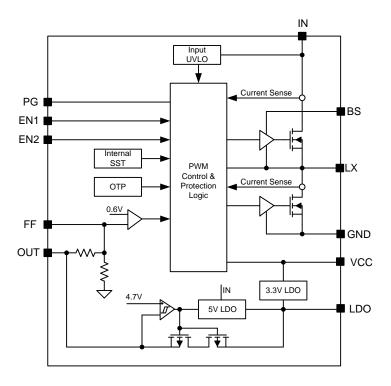


Figure 3. Block Diagram

Absolute Maximum Ratings (Note 1)	
IN, LX, PG	25V
BS-LX	4V
EN1, EN2	
VCC	
LDO	
OUT	
FF	6V
Power Dissipation,	
PD @ TA = 25°C QFN3x3-20	3.3W
Package Thermal Resistance (Note 2)	
$\theta$ Ja, QFN3x3-20	30°C/W
$\theta$ JC, QFN3x3-20	4.5°C/W
Junction Temperature Range	
Lead Temperature (Soldering, 10 sec.)	260°C
Storage Temperature Range	
Dynamic LX Voltage in 10ns Duration	IN+3V to GND-5V
<b>Recommended Operating Conditions</b> (Note 3)	
Supply Input Voltage	5.5V to 23V
Junction Temperature Range	
Ambient Temperature Range	



### **Electrical Characteristics**

 $(V_{IN} = 12V, C_{OUT} = 100 \mu F, T_A = 25 ^{\circ}C, I_{OUT} = 1A, unless otherwise specified)$ 

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Input Voltage Range	V <sub>IN</sub>	Test Conditions	5.5	тур	23	V
Quiescent Current	I <sub>Q</sub>	IOUT=0, V <sub>OUT</sub> =V <sub>SET</sub> ×105%	3.3	108	145	μA
Shutdown Current 1	I <sub>SHDN1</sub>	EN1=0, EN2=1		59	70	μΑ
Shutdown Current 2	I <sub>SHDN2</sub>	EN1=0, EN2=0		6	10	μΑ
Output Voltage Set-point	V <sub>SET</sub>	LIVI-0, LIV2-0	5.023	5.1	5.177	V
Top FET RON	R <sub>DS(ON)1</sub>		3.023	38	3.177	mΩ
Bottom FET RON	R <sub>DS(ON)1</sub>			19		mΩ
Output Discharge Current	I <sub>DIS</sub>			70		mA
HSFET FET Current Limit	I <sub>LMT,HSFET</sub>		12	70		A
Bottom FET Current Limit	I <sub>LMT,LSFET</sub>		8			A
Soft-start Time	tss		0	1.2		ms
EN Rising Threshold	V <sub>ENH</sub>		1	1.2		V
EN Falling Threshold	VENL		-		0.4	V
Input UVLO Threshold	V <sub>UVLO</sub>				5	V
UVLO hysteresis	V <sub>HYS</sub>			0.3		V
Switching Frequency	Fosc		510	600	690	kHz
Min ON Time	T <sub>ON,MIN</sub>	V <sub>IN</sub> =V <sub>INMAX</sub>	310	50	070	ns
Min OFF Time	T <sub>OFF,MIN</sub>	V IN— V INMAX		150		ns
VCC Output	V <sub>CC</sub>	V <sub>IN</sub> =4V	3.2	3.3	3.4	V
Output Over Voltage	* CC					
Threshold		V <sub>OUT</sub> rising	115	120	125	$%V_{SET}$
Output Over Voltage				_		
Hysteresis				2		$% V_{SET}$
Output OVP Delay				20		μs
Output Under Voltage				<b>~</b> 0		·
Protection Threshold			45	50	55	$% V_{SET}$
Output UVP Delay				200		us
Power Good Threshold	$V_{PG}$	V <sub>FB</sub> rising (good)	80	84	88	% V <sub>REF</sub>
Power Good Hysteresis	$V_{PG, HYS}$			2		% V <sub>REF</sub>
D. C. ID.I.	t <sub>PG, RISING</sub>	Low to high		200		μs
Power Good Delay	t <sub>PG, FALLING</sub>	High to low		10		μs
LDO Output Voltage	$V_{ m LDO}$	V <sub>IN</sub> =12V, no load	4.85	5	5.15	V
LDO Dropout Voltage	$V_{DROPOUT}$	I <sub>LDO</sub> =100mA		200		mV
LDO Output Current	_		150		200	
Limit	$I_{LMTLDO}$		150		300	mA
Bypass Switch RON	$R_{BYP}$			1.5		Ω
Bypass Switch Turn-on			15	47	4.0	V
Voltage	$V_{BYP}$		4.5	4.7	4.9	V
Bypass Switch				0.2		V
Switchover Hysteresis				0.2		V
Bypass Switch OVP				120		$%V_{LDO}$
Thermal Shutdown	$T_{SD}$			150		°C
Temperature	1 SD			130		C
Thermal Shutdown	THYS			15		°C
Hysteresis	11115			13		





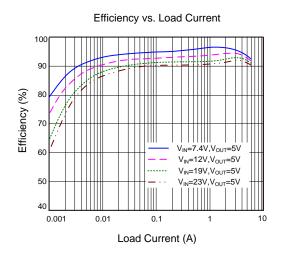
Note 1: Stresses beyond the "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability

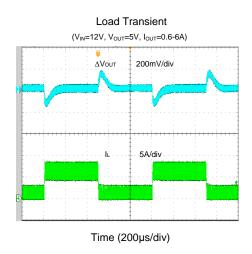
**Note 2**:  $\theta_{JA}$  is measured in the natural convection at  $T_A = 25^{\circ}\text{C}$  on a four-layer Silergy Evaluation Board.

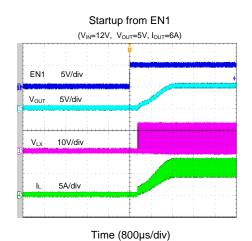
**Note 3:** The device is not guaranteed to function outside its operating conditions.

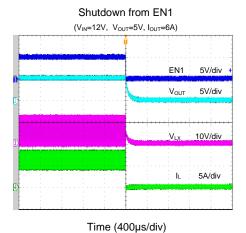


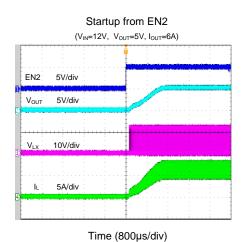
# **Typical Performance Characteristics**

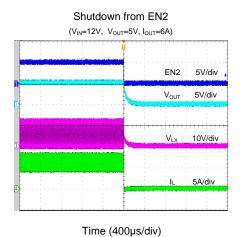






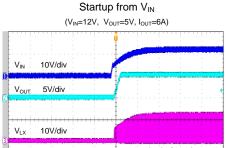






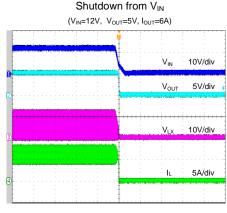






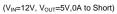
Time (4ms/div)

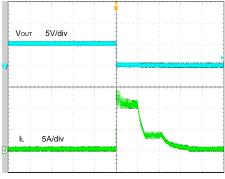
5Å/div



Time (4ms/div)

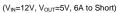
#### **Short Circuit Protection**

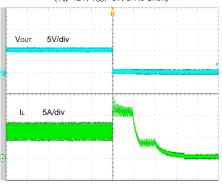




Time (100µs/div)

#### **Short Circuit Protection**

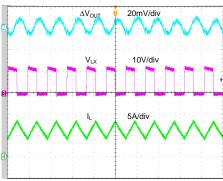




Time (100µs/div)

#### Output Ripple

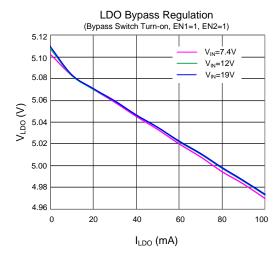
 $(V_{IN}\!\!=\!\!12V,\,V_{OUT}\!\!=\!\!5V,\,I_{OUT}\!\!=\!\!6A)$ 

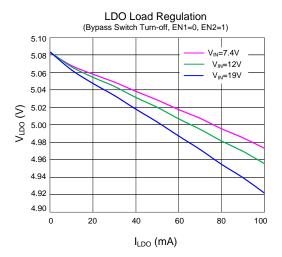


Time (2µs/div)













## **Operation**

The SY8286C develops a high efficiency synchronous step-down DC/DC regulator capable of delivering 6A current. The device integrates main switch and synchronous switch with very low  $R_{\rm DS(ON)}$  to minimize the conduction loss.

The SY8286C has a fixed 5V DC/DC output and integrates a 5V 100mA LDO with bypass switch and individual enable control.

The SY8286C operates over a wide input voltage range from 5.5V to 23V. The DC/DC regulator adopts the instant PWM architecture to achieve fast transient responses for high step down applications and high efficiency at light loads. In addition, it operates at pseudo-constant frequency of 600kHz under heavy load conditions to minimize the size of inductor and capacitor.

## **Applications Information**

Because of the high integration in the SY8286C, the application circuit based on this regulator is rather simple. Only input capacitor  $C_{\rm IN}$ , output capacitor  $C_{\rm OUT}$  and output inductor L need to be selected for the targeted applications specifications.

#### **Input Capacitor CIN:**

The ripple current through input capacitor is calculated as:

$$\boldsymbol{I}_{_{\text{CIN}\_\text{RMS}}} = \boldsymbol{I}_{_{\text{OUT}}} \cdot \sqrt{D(1-D)} \cdot$$

To minimize the potential noise problem, place a typical X5R or better grade ceramic capacitor really close to the IN and GND pins. Care should be taken to minimize the loop area formed by C<sub>IN</sub>, and IN/GND pins. In this case, a 10uF low ESR ceramic capacitor is recommended.

#### **Output Capacitor Cout:**

The output capacitor is selected to handle the output ripple noise requirements. Both steady state ripple and transient requirements must be taken into consideration when selecting this capacitor. For most applications, an X5R or better grade ceramic capacitor greater than  $66\mu F$  capacitance can work well. The capacitance derating with DC voltage must be considered.

#### **Output Inductor L:**

There are several considerations in choosing this inductor.

1) Choose the inductance to provide the desired ripple current. It is suggested to choose the ripple current to be about 40% of the maximum output current. The inductance is calculated as:

$$L = \frac{V_{\text{OUT}}(1 - V_{\text{OUT}}/V_{\text{IN,MAX}})}{F_{\text{SW}} \times I_{\text{OUT,MAX}} \times 40\%}$$

Where  $F_{SW}$  is the switching frequency and  $I_{OUT,MAX}$  is the maximum load current.

The SY8286C regulator IC is quite tolerant of different ripple current amplitude. Consequently, the final choice of inductance can be slightly off the calculation value without significantly impacting the performance.

 The saturation current rating of the inductor must be selected to be greater than the peak inductor current under full load conditions.

$$I_{SAT, \, MIN} > I_{OUT, \, MAX} + \frac{V_{OUT}(1\text{-}V_{OUT}/V_{IN, MAX})}{2 \cdot F_{SW} \cdot L}$$

3) The DCR of the inductor and the core loss at the switching frequency must be low enough to achieve the desired efficiency requirement. It is desirable to choose an inductor with DCR<10m $\Omega$  to achieve a good overall efficiency.

#### Soft-start

The SY8286C has a built-in soft-start to control the rise rate of the output voltage and limit the input current surge during IC start-up. The typical soft-start time is 1.2ms.

#### **Enable Operation**

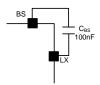
The device contains two enable pins to control the LDO and buck regulator. Driving EN2 pin high (>1V) enables the IC and the LDO. To enable the buck regulator, both EN1 and EN2 should be driven high. When EN1 and EN2 are driven below 0.4V, the device will be shut down, reducing input current <10uA. For automatic start-up, connect enable pins to VIN directly or through a 100k resistor.

#### **External Bootstrap Cap**

This capacitor provides the gate driver voltage for internal high side MOSEFET. A 100nF low ESR ceramic capacitor connected between BS pin and LX pin is recommended.







#### VCC LDO

The 3.3V VCC LDO provides the power supply for internal control circuit. Bypass this pin to ground with a  $2.2\mu f$  ceramic capacitor.



#### **LDO**

Internal 5V LDO output. Power supply for internal analog circuit and driving circuit. This pin should be bypassed to ground with a 4.7uf ceramic capacitor. This pin is also capable sourcing 100mA current for external load.

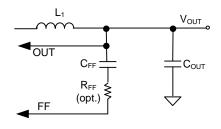


#### **Power Good Indication**

PG is an open-drain output pin. This pin will pull to ground if output voltage is lower than 90% of regulation voltage. Otherwise this pin will go to a high impedance state.

#### **Load Transient Considerations:**

The SY8286C regulator IC adopts the instant PWM architecture to achieve good stability and fast transient responses. In applications with high step load current, adding an RC network  $R_{FF}$  and  $C_{FF}$  between OUT and FF pin may further speed up the load transient responses.



#### **Layout Design:**

The layout design of SY8286C regulator is relatively simple. For the best efficiency and to minimize noise problem, we should place the following components close to the IC:  $C_{IN}$ ,  $C_{LDO}$ ,  $C_{VCC}$ , L.

- It is desirable to maximize the PCB copper area connecting to GND pin to achieve the best thermal and noise performance. If the board space allowed, a ground plane is highly desirable.
- 2) C<sub>IN</sub> must be close to Pins IN and GND. The loop area formed by C<sub>IN</sub> and GND must be minimized.
- The PCB copper area associated with LX pin must be minimized to avoid the potential noise problem.
- 4) If the system chip interfacing with the enable pin has a high impedance state at shutdown mode and the IN pin is connected directly to a power source such as a Li-Ion battery, A 1Mohm pull down resistor should be placed between the enable pin and GND pins to prevent the noise from falsely turning on the regulator at shutdown mode.



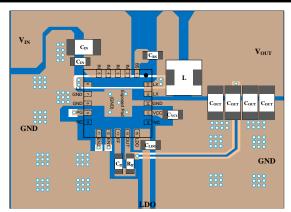
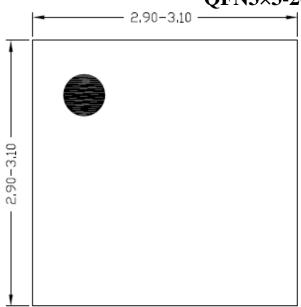
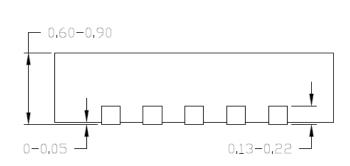


Figure 4. PCB Layout Suggestion



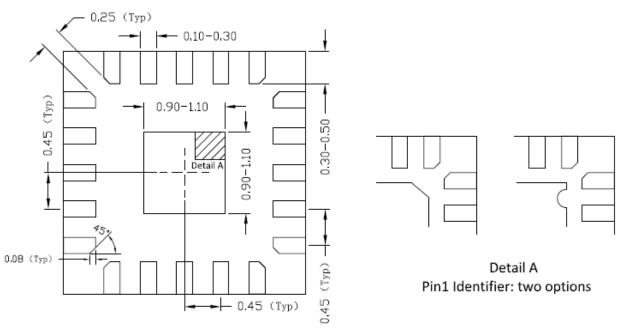
# QFN3×3-20 Package Outline





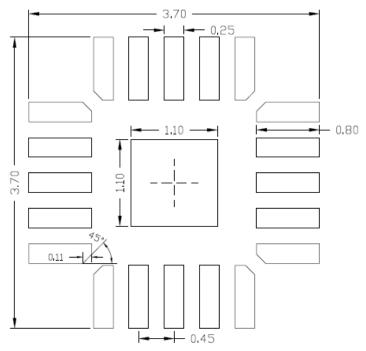
# Top view

## **Side view**



**Bottom view** 





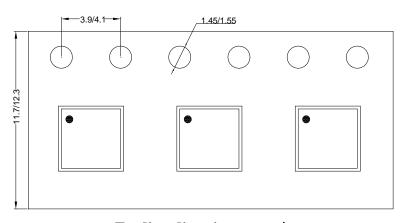
Recommended PCB layout (Reference only)

Notes: All dimension in millimeter and exclude mold flash & metal burr.



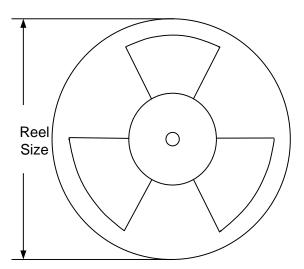
# **Taping & Reel Specification**

## 1. QFN3×3-20 taping orientation



## **Feeding direction**

## 2. Carrier Tape & Reel specification for packages



Package types	Tape width (mm)	Pocket pitch(mm)	Reel size (Inch)	Trailer length(mm)	Leader length (mm)	Qty per reel
QFN3×3	12	8	13"	400	400	5000

### 3. Others: NA



# **Revision History**

The revision history provided is for informational purpose only and is believed to be accurate, however, not warranted. Please make sure that you have the latest revision.

Date	Revision	Change	
Nov 21, 2019	Revision 0.9I	Update "Power Good Threshold" in EC table	
Jan 17, 2018	Revision 0.9H	Add "Revision History" in page 13, Important Notice in page 14.	
July 18, 2017	Revision 0.9G	Update "Power good indicator" in page 10.	
May 30, 2016	Revision 0.9F	Update package outline (Recommended PCB layout)	
May 5, 2016	Revision 0.9E	Update package outline (side view)	
April 13, 2016	Revision 0.9D	Update the data in EC table (I $_{Q}$ , I $_{SHDN}$ , I $_{Discharge}$ , V $_{ENH}$ , V $_{HYS}$ , V $_{PG}$ threshold/ HYS, V $_{LDO}$	
Dec 31, 2015	Revision 0.9C	Update general description.	
Nov 26, 2015	Revision 0.9B	<ol> <li>Features: Change "+/-1% Internal Vref" to "+/-1.5% output voltage accuracy"</li> <li>PG pin description</li> <li>Enable Operation in Application Information(Page9)</li> </ol>	
July 15, 2015	Revision 0.9A	Add " dynamic LX voltage" in Abs (Page3)	
July 12, 2015	Revision 0.9	Initial Release	



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