# **Application Note: SY8288A**



High Efficiency Fast Response 8A, 23V Input Synchronous Step Down Regulator

# **General Description**

The SY8288A develops a high efficiency synchronous step-down DC/DC regulator capable of delivering 8A current. The device integrates a main switch and a synchronous switch with very low  $R_{DS(ON)}$  to minimize the conduction loss.

The SY8288A operates over a wide input voltage range from 4V to 23V. The DC/DC regulator adopts the instant PWM architecture to achieve fast transient responses for high step down applications and high efficiency at light load. The device provides various protection features for reliable operation. In addition, it operates at pseudo-constant frequency of 600 kHz to minimize the size of inductor and capacitor.

**Temperature** Code

Optional Spec Code

Note

Package type

QFN3x3-20

Package Code

# **Ordering Information**

SY8288 □(□□)□

Ordering Number

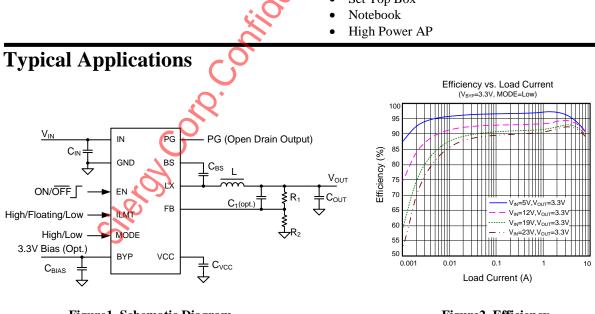
SY8288ARAC

## Features

- Low R<sub>DS(ON)</sub> for Internal Switches (Top/Bottom): 22 mΩ /11 mΩ
- Wide Input Voltage Range: 4-23V
- Instant PWM Architecture to Achieve Fast Transient Responses
- Internal 2ms Soft-start Limits the Inrush Current
- Pseudo-constant Frequency: 600kHz
- 8A Output Current Capability
- +/-1% Internal Reference Voltage
- PFM/PWM Selectable Light Load Operation Mode
- Optional Bypass Input
- Power Good Indicator
- Output Discharge Function
- Output Current Limit Protection
- Hiccup Mode Output Short Circuit Protection
- OutputOver Voltage Protection
- Input UVLO
- Over Temperature Protection with Auto Recovery
- RoHS Compliant and Halogen Free
- Compact Package: QFN3x3-20

## Applications

- LCD-TV/Net-TV/3DTV
- Set Top Box



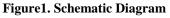
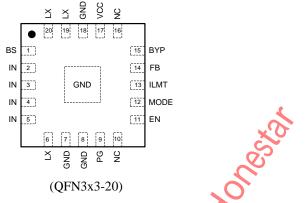


Figure2. Efficiency

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## Pinout (top view)



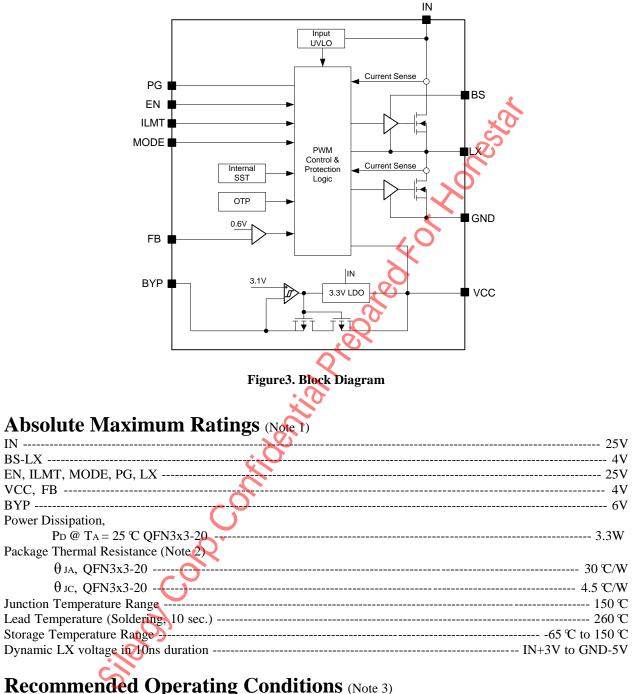
Top Mark: BDJxyz, (Device code: BDJ, x=year code, y=week code, z= lot number code)

Pin Number	Pin Description
1	Boot-strap pin. Supply high side gate driver. Decouple this pin to the LX pin with
	a 0.1 µF ceramic capacitor.
2,3,4,5	Input pin. Decouple this pin to the GND pin with at least a 10 µF ceramic
	capacitor.
6,19,20	Inductor pin. Connect this pin to the witching node of the inductor.
7,8,18,EP	Ground pin.
9	Power good Indicator. Open-drain output when the output voltage is within 90% to
	120% of regulation point.
10, 16	Not connected.
11	Enable pin. Pull this pin high to turn on the IC. Do not leave this pin floating.
12	Operating mode selection under light load. Pull this pin low for PFM operation,
	and pull this pin high for PWM operation. Do not leave this pin floating.
13	Output current limit threshold selection.
14	Output feedbackpin. Connect to the center point of the resistor divider.
15	External 3.3 by pass power supply input. Decouple this pin to GND with a 1 µF
	ceramic capacitor. Leave this pin floating if it is not used.
17	Internal 3.3V LDO output. Power supply for internal analog circuits and driving
	circuit. Decouple this pin to GND with a 2.2 µF ceramic capacitor.
	1 2,3,4,5 6,19,20 7,8,18,EP 9 10, 16 11 12 13 14 15

circuit De



## **Block Diagram**



	1	0	
Supply Input Voltage			 4V to 23V
Supply input voltage			+ 10 25 1
Junction Temperature	Range		 40 °C to 125 °C
Ambient Temperature	Range		 40 °C to 85 °C
millionent remperature	Runge		40 C 10 05 C



# **Electrical Characteristics**

(VIN = 12V, TA = 25 °C unless otherwise specified)

$(V_{IN} = 12V, T_A = 25 \ C \text{ unless of}$			Ma	T	Mari	II:4
Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Input Voltage Range	V <sub>IN</sub>		4		23	V
Input UVLO Threshold	V <sub>UVLO</sub>				3.9	V
UVLO Hysteresis	V <sub>HYS</sub>			0.3		V
Quiescent Current	I <sub>Q</sub>	IOUT=0, $V_{OUT}=V_{SET} \times 105\%$		120	145	μA
Shutdown Current	I <sub>SHDN</sub>	EN=0,		6	10	μA
Feedback Reference	V <sub>REF</sub>		0.594	0.6	0.606	V
Voltage				<u></u>		
Top FET RON	R <sub>DS(ON)1</sub>			22		mΩ
Bottom FET RON	R <sub>DS(ON)2</sub>			AL		mΩ
Output Discharge Current	I <sub>DIS</sub>			70		mA
HSFET FET Current	I <sub>LMT,HSFET</sub>			24		А
Limit						
Bottom FET Current	I <sub>LMT, LSFET1</sub>	ILMT=Low	8			А
Limit		ILMT=Floating	12			А
		ILMT=High	16			А
Bottom FET Reverse	I <sub>LIM,LSFET2</sub>		1.8	2.85	3.9	А
Current Limit	200,201 212					
Soft-start Time	t <sub>SS</sub>	S.		2		ms
EN/MODE Rising	V <sub>ENH</sub>		1			V
Threshold						
EN/MODE Falling	V <sub>ENL</sub>				0.4	V
Threshold	· ENL					
ILMT Rising Threshold	V <sub>ILMTH</sub>		Vcc-0.5			V
ILMT Falling Threshold	V <sub>ILMTL</sub>				0.5	V
Switching Frequency	f <sub>OSC</sub>		510	600	690	kHz
Min ON Time	t <sub>ON,MIN</sub>	V <sub>IN</sub> =V <sub>INMAX</sub>	010	50	070	ns
Min OFF Time	t <sub>OFF.MIN</sub>			150		ns
VCC Output	V <sub>CC</sub>	~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~	3.2	3.3	3.4	V
Output Over Voltage	V <sub>OVP</sub>	V <sub>FB</sub> rising	115	120	125	%V <sub>REF</sub>
Threshold	* OVP	-Brising	115	120	125	/0 V REF
Output Over Voltage	V <sub>OVP,HYS</sub>			2		%V <sub>REF</sub>
Hysteresis	• OVP,HYS	P		2		/0 V REF
Output OVP Delay	t <sub>OVP,DLY</sub>			10		μs
Output Under Voltage	V <sub>UVP</sub>		57.5	62.5	67.5	%V <sub>REF</sub>
Protection Threshold	• UVP		51.5	02.5	07.5	∕♥ ▼ REF
Output UVP Delay				200		116
Power Good Threshold	$t_{UVP,DLY}$ V <sub>PG</sub>	V <sub>FB</sub> rising (Good)	87.5	92.5	97.5	μs %V <sub>REF</sub>
Power Good Hysteresis		* FB HSING (GOOD)	07.5	2	91.J	% V <sub>REF</sub>
Power Good Delay	V <sub>PG,HYS</sub>	Low to high		200		
	tPG,RISING	High to low		200		μs
Bypass Switch Turn-on	t <sub>PG,FALLING</sub>		2.97	3.1	3.21	μs V
Voltage	V <sub>BYP</sub>		2.97	5.1	3.21	v
Bypass Switch	V			0.2		V
	V <sub>BYP,HYS</sub>			0.2		v
Switchover Hysteresis	V			100		0/ 17
Bypass Switch OVP	V <sub>BYP,OVP</sub>			120		%V <sub>CC</sub>
Thermal Shutdown	T <sub>SD</sub>			150		C
Temperature				1.7		~
Thermal Shutdown	Thys			15		C
hysteresis						



Note 1: Stresses beyond the "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Note 2: Package thermal resistance is measured in the natural convection at  $T_A = 25$  °C on a four-layer Silergy Evaluation Board.

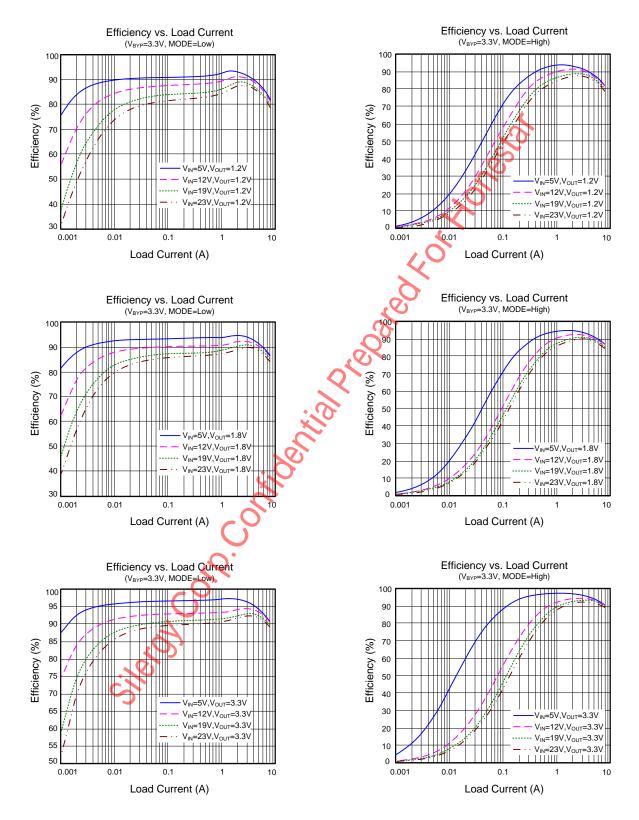
Note 3: The device is not guaranteed to function outside its operating conditions.

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# **Typical Performance Characteristics**

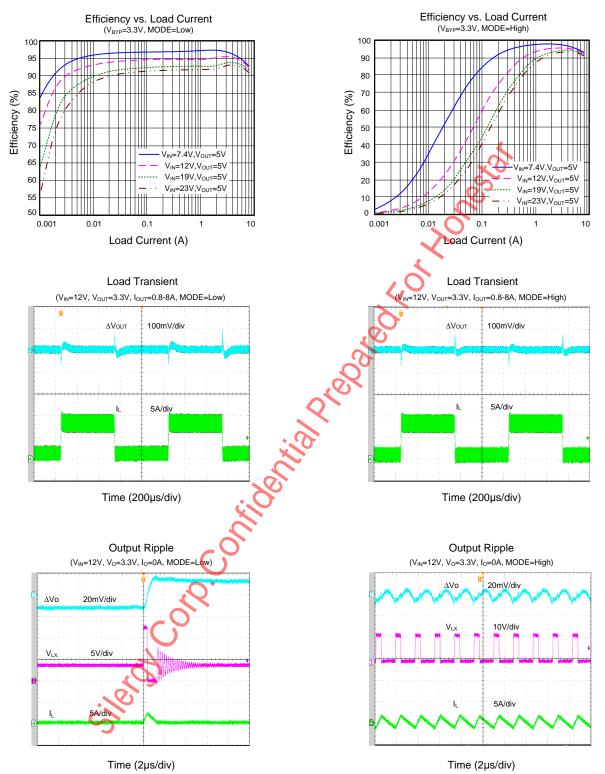
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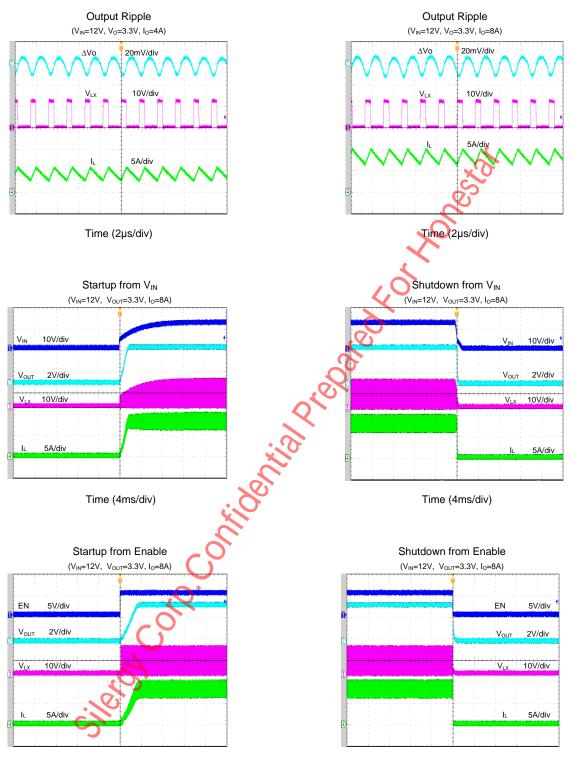
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# SY8288A



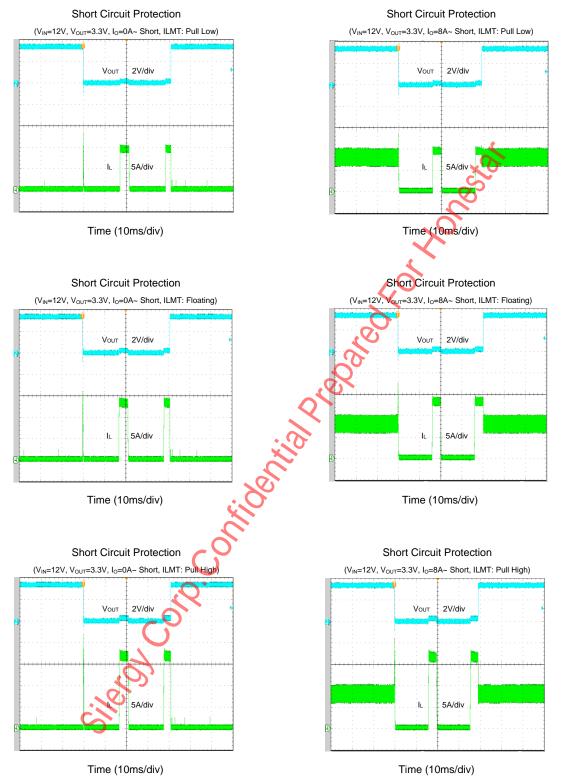




Time (2ms/div)

Time (2ms/div)







# Operation

The SY8288A develops a high efficiency synchronous step-down DC/DC regulator capable of delivering 8A current. The device integrates a main switch and a synchronous switch with very low  $R_{DS(ON)}$  to minimize the conduction loss.

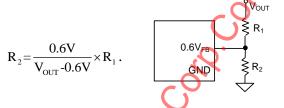
The SY8288A operates over a wide input voltage range from 4V to 23V. The DC/DC regulator adopts the instant PWM architecture to achieve fast transient responses for high step down applications and high efficiency at light load. The device provides various protection features for reliable operation. In addition, it operates at pseudo-constant frequency of 600 kHz to minimize the size of inductor and capacitor.

# **Applications Information**

Because of the high integration in the SY8288A, the application circuit based on this regulator is rather simple. Only the input capacitor  $C_{IN}$ , the output capacitor Cour, the output inductor L and the feedback resistors ( $R_1$  and  $R_2$ ) need to be selected for the targeted applications specifications.

### Feedback Resistor Dividers R1 and R2

Choose  $R_1$  and  $R_2$  to program the proper output voltage. To minimize the power consumption under light load it is desirable to choose large resistance values for both  $R_1$  and  $R_2$ . A value of between  $10k\Omega$  and  $1M\Omega$  is highly recommended for both resistors. If  $V_{OUT}$  (\$ :2V,  $R_1=100k\Omega$  is chosen, then using the following equation,  $R_2$  can be calculated to be  $100k\Omega$ :



### Input Capacitor CIN

The ripple current through input capacitor is calculated as:

$$I_{\text{CIN_RMS}} = I_{\text{OUT}} \times \sqrt{D \times (1-D)}$$

To minimize the potential noise problem, place a typical X5R or better grade ceramic capacitor really close to the IN and GND pins. Care should be taken to minimize the loop area formed by  $C_{IN}$ , and IN/GND pins. In this case, a 10 µF low ESR ceramic capacitor is recommended.

## **Output Capacitor Cout**

The output capacitor is selected to handle the output ripple noise requirements. Both steady state ripple and transient requirements must be taken into consideration when selecting this capacitor. For most applications, an X5R or better grade ceramic capacitor larger than  $66 \,\mu\text{F}$  capacitance can work well. The capacitance derating with DC voltage must be considered.

### <u>Output Inductor L</u>

There are several considerations in choosing this inductor.

 Choose the inductance to provide the desired ripple current. It is suggested to choose the ripple current to be about 30~40% of the maximum output current. In PWM mode operation, 30% of the maximum output current is suggested, in order not to trigger bottom FET reverse current limit at light load condition. The inductance is calculated as:

$$L \neq V_{OUT} (1 - V_{OUT} / V_{IN,MAX})$$
  
$$f_{SW} \times I_{OUT,MAX} \times 30\%$$

Where  $f_{SW}$  is the switching frequency and  $I_{OUT,MAX}$  is the maximum load current.

The SY8288A is quite tolerant of different ripple current amplitude. Consequently, the final choice of inductance can be slightly off the calculation value without significantly impacting the performance.

2) The saturation current rating of the inductor must be selected to be greater than the peak inductor current under full load condition.

$$I_{\text{SAT, MIN}} > I_{\text{OUT, MAX}} + \frac{V_{\text{OUT}}(1\text{-}V_{\text{OUT}}/V_{\text{IN, MAX}})}{2 \times f_{\text{SW}} \times L}$$

3) The DCR of the inductor and the core loss at the switching frequency must be low enough to achieve the desired efficiency requirement. It is desirable to choose an inductor with DCR<10m $\Omega$  to achieve a good overall efficiency.

### **Current Limit Setting**

The SY8288A features both cycle-by-cycle peak and valley current limit. The high side MOSFET is turned off and low side MOSFET is turned on when peak current limit is triggered. When the valley current limit is triggered, the device will not allow high side MOSFET turning on until the valley current drops below the threshold. The valley current limit threshold

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is selectable by pulling ILMT pin low, high or leaving it floating.

#### <u>Soft-start</u>

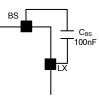
The SY8288A has a built-in soft-start to control the rise rate of the output voltage and limit the input current surge during the IC starts up. The typical soft-start time is 2ms.

### Enable Operation

Pulling the EN pin low will shut down the device. During shutdown mode, the SY8288A shutdown current drops to lower than  $10 \,\mu$ A. Driving the EN pin high will turn on the IC again.

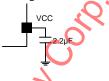
#### **External Bootstrap Capacitor**

This capacitor provides the gate driver voltage for internal high side MOSEFET. A 100nF low ESR ceramic capacitor connected between the BS pin and the LX pin is recommended.



#### VCC LDO and BYP Input

The 3.3V VCC LDO provides the power supply for internal control and drive circuit. Bypass this pin to ground with a  $2.2 \,\mu\text{F}$  ceramic capacitor. The control and drive circuit can also be powered by external 3.3V power supply. When a 3.3V external power supply is connected to the BYP pin, the VCC LDO is turned off and the switch between BYP and VCC is turned on. The overall efficiency may be improved by connecting the BYP pin to external 3.3V switching power supply. Leave the BYP pin floating if this feature is not used.



#### Power Good Indication

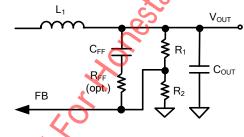
PG is an open-drain output pin. This pin will pull to ground if output voltage is lower than 90% or higher than 120% of the regulation voltage. Otherwise this pin will go to a high impedance state.

### **Light Load Operation Mode Selection**

PFM or PWM light load operation is selected by the MODE pin. Pull the MODE pin low for PFM operation, and pull this pin high for PWM operation.

#### **Load Transient Considerations**

The SY8288A adopts the instant PWM architecture to achieve good stability and fast transient responses. In applications with high step load current, adding an RC network  $R_{FF}$  and  $C_{FF}$  parallel with  $R_1$  may further speed up the load transient responses.



#### Layout Design

The layout design of SY8288A is relatively simple. For the best efficiency and minimum noise problem, the following components should be close to the IC:  $C_{IN}$ ,  $C_{VCC}$ ,  $R_1$  and  $R_2$ .

It is desirable to maximize the PCB copper area connecting to the GND pin to achieve the best thermal and noise performance. If the board space allows, a ground plane is highly desirable.

2)  $C_{IN}$  must be close to the IN and GND pins. The loop area formed by  $C_{IN}$  and GND must be minimized.

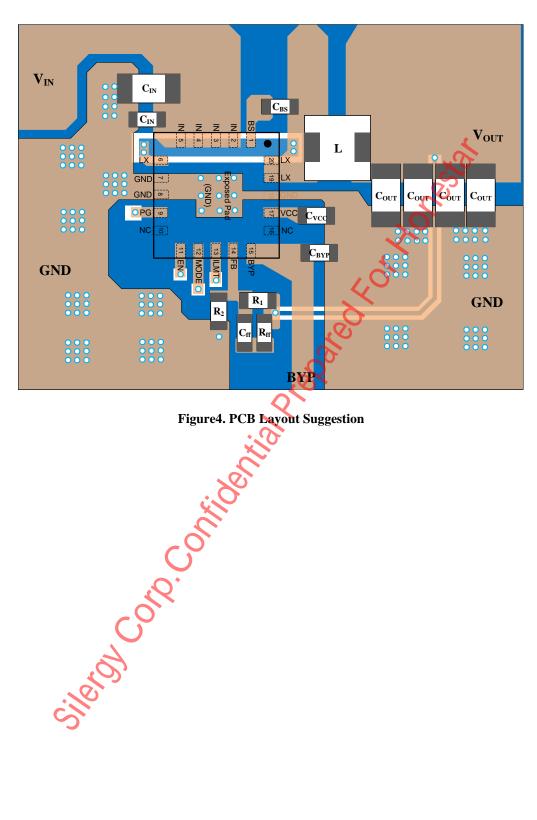
3) The PCB copper area associated with the LX pin must be minimized to avoid the potential noise problem.

4) The components  $R_1$  and  $R_2$  and the trace connecting to the FB pin must NOT be adjacent to the LX net on the PCB layout to avoid the noise problem.

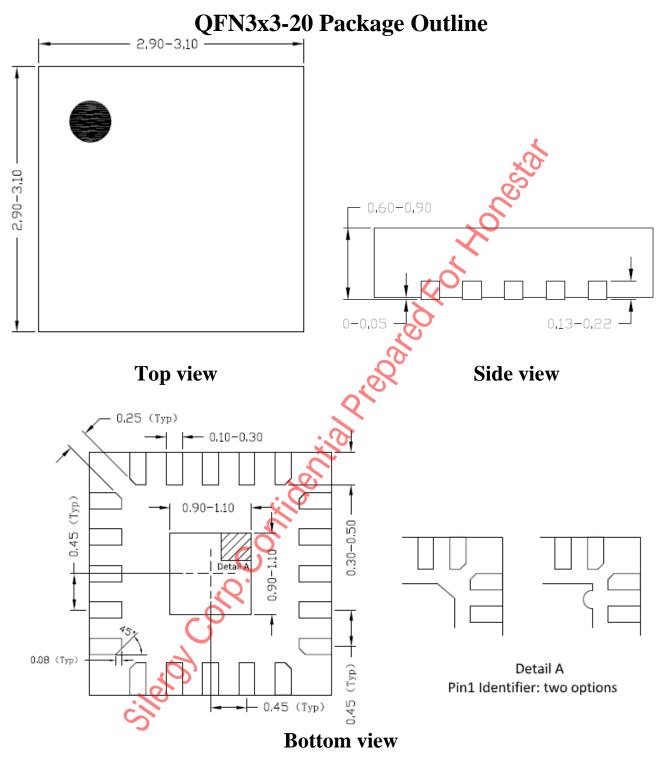
5) If the system chip interfacing with the EN pin has a high impedance state at shutdown mode and the IN pin is connected directly to a power source such as a Li-Ion battery. A 1M $\Omega$  pull down resistor should be placed between the EN and GND pins to prevent the noise from falsely turning on the regulator at shutdown mode.

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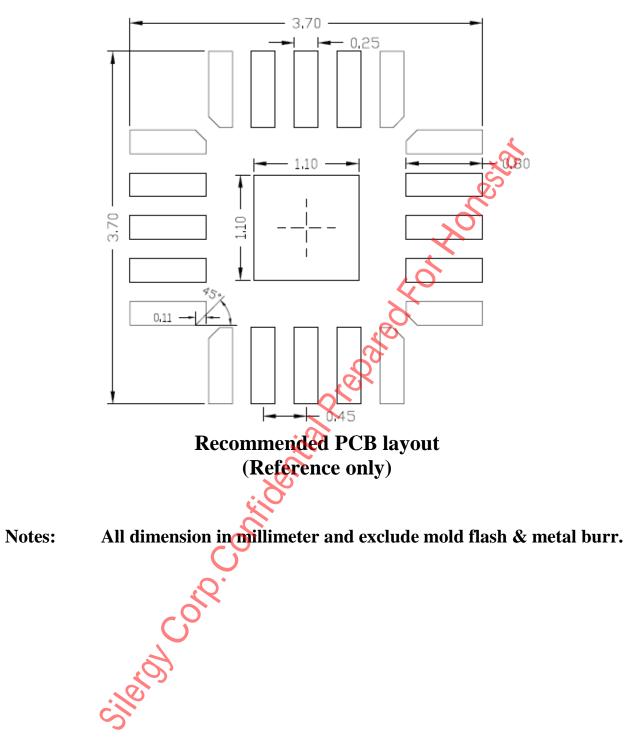








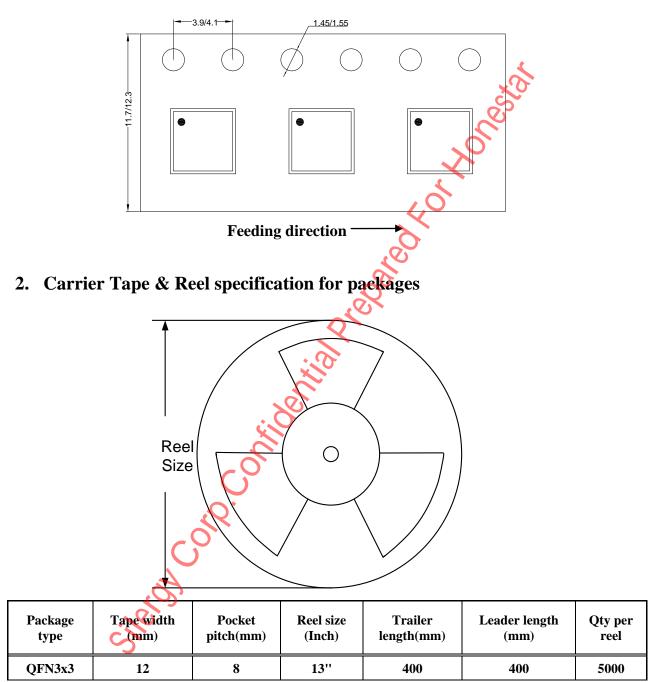






# **Taping & Reel Specification**

## 1. QFN3x3-20 taping orientation



## 3. Others: NA



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