

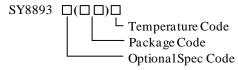
High Efficiency, 1.2MHz, 3A Synchronous Step Down Regulator

General Description

The SY8893 is a high efficiency 1.2MHz synchronous step down DC/DC regulator IC, which is capable of delivering up to 3A output currents. It can operate over a wide input voltage range from 2.5V to 5.5V and integrate main switch and synchronous switch with very low $R_{\rm DS\,(ON)}$ to minimize the conduction loss.

The SY8893 is in a space saving, low profile SOT563 package.

Ordering Information



Ordering Number	Package type	Note
SY8893ARC	SOT563	

Features

- 2.5V to 5.5V Input Voltage Range
- 50μA Low Quiescent Current
- Low Rds(on) for Internal Switches (Top/Bottom) $100m\Omega/60m\Omega$
- High Switching Frequency 1.2MHz Minimizes the External Components
- Internal Soft-start Limits the Inrush Current
- 100% Dropout Operation
- Hic-cup for Short Circuit Protection
- Output Auto Discharge Function
- RoHS Compliant and Halogen Free
- Compact Package: SOT563

Applications

- Set Top Box
- USB Dongle
- Media Player
- Smart phone

Typical Applications

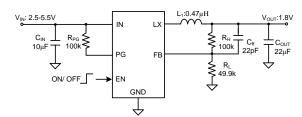


Figure 1. Schematic Diagram

Inductor and Cour Selection Table

V _{OUT} [V]	L[μH]	C _{OUT} [μF]			
		10	22	2×22	3×22
	0.47		$\stackrel{\wedge}{\simeq}$		
1.2/1.8/3.3	1.0			V	V
	2.2				

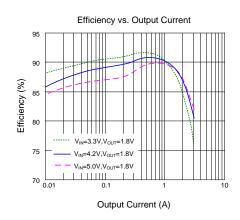
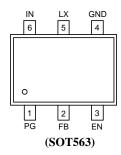


Figure 2. Efficiency vs. Output Current



Pinout (Top View)



Top Mark: fQxyz (device code: fQ, x=year code, y=week code, z=lot number code)

Pin Name	Pin Number	Pin Description
PG	1	Power good indicator (open drain output). Low if the output < 90% or the output >120% of regulation voltage; High otherwise. Connect a pull-up resistor to the input.
FB	2	Output Feedback Pin. Connect this pin to the center point of the output resistor divider (as shown in Figure 1) to program the output voltage: $V_{OUT}=0.6\times(1+R_H/R_L)$.
EN	3	Enable control. Pull high to turn on. Do not float.
GND	4	Ground pin.
LX	5	Inductor pin. Connect this pin to the switching node of inductor.
IN	6	Input pin. Decouple this pin to GND pin with at least 10µF ceramic capacitor.

Block Diagram

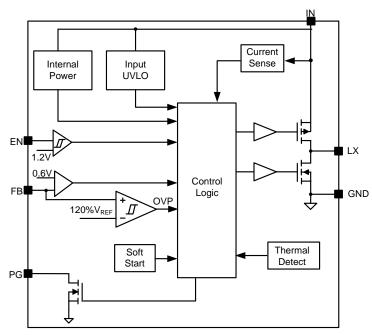


Figure 3. Block Diagram





Absolute Maximum Ratings (Note 1)

8 \ '	
Supply Input Voltage	
PG, FB, EN Voltage	
LX Voltage	
Power Dissipation, P_D @ $T_A = 25^{\circ}C$	1.4W
Package Thermal Resistance (Note 2)	
$ heta_{ m JA}$	70°C/W
$ heta_{ m JC}$	8°C/W
Junction Temperature Range	
Lead Temperature (Soldering, 10 sec.)	
Storage Temperature Range	65°C to 150°C
(*1) LX Voltage Tested Down to -3V<40ns	
(*2) LX Voltage Tested Up to +7V<40ns	
Recommended Operating Conditions (Note 3)	
<u> </u>	2.511
Supply Input Voltage	
Junction Temperature Range	



Electrical Characteristics

 $(V_{IN} = 5V, V_{OUT} = 1.8V, L = 0.47\mu H, C_{OUT} = 22\mu F, T_A = 25$ °C, unless otherwise specified)

$\begin{array}{ c c }\hline \textbf{Parameter} \\ \hline \end{array}$	Symbol	Test Conditions	Min	Тур	Max	Unit
Input Voltage Range	V_{IN}		2.5		5.5	V
Input UVLO Threshold	$V_{\rm UVLO}$				2.5	V
Input UVLO Hysteresis	V _{HYS}			0.15		V
Quiescent Current	I_Q	$V_{FB}=V_{REF}\times 105\%$		50	70	μA
Shutdown Current	I _{SHDN}	V _{EN} =0V		0.1	1	μA
Feedback Reference Voltage	V_{REF}	I _{OUT} =1A, CCM	591	600	609	mV
Output Discharge Resistance	R_{DIS}			50		Ω
Top FET R _{ON}	$R_{DS(ON)1}$			100		m Ω
Bottom FET RON	$R_{DS(ON)2}$			60		m Ω
EN Input Voltage High	$V_{\rm EN,H}$		1.2			V
EN Input Voltage Low	$V_{\mathrm{EN,L}}$				0.4	V
PG Threshold for Under Voltage Detection	$V_{PG,UVP}$			90		%V _{REF}
PG Low Delay Time for Under Voltage Detection	t _{UVP,DLY}			20		μs
PG Threshold for Over Voltage Detection	$V_{PG,OVP}$			120		%V _{REF}
PG Low Delay Time for Over Voltage Detection	t _{OVP,DLY}			20		μs
Min ON Time	t _{ON,MIN}			60		ns
Maximum Duty Cycle	D_{MAX}		100			%
Turn On Delay	t _{ON,DLY}	from EN high to LX start switching		300		μs
Soft-start Time	t_{SS}			700		μs
Switching Frequency	f_{SW}	I _{OUT} =1A, CCM		1.2		MHz
Top FET Current Limit	I _{LMT,TOP}		4			A
Thermal Shutdown Temperature	T_{SD}			160		°C
Thermal Shutdown Hysteresis	T_{HYS}			20		°C

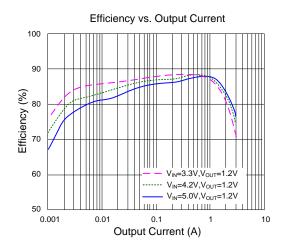
Note 1: Stresses beyond the "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

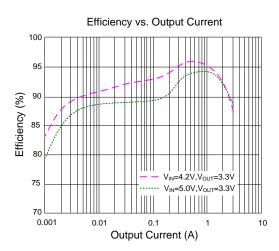
Note 2: θ_{JA} is measured in the natural convection at $T_A = 25$ °C on 2OZ four-layer Silergy evaluation board. Pin 5 of case position for SY8893ARC θ_{JC} measurement.

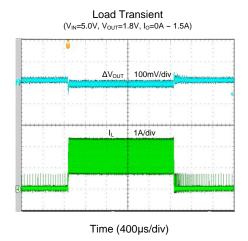
Note 3: The device is not guaranteed to function outside its operating conditions.

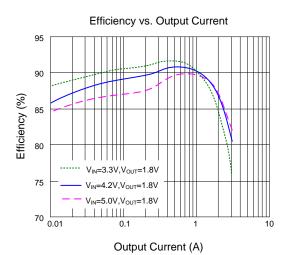


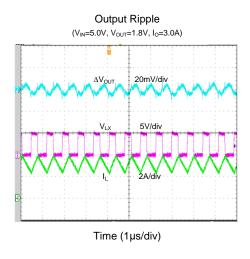
Typical Performance Characteristics

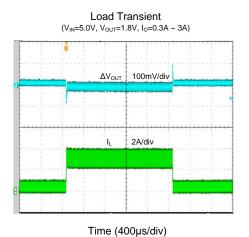








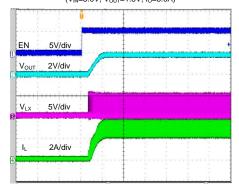






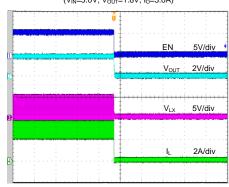


Startup from Enable (V_{IN} =5.0V, V_{OUT} =1.8V, I_{O} =3.0A)



Time (800µs/div)

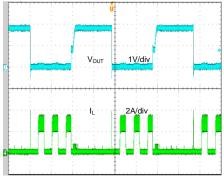
Shutdown from Enable $(V_{IN}=5.0V, V_{OUT}=1.8V, I_O=3.0A)$



Time (800µs/div)

Short Circuit Protection

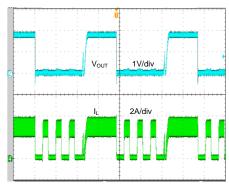




Time (4ms/div)

Short Circuit Protection

(V_{IN} =5.0V, V_{OUT} =1.8V, I_{O} =3A ~ Short)



Time (4ms/div)





Operation

The SY8893 is a high efficiency 1.2MHz synchronous step down DC/DC regulator IC, which is capable of delivering up to 3A output currents. It can operate over a wide input voltage range from 2.5V to 5.5V and integrate main switch and synchronous switch with very low R_{DS (ON)} to minimize the conduction loss.

The SY8893 is in a space saving, low profile SOT563 package.

Applications Information

Because of the high integration in the SY8893, the application circuit based on this regulator is rather simple. Only input capacitor C_{IN} , output capacitor C_{OUT} , output inductor L and feedback resistors (R_{H} and R_{L}) need to be selected for the targeted application specifications.

Feedback Resistor Dividers R_H and R_L

Choose R_H and R_L to program the proper output voltage. To minimize the power consumption under light loads, it is desirable to choose large resistance values for both R_H and R_L . A value of between $100k\Omega$ and $1M\Omega$ is highly recommended for both resistors. If R_L =120k Ω is chosen, then R_H can be calculated to be:

$$R_{\rm H} = \frac{(V_{\rm OUT} - 0.6\,V) \cdot R_{\rm L}}{0.6V}$$

Input Capacitor C_{IN}

A typical X5R or better grade ceramic capacitor with 6.3V rating and greater than 10uF capacitance is recommended. To minimize the potential noise problem, this ceramic capacitor should be placed really close to the IN and GND pins. Care should be taken to minimize the loop area formed by C_{IN}, and IN/GND pins.

Output Inductor L

There are several considerations in choosing this inductor.

1) Choose the inductance to provide the desired ripple current. It is suggested to choose the ripple current to be about 40% of the maximum output current. The inductance is calculated as:

$$L = \frac{V_{\text{OUT}}(1 - V_{\text{OUT}}/V_{\text{IN,MAX}})}{F_{\text{SW}} \times I_{\text{OUT,MAX}} \times 40\%}$$

Where F_{SW} is the switching frequency and $I_{OUT,MAX}$ is the maximum load current.

 The saturation current rating of the inductor must be selected to be greater than the peak inductor current under full load conditions.

$$I_{SAT, \, MIN} > I_{OUT, \, MAX} + \frac{V_{OUT}(1\text{-}V_{OUT}/V_{IN, MAX})}{2 \cdot F_{SW} \cdot L}$$

3) The DCR of the inductor and the core loss at the switching frequency must be low enough to achieve the desired efficiency requirement. It is desirable to choose an inductor with DCR<25m Ω to achieve a good overall efficiency.

Load Transient Considerations

The SY8893 regulator IC integrates the compensation components to achieve good stability and fast transient responses. In some applications, adding a 22pF ceramic capacitor in parallel with $R_{\rm H}$ may further speed up the load transient responses and is thus recommended for applications with large load transient step requirements.

Layout Design

The layout design of the SY8893 is relatively simple. For the best efficiency and to minimize noise problems, the following components should be placed close to the IC: $C_{\rm IN}$, L, $R_{\rm H}$ and $R_{\rm L}$.

- It is desirable to maximize the PCB copper area connecting to GND pin to achieve the best thermal and noise performance. If the board space allowed, a ground plane is highly desirable. Reasonable paths are suggested to be placed underneath the ground pad to enhance the soldering quality and thermal performance.
- 2) C_{IN} must be close to Pins IN and GND. The loop area formed by C_{IN} and GND must be minimized.
- The PCB copper area associated with LX pin must be minimized to avoid the potential noise problem.
- 4) The components R_H and R_L, and the trace connecting to the FB pin must NOT be adjacent to the LX net on the PCB layout to avoid the noise problem.



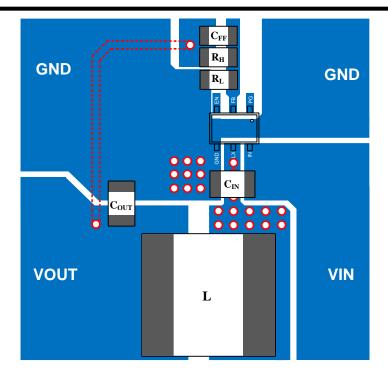
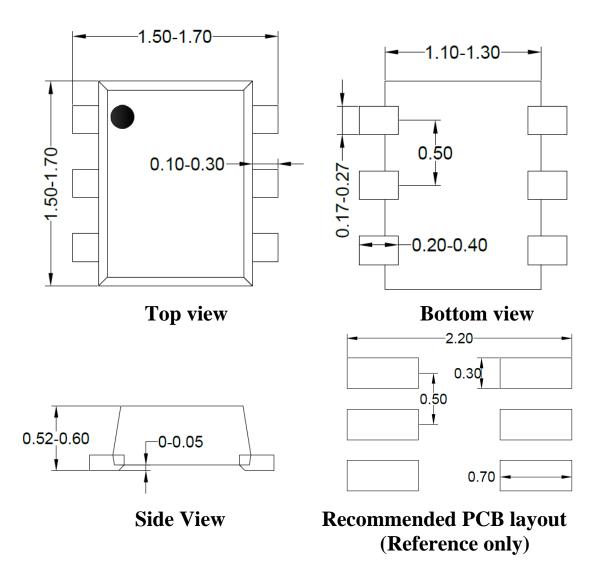


Figure 4. PCB Layout Suggestion



SOT563 Package Outline Drawing



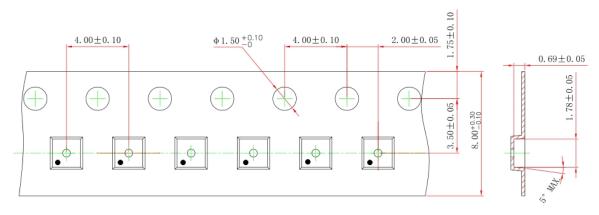
Notes: All dimension in millimeter and exclude mold flash & metal burr.



Taping & Reel Specification

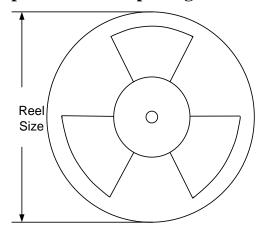
1. Taping Orientation

SOT563



Feeding Direction

2. Carrier Tape & Reel specification for packages



Package types	Tape width (mm)	Pocket pitch(mm)	Reel size (Inch)	Trailer * length(mm)	Leader * length (mm)	Qty per reel (pcs)
SOT563	8	4	7''	280	160	5000

3. Others: NA





Revision History

The revision history provided is for informational purpose only and is believed to be accurate, however, not warranted. Please make sure that you have the latest revision.

Date	Revision	Change	
Dec.17, 2020	Revision 0.9A	Update Taping Specification in page 10	
Jan.04, 2019	Revision 0.9	Initial Release	



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