

# 1.5MHz, 1A Synchronous Step-Down Converter

### **FEATURES**

- High Efficiency: Up to 96%(@3.3V)
- . 1.5MHz Constant Frequency Operation
- . 1.0A Output Current
- . No Schottky Diode Required
- . 2.5V to 5.5V Input Voltage Range
- . Output Voltage as Low as 0.6V
- . 100% Duty Cycle in Dropout
- . Low Quiescent Current: 50µA
- Slope Compensated Current Mode Control for Excellent Line and Load Transient Response
- . Short Circuit Protection
- . Thermal Fault Protection
- . Inrush Current Limit and Soft Start
- Input over voltage protection(OVP)
- . <1µA Shutdown Current
- SOT23-5 Package

#### **GENERAL DESCRIPTION**

The TMI3108 is a constant frequency, current mode PWM step-down converter. The device integrates a main switch and a synchronous rectifier for high efficiency without an external Schottky diode. It is ideal for powering portable equipment that runs from a single cell Lithium-Ion (Li+) battery. The output voltage can be regulated as low as 0.6V. The TMI3108 can also run at 100% duty cycle for low dropout operation, extending battery life in portable system. This device offers two operation modes, PWM control and PFM Mode switching control, which allows a high efficiency over the wider range of the load.

### **APPLICATIONS**

- Cellular and Smart Phones
- . Wireless and DSL Modems
- . PDA/MID/PAD
- . Digital Still and Video Cameras

## TYPICAL APPILCATION

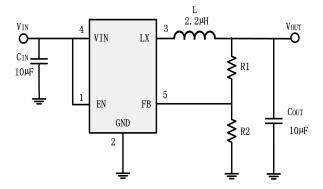


Figure 1. Basic Application Circuit

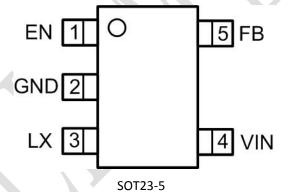
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# **ABSOLUTE MAXIMUM RATINGS(Note 1)**

Parameter	Value	Unit
Input Supply Voltages	-0.3~6.5	V
LX Voltages	-0.3~6.5	V
EN,FB Voltage	-0.3~6.5	V
Storage Temperature Range	-65~150	°C
Junction Temperature (Note 2)	-40~150	°C
Power Dissipation	600	mW
Lead Temperature Soldering,10Sec	260	°C

# **PIN CONFIGURATION**



Top Mark:T50XXX (T50:Device Code, XXX: Inside Code)

Part Number	Package	Top mark	Quantity/ Reel
TMI3108	SOT23-5	T50XXX	3000



## **PIN FUNCTIONS**

Pin	Name	Function
1	EN	Chip Enable Pin. Drive EN above 1.5V to turn on the part. Drive EN below 0.3V to turn it off. Do not leave EN floating.
2	GND	Ground Pin
3	LX	Power Switch Output. It is the switch node connection to Inductor.
4	VIN	Power Supply Input. Must be closely decoupled to GND with a $4.7\mu F$ or greater ceramic capacitor.
5	FB	Output Voltage Feedback Pin.

# **ESD RATING**

Items	Description	Value	Unit
V <sub>ESD</sub>	Human Body Model for all pins	±2000	V

# **JEDEC specification JS-001**

# **RECOMMENDED OPERATING CONDITIONS**

Items	Description	Min Max	Unit
Voltage Range	IN	2.5 5.5	5 V
TA	Operating Temperature Range	-40 85	°C



# **ELECTRICAL CHARACTERISTICS (Note 3)**

( $V_{IN}=V_{EN}=3.6V$ ,  $V_{OUT}=1.8V$ ,  $T_A=25$ °C, unless otherwise noted.)

Parameter	Test Conditions	Min	Тур	Max	Unit
Input Voltage Range		2.5		5.5	V
OVP Threshold			6.0		V
UVLO Threshold			2.1		V
Quiescent Current	V <sub>EN</sub> =2.0V, I <sub>OUT</sub> =0, V <sub>FB</sub> =V <sub>REF</sub> *105%		50	65	μΑ
Shutdown Current	V <sub>EN</sub> =0V		0.1	10	μΑ
	T <sub>A</sub> = 25°C	0.588	0.600	0.612	V
Regulated Feedback Voltage	$T_A = 0$ °C $\leq T_A \leq 85$ °C	0.586	0.600	0.613	V
	T <sub>A</sub> = -40°C ≤T <sub>A</sub> ≤ 85°C	0.585	0.600	0.615	V
Reference Voltage Line Regulation	V <sub>IN</sub> = 2.5V to 6.0V		0.04	0.40	%/V
Output Voltage Line Regulation	V <sub>IN</sub> = 2.5V to 6.0V		0.04	0.4	%
Output Voltage Load Regulation			0.5		%
Oscillation Frequency	V <sub>OUT</sub> =100%		1.5		MHz
Oscillation Frequency	V <sub>OUT</sub> =0V		400		KHz
On Resistance of PMOS	I <sub>LX</sub> =100mA		0.29		Ω
On Resistance of NMOS	I <sub>LX</sub> =-100mA		0.18		Ω
Peak Current Limit	V <sub>IN</sub> =5V,V <sub>OUT</sub> =1.2V,L=4.7uH/2A	1.5			Α
EN Input Low Level				0.30	С
EN Input High Level		1.5			V
EN Leakage Current			±0.01	±1.0	μΑ
LX Leakage Current	V <sub>EN</sub> =0V,V <sub>IN</sub> =V <sub>LX</sub> =5V		±0.01	±1.0	μΑ

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2:  $T_J$  is calculated from the ambient temperature  $T_A$  and power dissipation  $P_D$  according to the following formula:  $T_J = T_A + (P_D) \times (250^{\circ}C/W)$ .

Note 3: 100% production test at +25°C. Specifications over the temperature range are guaranteed by design and characterization.

Note 4: Dynamic supply current is higher due to the gate charge being delivered at the switching frequency



#### **FUNCTION DESCRIPTION**

The TMI3108 is a high performance 1.0A 1.5MHz monolithic step-down converter. The TMI3108 requires only three external power components ( $C_{in}$ ,  $C_{out}$  and L). The adjustable version can be programmed with external feedback to any voltage, ranging from 0.6V to the input voltage.

At dropout, the converter duty cycle increases to 100% and the output voltage tracks the input voltage minus the Rdson drop of the high-side MOSFET.

The internal error amplifier and compensation provides excellent transient response, load, and line regulation. Soft start function prevents input inrush current and output overshoot during start up.

## **FUNCTIONAL BLOCK DIAGRAM**

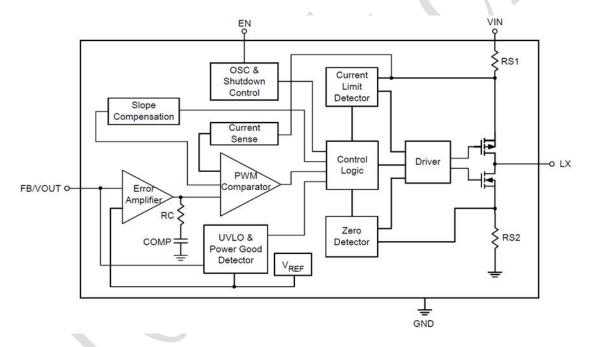


Figure 3. TMI3108 Block Diagram



#### APPLICATION INFORMATION

#### **Setting the Output Voltage**

Figure 1 shows the basic application circuit for the TMI3108. The TMI3108 can be externally programmed. Resistors R1 and R2 in Figure 1 program the output to regulate at a voltage higher than 0.6V. To limit the bias current required for the external feedback resistor string while maintaining good noise immunity, the minimum suggested value for R2 is  $59k\Omega$ . Although a larger value will further reduce quiescent current, it will also increase the impedance of the feedback node, making it more sensitive to external noise and interference.

The external resistor sets the output voltage according to the following equation:

$$V_{OUT} = 0.6 \times (1 + \frac{R_1}{R_2})$$

$$R_1 = (V_{OUT} / 0.6 - 1) \times R_2$$

#### **Inductor Selection**

For most designs, the TMI3108 operates with inductors of  $2.2\mu H$  to  $10\mu H$ . Low inductance values are physically smaller but require faster switching, which results in some efficiency loss. The inductor value can be derived from the following equation:

$$L = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times \Delta I_L \times f_{OSC}}$$

Where  $^{\Delta I_L}$  is inductor Ripple Current. Large value inductors result in lower ripple current and small value inductors result in high ripple current. For optimum voltage-positioning load transients, choose an inductor with DC series resistance in the  $50 \text{m}\Omega$  to  $150 \text{m}\Omega$  range.

#### **Input Capacitor Selection**

The input capacitor reduces the surge current drawn from the input and switching noise from the device.

The input capacitor impedance at the switching frequency should be less than input source impedance to prevent high frequency switching current passing to the input.

A low ESR input capacitor sized for maximum RMS current must be used. Ceramic capacitors with X5R or X7R dielectrics are highly recommended because of their low ESR and small temperature coefficients.



A  $4.7\mu F$  ceramic capacitor for most applications is sufficient. A large value may be used for improved input voltage filtering.

#### **Output Capacitor Selection**

The output capacitor is required to keep the output voltage ripple small and to ensure regulation loop stability. The output capacitor must have low impedance at the switching frequency. Ceramic capacitors with X5R or X7R dielectrics are recommended due to their low ESR and high ripple current ratings. The output ripple V<sub>OUT</sub> is determined by:

$$\Delta V_{OUT} \leq \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times f_{OSC} \times L} \times \left(ESR + \frac{1}{8 \times f_{osc} \times C3}\right)$$

A 10µF ceramic can satisfy most applications.

#### **Layout Consideration**

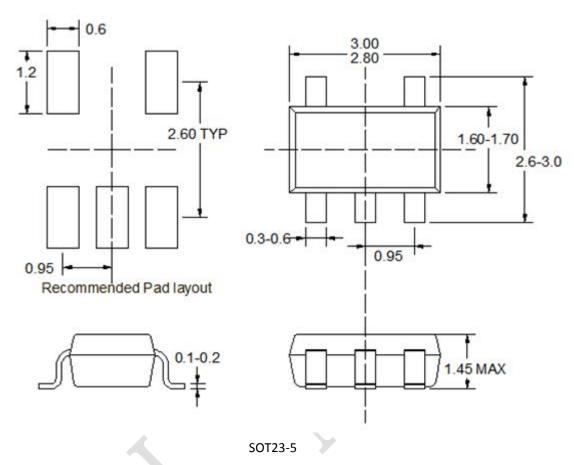
When laying out the printed circuit board, the Following checking should be used to ensure proper operation of the TMI3108. Check the following in your layout:

- 1. The power traces, consisting of the GND trace, the LX trace and the VIN trace should be kept short, direct and wide.
- 2.Does the (+) plates of Cin connect to Vin as closely as possible. This capacitor provides the AC current to the internal power MOSFETs.
- 3. Keep the switching node, LX, away from the sensitive VOUT node.
- 4. Keep the (-) plates of Cin and Cout as close as possible





### **PACKAGE INFORMATION**



#### Note:

- 1) All dimensions are in millimeters.
- 2) Package length does not include mold flash, protrusion or gate burr.
- 3) Package width does not include inter lead flash or protrusion.
- 4) Lead popularity (bottom of leads after forming) shall be 0.10 millimeters max.
- 5) Pin 1 is lower left pin when reading top mark from left to right.