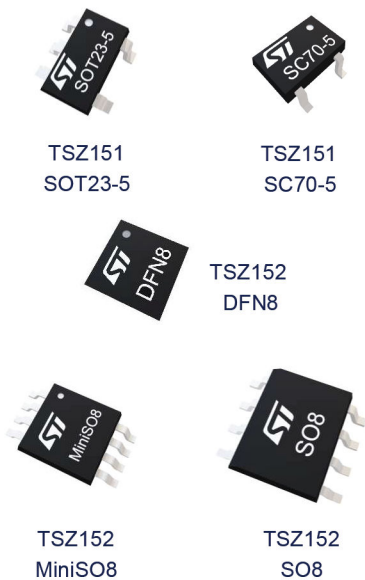


Very high accuracy (7  $\mu\text{V}$ ) high bandwidth (1.6 MHz) zero-drift 5 V op amp


## Features

- Very high accuracy and stability: offset voltage
  - 7  $\mu\text{V}$  max. at 25 °C
  - 10  $\mu\text{V}$  over full temperature range (-40 °C to 125 °C)
- Rail-to-rail input and output
- Low supply voltage: from 1.8 to 5.5 V
- Low power consumption: 210  $\mu\text{A}$  at 5 V
- Gain bandwidth product: 1.6 MHz
- Extended temperature range: -40 to 125 °C
- AEC-Q100 qualified
- Benefits:
  - Higher accuracy without calibration
  - Accuracy is virtually unaffected by temperature change

## Applications

- High accuracy signal conditioning
- Automotive current measurement and sensors signal conditioning

## Description

The **TSZ151** and **TSZ152** are single and dual operational amplifier featuring very low offset voltages with virtually zero-drift versus temperature changes.

The **TSZ151** and **TSZ152** are single and dual offers rail-to-rail input and output, an excellent speed/power consumption ratio, and 1.6 MHz gain bandwidth product, while consuming just 210  $\mu\text{A}$  at 5 V.

The device also features an ultra-low input bias current. These features make the **TSZ151** and **TSZ152** are ideal for high-accuracy sensor interfaces.

Product status link	Channel	Automotive	Package
TSZ151ICT	1		SC70-5
TSZ151IYCT	1	•	SC70-5
TSZ151ILT	1		SOT23-5
TSZ151IYLT	1	•	SOT23-5
TSZ152IDT	2		SO8
TSZ152IYDT	2	•	SO8
TSZ152IST	2		MiniSO8
TSZ152IYST	2	•	MiniSO8
TSZ152IQT	2		DFN8
TSZ152IYQT	2	•	DFN8

Related products	
TSZ121 TSZ122	Zero drift amplifier for more power savings (400 kHz, 31 $\mu\text{A}$ )
TSZ181 TSZ182	Zero drift amplifier for higher bandwidth (3 MHz, 800 $\mu\text{A}$ )

## 1 Pin description

### 1.1 TSZ151 single operational amplifier (SC70-5)

Figure 1. Pin connection (top view)

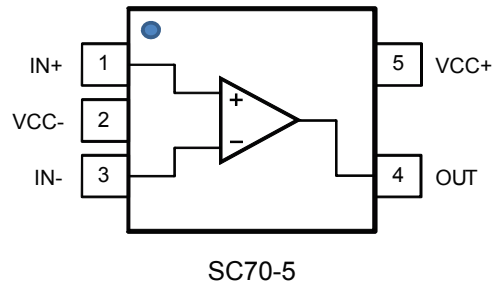


Table 1. Pin description

Pin n°	Pin name	Description
1	IN+	Non-inverting input channel
2	VCC-	Negative supply voltage
3	IN-	Inverting input channel
4	OUT	Output channel
5	VCC+	Positive supply voltage

### 1.2 TSZ151 single operational amplifier (SOT23-5)

Figure 2. Pin connection (top view)

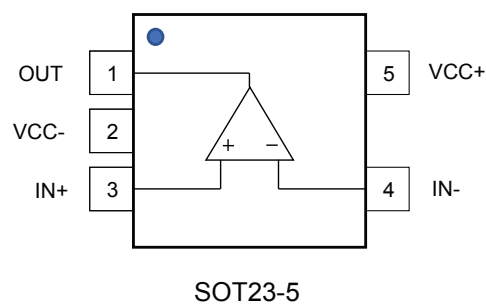
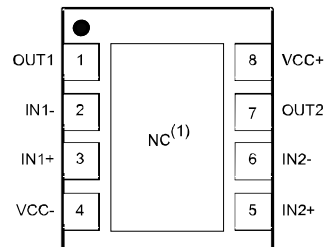


Table 2. Pin description

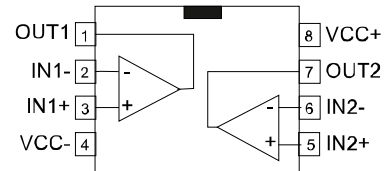
Pin n°	Pin name	Description
1	OUT	Output channel
2	VCC-	Negative supply voltage
3	IN+	Non-inverting input channel
4	IN-	Inverting input channel
5	VCC+	Positive supply voltage

### 1.3 TSZ152 dual operational amplifier

**Figure 3. Pin connections (top view)**



DFN8 2x2 (TSZ152)



MiniSO8 and SO8 (TSZ152)

1. The exposed pad of the DFN8 2x2 can be connected to  $V_{CC-}$  or left floating.

**Table 3. Pin description**

Pin n°	Pin name	Description
1	OUT1	Output channel
2	IN1-	Inverting input channel
3	IN1+	Non-inverting input channel
4	VCC-	Negative supply voltage
5	IN2+	Non-inverting input channel
6	IN2-	Inverting input channel
7	OUT2	Output channel
8	VCC+	Positive supply voltage

## 2 Maximum ratings

**Table 4. Absolute maximum ratings**

Symbol	Parameter <sup>(1)</sup>	Value	Unit	
V <sub>CC</sub>	Supply voltage	6	V	
V <sub>id</sub>	Differential input voltage (V <sub>IN+</sub> - V <sub>IN-</sub> )	±V <sub>CC</sub>		
V <sub>in</sub>	Input voltage <sup>(2)</sup>	(V <sub>CC-</sub> ) -0.2 to (V <sub>CC+</sub> ) +0.2		
I <sub>in</sub>	Input current	±10	mA	
T <sub>stg</sub>	Storage temperature	-65 to +150	°C	
T <sub>j</sub>	Maximum junction temperature	150		
R <sub>th-ja</sub>	Thermal resistance junction-to-ambient <sup>(3)</sup>	SC70-5	205	°C/W
		SOT23-5	250	
		SO8	113	
		MiniSO8	127	
		DFN8 2x2	76	
ESD	HBM: human body model (industrial grade) <sup>(4)</sup>	4000	V	
	HBM: human body model (automotive grade) <sup>(5)</sup>	4000		
	CDM: charged device model	1000		

1. All voltage values are with respect to the VCC- pin, unless otherwise specified.
2. The maximum input voltage value may be extended on the condition that the input current is limited to ±10 mA.
3. R<sub>th-ja</sub> is a typical value, obtained with PCB according to JEDEC 2s2p without vias.
4. Human body model: HBM test according to the standard ESDA-JS-001-2017.
5. Human body model: HBM test according to the standard AEC-Q100-002.

**Table 5. Operating conditions**

Symbol	Parameter	Value	Unit
V <sub>CC</sub>	Supply voltage	1.8 to 5.5	V
V <sub>icm</sub>	Common-mode input voltage range	(V <sub>CC-</sub> ) -0.1 V to (V <sub>CC+</sub> ) +0.1	
T <sub>oper</sub>	Operating free-air temperature range	-40 to 125	°C

### 3 Electrical characteristics

**Table 6. Electrical characteristics at  $V_{CC} = 5\text{ V}$ ,  $V_{icm} = V_{OUT} = V_{CC} / 2$ ,  $T = 25\text{ °C}$  (unless otherwise specified),  $R_L$  and  $C_L$  (if any specified) connected to  $V_{CC} / 2$ .**

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
<b>DC performance</b>						
$V_{io}$	Input offset voltage	$T = 25\text{ °C}$		1	$\pm 7$	$\mu\text{V}$
		$-40\text{ °C} \leq T \leq 125\text{ °C}$			$\pm 10$	
$\Delta V_{io}/\Delta T$	Input offset voltage temperature drift <sup>(1)</sup>	$-40\text{ °C} \leq T \leq 125\text{ °C}$	-15		35	$\text{nV}/\text{°C}$
$I_{ib}$	Input bias current	$T = 25\text{ °C}$		50	200	$\text{pA}$
		$-40\text{ °C} \leq T \leq 125\text{ °C}$			300	
$I_{io}$	Input offset current	$T = 25\text{ °C}$		100	400	$\text{pA}$
		$-40\text{ °C} \leq T \leq 125\text{ °C}$			600	
$A_{VD}$	Open loop gain	$V_{CC-} + 200\text{ mV} \leq V_{OUT} \leq V_{CC+} - 200\text{ mV}$ , $R_L = 10\text{ k}\Omega$ , $T = 25\text{ °C}$	120	135		$\text{dB}$
		$V_{CC-} + 200\text{ mV} \leq V_{OUT} \leq V_{CC+} - 200\text{ mV}$ , $R_L = 10\text{ k}\Omega$ , $-40\text{ °C} \leq T \leq 125\text{ °C}$	115			
CMR	Common-mode rejection ratio $20 \cdot \log(\Delta V_{io}/\Delta V_{icm})$	$V_{CC-} \leq V_{icm} \leq V_{CC+}$ , $T = 25\text{ °C}$	120	132		$\text{dB}$
		$V_{CC-} \leq V_{icm} \leq V_{CC+}$ , $-40\text{ °C} \leq T \leq 125\text{ °C}$	115			
SVR	Supply-voltage rejection ratio $20 \cdot \log(\Delta V_{io}/\Delta V_{CC})$	$1.8\text{ V} \leq V_{CC} \leq 5.5\text{ V}$ , $V_{icm} = 0\text{ V}$ , $T = 25\text{ °C}$	120	140		$\text{dB}$
		$1.8\text{ V} \leq V_{CC} \leq 5.5\text{ V}$ , $V_{icm} = 0\text{ V}$ , $-40\text{ °C} \leq T \leq 125\text{ °C}$	120			
$V_{OH}$	High level output voltage drop ( $V_{OH} = V_{CC+} - V_{OUT}$ )	$R_L = 10\text{ k}\Omega$ , $T = 25\text{ °C}$			30	$\text{mV}$
		$R_L = 10\text{ k}\Omega$ , $-40\text{ °C} \leq T \leq 125\text{ °C}$			70	
$V_{OL}$	Low level output voltage drop ( $V_{OL} = V_{OUT}$ )	$R_L = 10\text{ k}\Omega$ , $T = 25\text{ °C}$			30	$\text{mV}$
		$R_L = 10\text{ k}\Omega$ , $-40\text{ °C} \leq T \leq 125\text{ °C}$			70	
$I_{OUT}$	$I_{SINK}$ , OUT connected to $V_{CC+}$	$V_{ID} = -0.1\text{ V}$ , $T = 25\text{ °C}$	20	30		$\text{mA}$
		$V_{ID} = -0.1\text{ V}$ , $-40\text{ °C} \leq T \leq 125\text{ °C}$	15			
	$I_{SOURCE}$ , OUT connected to $V_{CC-}$	$V_{ID} = +0.1\text{ V}$ , $T = 25\text{ °C}$	20	30		
		$V_{ID} = +0.1\text{ V}$ , $-40\text{ °C} \leq T \leq 125\text{ °C}$	15			
$I_{CC}$	Supply current (by operational amplifier)	$T = 25\text{ °C}$		210	300	$\mu\text{A}$
		$-40\text{ °C} \leq T \leq 125\text{ °C}$			300	
<b>AC performance</b>						
GBP	Gain bandwidth product	$C_L = 100\text{ pF}$		1.6		$\text{MHz}$
SR	Slew rate <sup>(2)</sup> measured from 10% to 90%	$A_V = 1\text{ V/V}$ , $V_{in} = 0.3\text{ V}$ to $V_{CC+} - 0.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$		0.8		$\text{V}/\mu\text{s}$
$\Phi_m$	Phase margin	$C_L = 100\text{ pF}$		60		degrees
en	Input voltage noise density	$f = 1\text{ kHz}$		27		$\text{nV}/\sqrt{\text{Hz}}$
en p-p	Input noise voltage	$0.1\text{ Hz} \leq f \leq 10\text{ Hz}$		0.5		$\mu\text{V}_{pp}$

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
$t_{rec}$	Overload recovery time	$V_{in}$ from ( $V_{CC+} + 100$ mV) to ( $V_{CC+} - 1$ V), $V_{OUT}$ measured at ( $V_{CC+} - 100$ mV), $A_V = +1$		5		$\mu$ s
$t_{init}$	Initialization time, $V_{OUT}$ at 100 mV from final value	$T = 25$ °C		80		$\mu$ s
EMIRR	EMI rejection rate = $-20 \log (V_{RFpeak}/\Delta V_{io})$	$V_{RF} = 100$ mVpp, $f = 400$ MHz		84		dB
		$V_{RF} = 100$ mVpp, $f = 900$ MHz		87		
		$V_{RF} = 100$ mVpp, $f = 1800$ MHz		90		
		$V_{RF} = 100$ mVpp, $f = 2400$ MHz		91		

**Table 7. Electrical characteristics at  $V_{CC} = 3.3\text{ V}$ ,  $V_{icm} = V_{OUT} = V_{CC} / 2$ ,  $T = 25\text{ }^{\circ}\text{C}$  (unless otherwise specified),  $R_L$  and  $C_L$  (if any specified) connected to  $V_{CC} / 2$ .**

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
<b>DC performance</b>						
$V_{io}$	Input offset voltage	$T = 25\text{ }^{\circ}\text{C}$		1	$\pm 7$	$\mu\text{V}$
		$-40\text{ }^{\circ}\text{C} \leq T \leq 125\text{ }^{\circ}\text{C}$			$\pm 10$	
$\Delta V_{io}/\Delta T$	Input offset voltage temperature drift <sup>(1)</sup>	$-40\text{ }^{\circ}\text{C} \leq T \leq 125\text{ }^{\circ}\text{C}$	-18		38	$\text{nV}/^{\circ}\text{C}$
$I_{ib}$	Input bias current	$T = 25\text{ }^{\circ}\text{C}$		50	200	$\mu\text{A}$
		$-40\text{ }^{\circ}\text{C} \leq T \leq 125\text{ }^{\circ}\text{C}$			300	
$I_{io}$	Input offset current	$T = 25\text{ }^{\circ}\text{C}$		100	400	$\mu\text{A}$
		$-40\text{ }^{\circ}\text{C} \leq T \leq 125\text{ }^{\circ}\text{C}$			600	
$A_{VD}$	Open loop gain	$V_{CC-} + 200\text{ mV} \leq V_{OUT} \leq V_{CC+} - 200\text{ mV}$ , $R_L = 10\text{ k}\Omega$ , $T = 25\text{ }^{\circ}\text{C}$	116	131		dB
		$V_{CC-} + 200\text{ mV} \leq V_{OUT} \leq V_{CC+} - 200\text{ mV}$ , $R_L = 10\text{ k}\Omega$ , $-40\text{ }^{\circ}\text{C} \leq T \leq 125\text{ }^{\circ}\text{C}$	111			
CMR	Common-mode rejection ratio $20 \cdot \log(\Delta V_{io}/\Delta V_{icm})$	$V_{CC-} \leq V_{icm} \leq V_{CC+}$ , $T = 25\text{ }^{\circ}\text{C}$	116	128		dB
		$V_{CC-} \leq V_{icm} \leq V_{CC+}$ , $-40\text{ }^{\circ}\text{C} \leq T \leq 125\text{ }^{\circ}\text{C}$	111			
$V_{OH}$	High level output voltage drop ( $V_{OH} = V_{CC+} - V_{OUT}$ )	$R_L = 10\text{ k}\Omega$ , $T = 25\text{ }^{\circ}\text{C}$			25	$\text{mV}$
		$R_L = 10\text{ k}\Omega$ , $-40\text{ }^{\circ}\text{C} \leq T \leq 125\text{ }^{\circ}\text{C}$			60	
$V_{OL}$	Low level output voltage drop ( $V_{OL} = V_{OUT}$ )	$T = 25\text{ }^{\circ}\text{C}$			25	$\text{mV}$
		$-40\text{ }^{\circ}\text{C} \leq T \leq 125\text{ }^{\circ}\text{C}$			60	
$I_{OUT}$	$I_{SINK}$ , OUT connected to $V_{CC+}$	$V_{ID} = -0.1\text{ V}$ , $T = 25\text{ }^{\circ}\text{C}$	14	21		$\text{mA}$
		$V_{ID} = -0.1\text{ V}$ , $-40\text{ }^{\circ}\text{C} \leq T \leq 125\text{ }^{\circ}\text{C}$	10			
	$I_{SOURCE}$ , OUT connected to $V_{CC-}$	$V_{ID} = +0.1\text{ V}$ , $T = 25\text{ }^{\circ}\text{C}$	14	21		
		$V_{ID} = +0.1\text{ V}$ , $-40\text{ }^{\circ}\text{C} \leq T \leq 125\text{ }^{\circ}\text{C}$	10			
$I_{CC}$	Supply current (by operational amplifier)	$T = 25\text{ }^{\circ}\text{C}$		210	300	$\mu\text{A}$
		$-40\text{ }^{\circ}\text{C} \leq T \leq 125\text{ }^{\circ}\text{C}$			300	
<b>AC performance</b>						
GBP	Gain bandwidth product	$C_L = 100\text{ pF}$		1.6		MHz
SR	Slew rate <sup>(2)</sup> measured from 10% to 90%	$A_V = 1\text{ V/V}$ , $V_{in} = 0.3\text{ V}$ to $V_{CC+} - 0.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$		0.8		$\text{V}/\mu\text{s}$
$\Phi_m$	Phase margin	$C_L = 100\text{ pF}$		60		degrees
$e_n$	Input voltage noise density	$f = 1\text{ kHz}$		28		$\text{nV}/\sqrt{\text{Hz}}$
$e_n\text{ p-p}$	Input noise voltage	$0.1\text{ Hz} \leq f \leq 10\text{ Hz}$		0.5		$\mu\text{V}_{pp}$
$t_{rec}$	Overload recovery time	$V_{in}$ from $(V_{CC+} + 100\text{ mV})$ to $(V_{CC+} - 1\text{ V})$ , $V_{OUT}$ measured at $(V_{CC+} - 100\text{ mV})$ , $A_V = +1$		3		$\mu\text{s}$
$t_{init}$	Initialization time, $V_{OUT}$ at 100 mV from final value	$T = 25\text{ }^{\circ}\text{C}$		70		$\mu\text{s}$
EMIRR	EMI rejection rate = $-20 \log(V_{RFpeak}/\Delta V_{io})$	$V_{RF} = 100\text{ mV}_{pp}$ , $f = 400\text{ MHz}$		84		dB
		$V_{RF} = 100\text{ mV}_{pp}$ , $f = 900\text{ MHz}$		87		
		$V_{RF} = 100\text{ mV}_{pp}$ , $f = 1800\text{ MHz}$		90		

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
EMIRR	EMI rejection rate = $-20 \log (V_{RFpeak}/\Delta V_{io})$	$V_{RF} = 100 \text{ mVpp}, f = 2400 \text{ MHz}$		91		dB



**Table 8. Electrical characteristics at  $V_{CC} = 1.8\text{ V}$ ,  $V_{icm} = V_{OUT} = V_{CC} / 2$ ,  $T = 25\text{ }^\circ\text{C}$  (unless otherwise specified),  $R_L$  and  $C_L$  (if any specified) connected to  $V_{CC} / 2$ .**

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
<b>DC performance</b>						
$V_{io}$	Input offset voltage	$T = 25\text{ }^\circ\text{C}$		1	$\pm 7$	$\mu\text{V}$
		$-40\text{ }^\circ\text{C} \leq T \leq 125\text{ }^\circ\text{C}$			$\pm 10$	
$\Delta V_{io}/\Delta T$	Input offset voltage temperature drift <sup>(1)</sup>	$-40\text{ }^\circ\text{C} \leq T \leq 125\text{ }^\circ\text{C}$	-38		43	$\text{nV}/^\circ\text{C}$
$I_{ib}$	Input bias current	$T = 25\text{ }^\circ\text{C}$		50	200	$\mu\text{A}$
		$-40\text{ }^\circ\text{C} \leq T \leq 125\text{ }^\circ\text{C}$			300	
$I_{io}$	Input offset current	$T = 25\text{ }^\circ\text{C}$		100	400	$\mu\text{A}$
		$-40\text{ }^\circ\text{C} \leq T \leq 125\text{ }^\circ\text{C}$			600	
$A_{VD}$	Open loop gain	$V_{CC-} + 200\text{ mV} \leq V_{OUT} \leq V_{CC+} - 200\text{ mV}$ , $R_L = 10\text{ k}\Omega$ , $T = 25\text{ }^\circ\text{C}$	110	125		dB
		$V_{CC-} + 200\text{ mV} \leq V_{OUT} \leq V_{CC+} - 200\text{ mV}$ , $R_L = 10\text{ k}\Omega$ , $-40\text{ }^\circ\text{C} \leq T \leq 125\text{ }^\circ\text{C}$	105			
CMR	Common-mode rejection ratio $20 \cdot \log(\Delta V_{io}/\Delta V_{icm})$	$V_{CC-} \leq V_{icm} \leq V_{CC+}$ , $T = 25\text{ }^\circ\text{C}$	110	122		dB
		$V_{CC-} \leq V_{icm} \leq V_{CC+}$ , $-40\text{ }^\circ\text{C} \leq T \leq 125\text{ }^\circ\text{C}$	105			
$V_{OH}$	High level output voltage drop ( $V_{OH} = V_{CC+} - V_{OUT}$ )	$R_L = 10\text{ k}\Omega$ , $T = 25\text{ }^\circ\text{C}$			20	$\text{mV}$
		$R_L = 10\text{ k}\Omega$ , $-40\text{ }^\circ\text{C} \leq T \leq 125\text{ }^\circ\text{C}$			50	
$V_{OL}$	Low level output voltage drop ( $V_{OL} = V_{OUT}$ )	$R_L = 10\text{ k}\Omega$ , $T = 25\text{ }^\circ\text{C}$			20	$\text{mV}$
		$R_L = 10\text{ k}\Omega$ , $-40\text{ }^\circ\text{C} \leq T \leq 125\text{ }^\circ\text{C}$			50	
$I_{OUT}$	$I_{SINK}$ , OUT connected to $V_{CC+}$	$V_{ID} = -0.1\text{ V}$ , $T = 25\text{ }^\circ\text{C}$	7	11		$\text{mA}$
		$V_{ID} = -0.1\text{ V}$ , $-40\text{ }^\circ\text{C} \leq T \leq 125\text{ }^\circ\text{C}$	5			
	$I_{SOURCE}$ , OUT connected to $V_{CC-}$	$V_{ID} = +0.1\text{ V}$ , $T = 25\text{ }^\circ\text{C}$	7	11		
		$V_{ID} = +0.1\text{ V}$ , $-40\text{ }^\circ\text{C} \leq T \leq 125\text{ }^\circ\text{C}$	5			
$I_{CC}$	Supply current (by operational amplifier)	$T = 25\text{ }^\circ\text{C}$		210	300	$\mu\text{A}$
		$-40\text{ }^\circ\text{C} \leq T \leq 125\text{ }^\circ\text{C}$			300	
<b>AC performance</b>						
GBP	Gain bandwidth product	$C_L = 100\text{ pF}$		1.6		MHz
SR	Slew rate <sup>(2)</sup> measured from 10% to 90%	$A_V = 1\text{ V/V}$ , $V_{in} = 0.3\text{ V}$ to $V_{CC+} - 0.3\text{ V}$ , $R_L = 10\text{ k}\Omega$ , $C_L = 100\text{ pF}$		0.8		$\text{V}/\mu\text{s}$
$\Phi_m$	Phase margin	$C_L = 100\text{ pF}$		60		degrees
$e_n$	Input voltage noise density	$f = 1\text{ kHz}$		33		$\text{nV}/\sqrt{\text{Hz}}$
$e_n\text{ p-p}$	Input noise voltage	$0.1\text{ Hz} \leq f \leq 10\text{ Hz}$		0.57		$\mu\text{V}_{pp}$
$t_{rec}$	Overload recovery time	$V_{in}$ from $(V_{CC+} + 100\text{ mV})$ to $(V_{CC+} - 1\text{ V})$ , $V_{OUT}$ measured at $(V_{CC+} - 100\text{ mV})$ , $A_V = +1$		2		$\mu\text{s}$
$t_{init}$	Initialization time, $V_{OUT}$ at 100 mV from final value	$T = 25\text{ }^\circ\text{C}$		60		$\mu\text{s}$
EMIRR	EMI rejection rate = $-20 \log(V_{RFpeak}/\Delta V_{io})$	$V_{RF} = 100\text{ mV}_{pp}$ , $f = 400\text{ MHz}$		84		dB
		$V_{RF} = 100\text{ mV}_{pp}$ , $f = 900\text{ MHz}$		87		
		$V_{RF} = 100\text{ mV}_{pp}$ , $f = 1800\text{ MHz}$		90		

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
EMIRR	EMI rejection rate = $-20 \log (V_{RFpeak}/\Delta V_{io})$	$V_{RF} = 100 \text{ mVpp}, f = 2400 \text{ MHz}$		91		dB

1. See section 4.2 Input offset voltage drift vs. temperature.
2. The slew rate value is the average of rising and falling values.

## 4 Typical performance characteristics

Figure 4. Input offset voltage distribution at  $V_{CC} = 1.8\text{ V}$

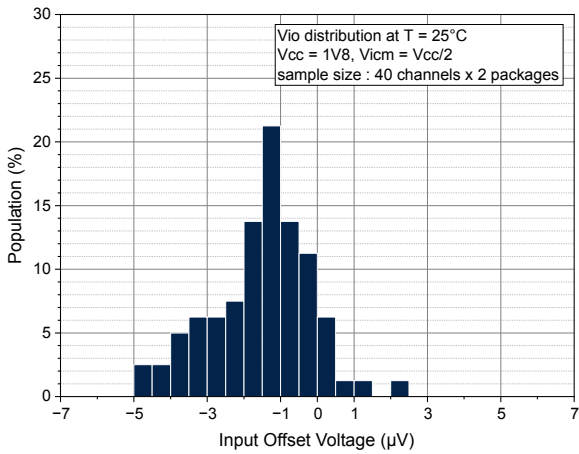


Figure 5. Input offset voltage distribution at  $V_{CC} = 3.3\text{ V}$

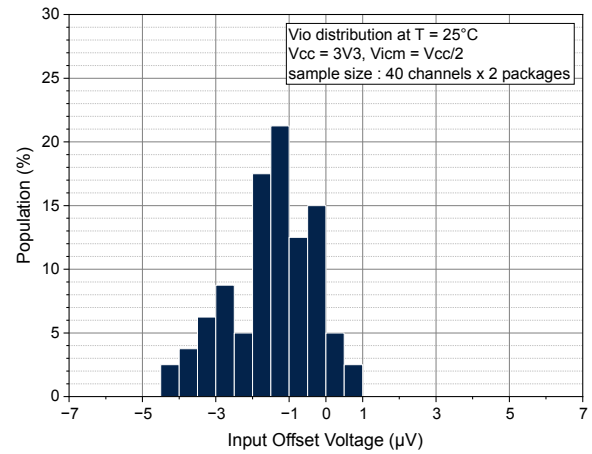


Figure 6. Input offset voltage distribution at  $V_{CC} = 5\text{ V}$

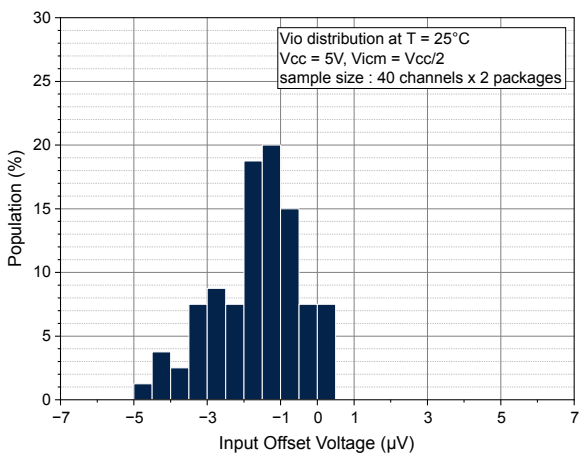
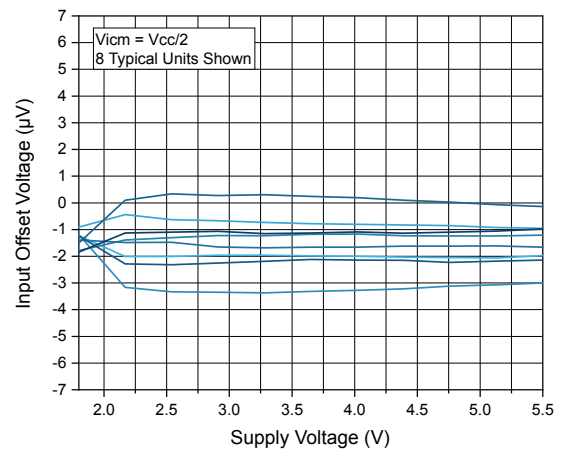
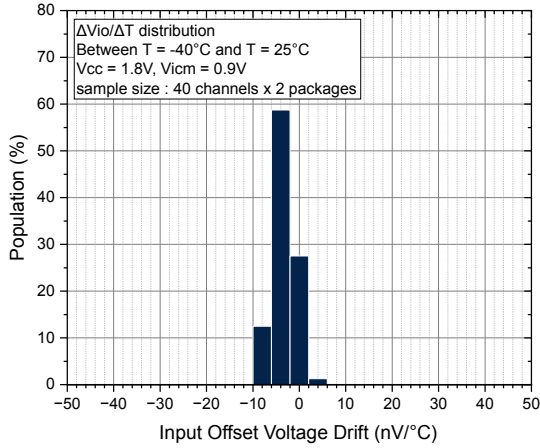


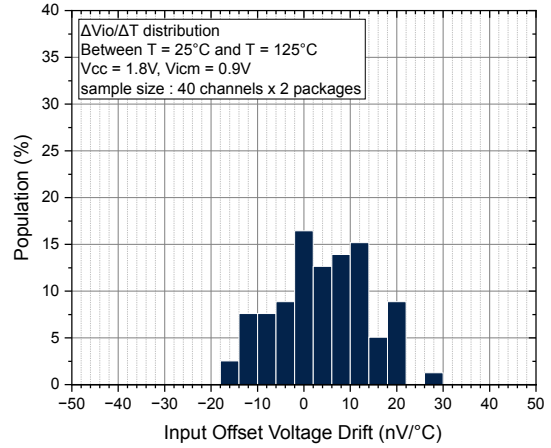
Figure 7. Input offset voltage vs. supply voltage



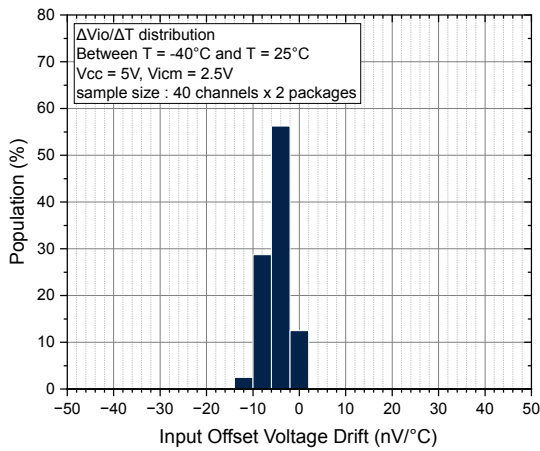
**Figure 8. Input offset voltage temperature drift distribution at  $V_{CC} = 1.8\text{ V}$  ( $-40\text{ }^{\circ}\text{C}$  to  $25\text{ }^{\circ}\text{C}$ )**



**Figure 9. Input offset voltage temperature drift distribution at  $V_{CC} = 1.8\text{ V}$  ( $25\text{ }^{\circ}\text{C}$  to  $125\text{ }^{\circ}\text{C}$ )**



**Figure 10. Input offset voltage temperature drift distribution at  $V_{CC} = 5\text{ V}$  ( $-40\text{ }^{\circ}\text{C}$  to  $25\text{ }^{\circ}\text{C}$ )**



**Figure 11. Input offset voltage temperature drift distribution at  $V_{CC} = 5\text{ V}$  ( $25\text{ }^{\circ}\text{C}$  to  $125\text{ }^{\circ}\text{C}$ )**

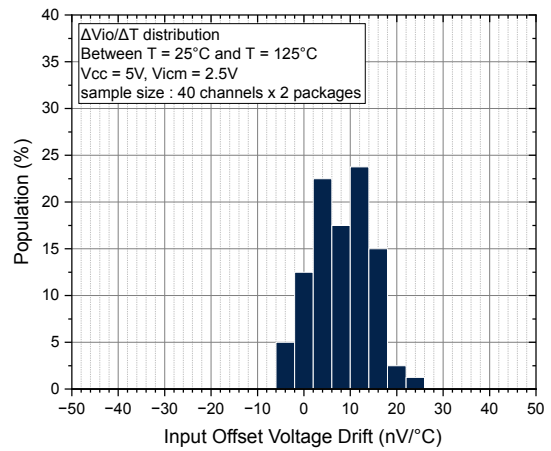


Figure 12. Input offset voltage vs. temperature

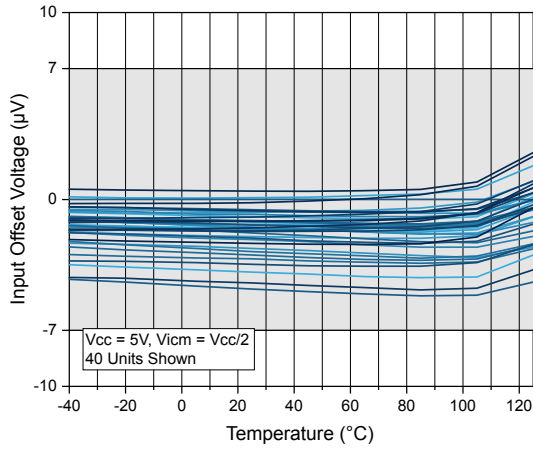


Figure 13. Input offset voltage vs. input common-mode voltage at  $V_{CC} = 1.8V$

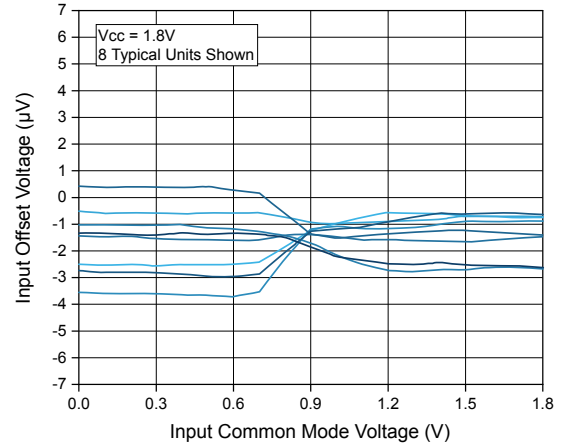


Figure 14. Input offset voltage vs. input common-mode voltage at  $V_{CC} = 3.3V$

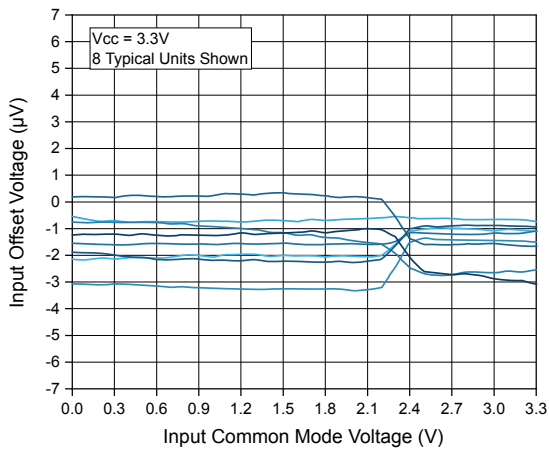


Figure 15. Input offset voltage vs. input common-mode voltage at  $V_{CC} = 5V$

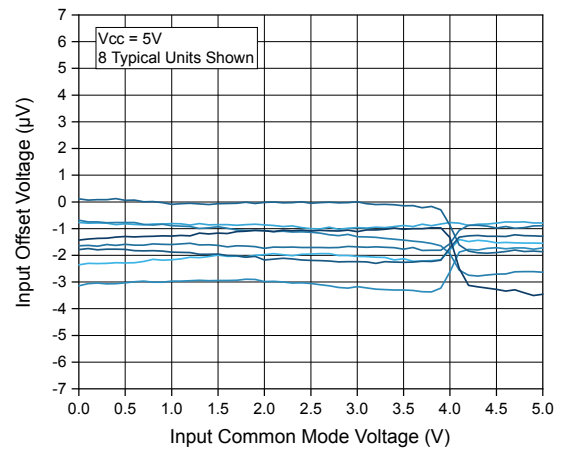


Figure 16. Supply current vs. input common-mode voltage at  $V_{CC} = 5\text{ V}$

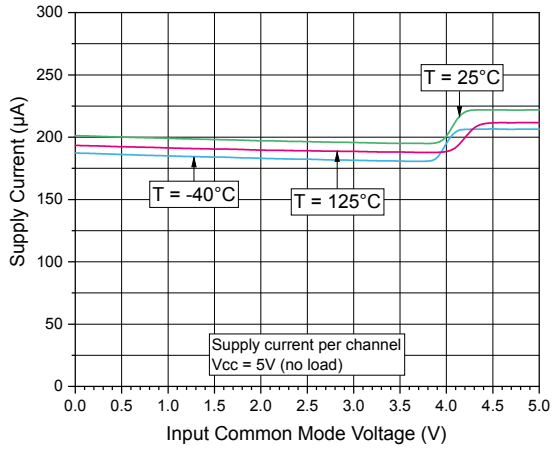


Figure 17. Supply current vs. supply voltage

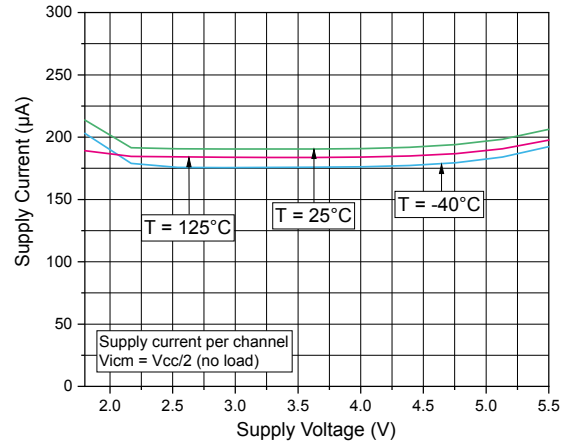


Figure 18. Output drop voltage  $V_{OH}$  vs. supply voltage

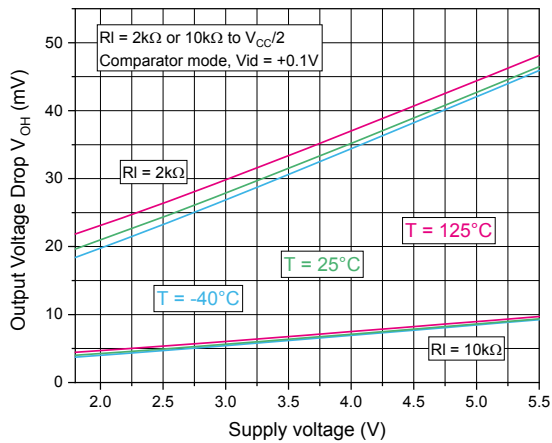


Figure 19. Output drop voltage  $V_{OL}$  vs. supply voltage

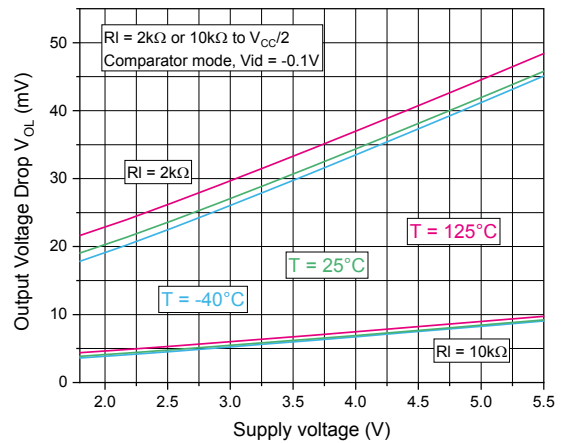


Figure 20. Output current vs. output voltage at  $V_{CC} = 1.8\text{ V}$

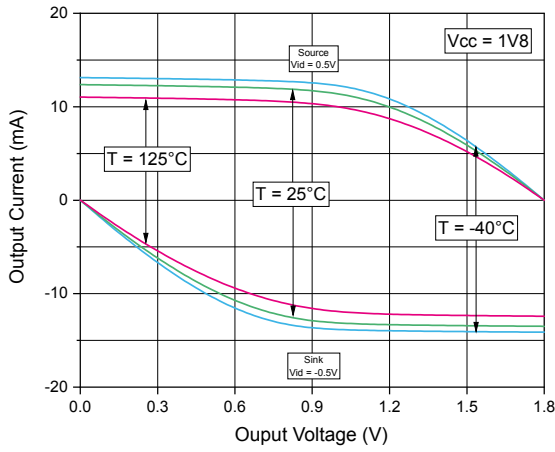


Figure 21. Output current vs. output voltage at  $V_{CC} = 3.3\text{ V}$

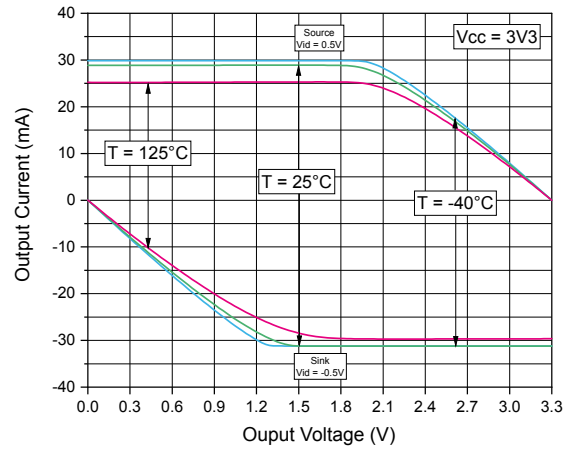


Figure 22. Output current vs. output voltage at  $V_{CC} = 5\text{ V}$

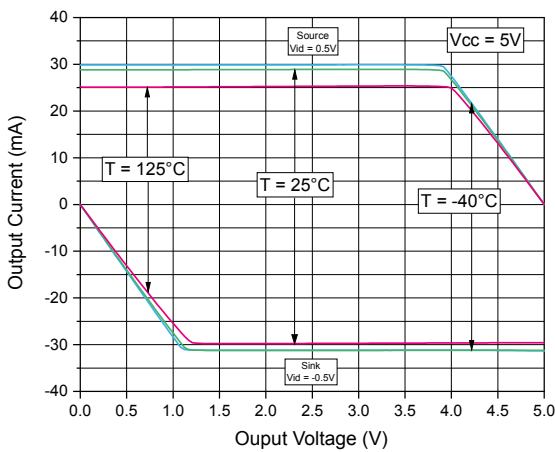


Figure 23. Input bias current vs. input common-mode voltage at  $V_{CC} = 1.8\text{ V}$

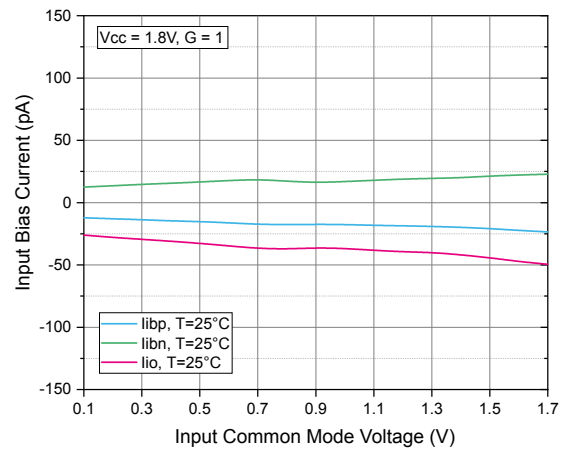


Figure 24. Input bias current vs. input common-mode voltage at  $V_{CC} = 5\text{ V}$

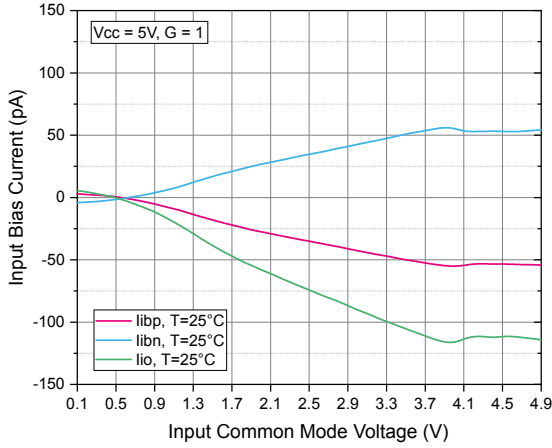


Figure 25. Input bias current vs. input common-mode voltage at  $T = 125^\circ\text{C}$

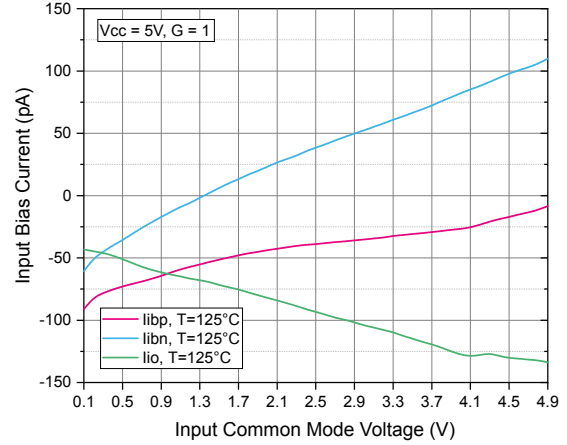


Figure 26. Bode diagram vs. temperature at  $V_{CC} = 1.8\text{ V}$

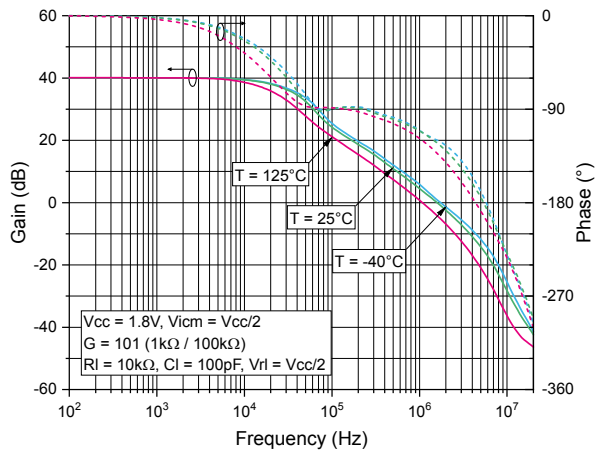


Figure 27. Bode diagram vs. temperature at  $V_{CC} = 3.3\text{ V}$

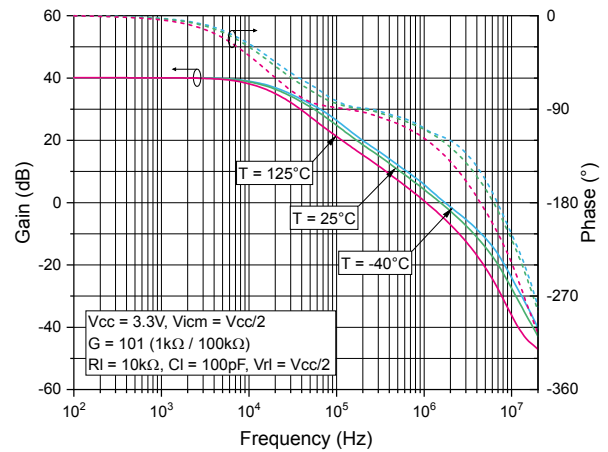




Figure 28. Bode diagram vs. temperature at  $V_{CC} = 5\text{ V}$

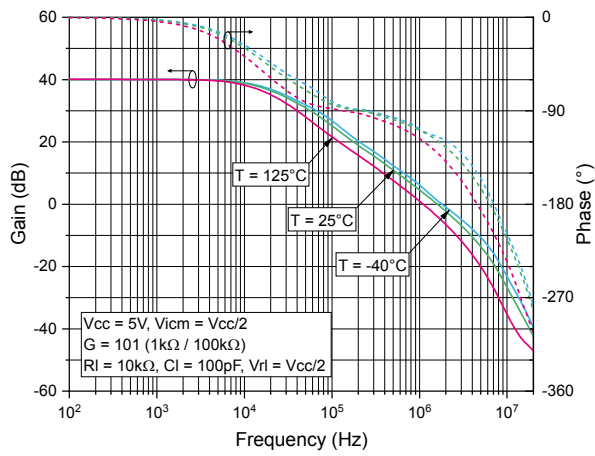


Figure 29. Bode diagram vs. output current at  $V_{CC} = 1.8\text{ V}$

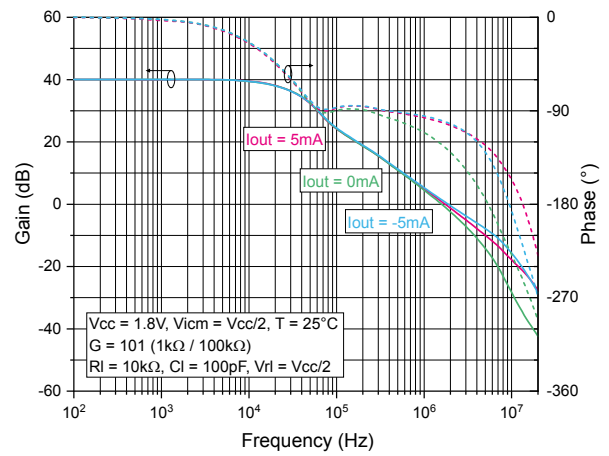


Figure 30. Bode diagram vs. output current at  $V_{CC} = 3.3\text{ V}$

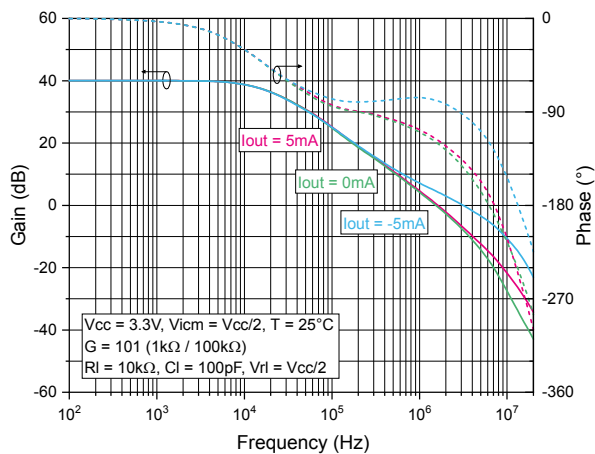


Figure 31. Bode diagram vs. output current at  $V_{CC} = 5\text{ V}$

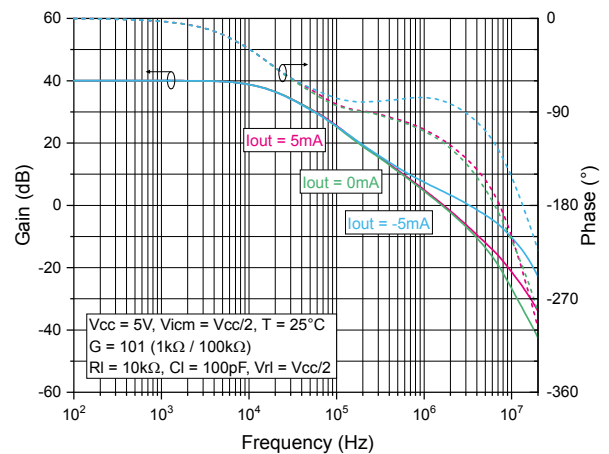


Figure 32. Step overshoot vs. load capacitance

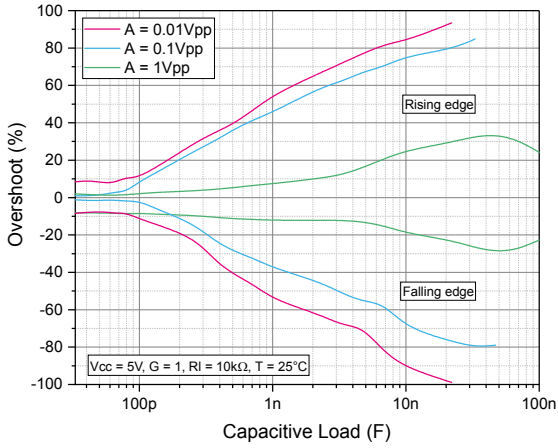


Figure 33. Small step overshoot vs. load capacitance and Riso

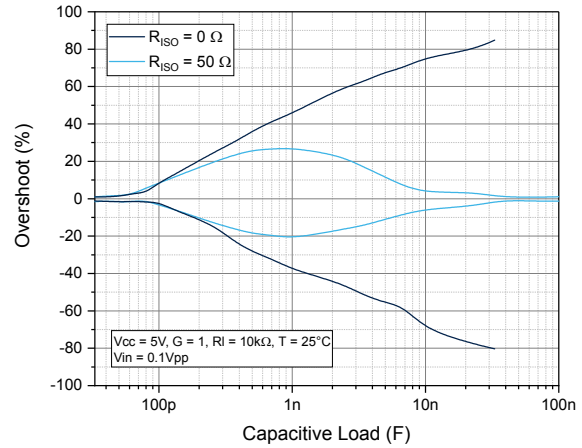


Figure 34. Small step overshoot vs. Riso

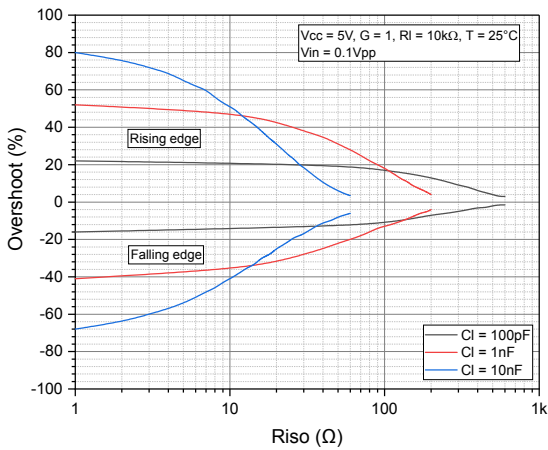
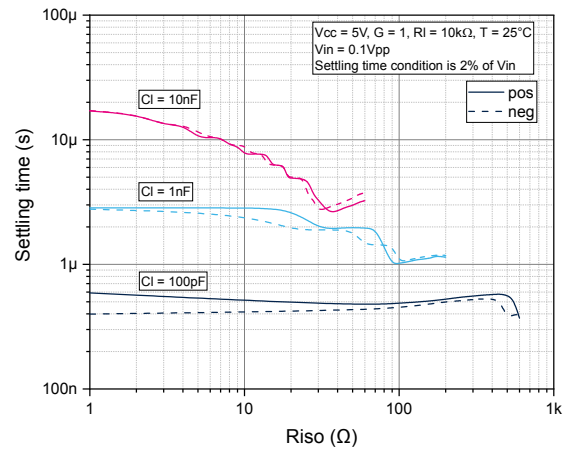
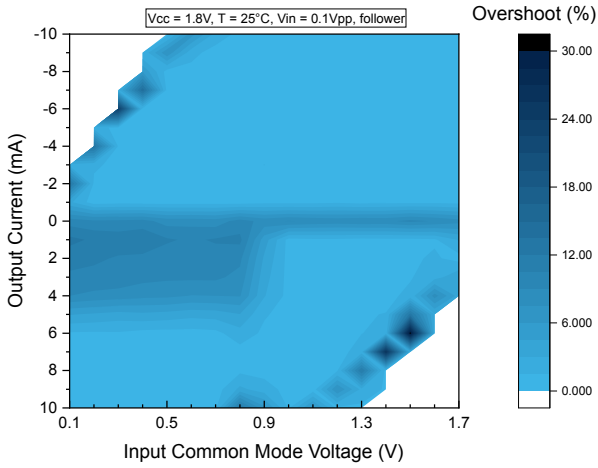


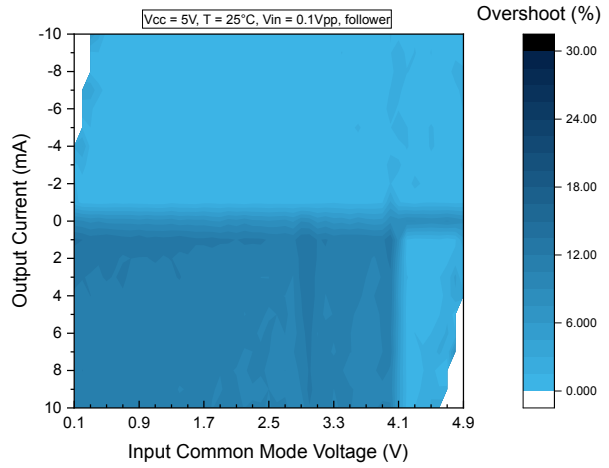
Figure 35. Small step settling time vs. Riso



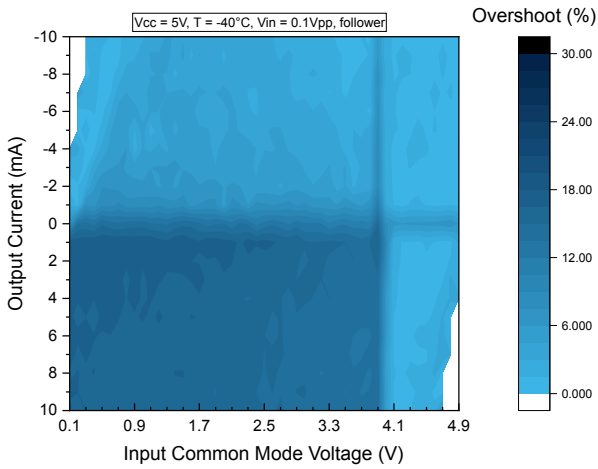
**Figure 36. Overshoot vs. input common-mode voltage and output current at  $V_{CC} = 1.8\text{ V}$**



**Figure 37. Overshoot vs. input common-mode voltage and output current at  $V_{CC} = 5\text{ V}$**



**Figure 38. Overshoot vs. input common-mode voltage and output current at  $T = -40^\circ\text{C}$**



**Figure 39. Overshoot vs. input common-mode voltage and output current at  $T = 125^\circ\text{C}$**

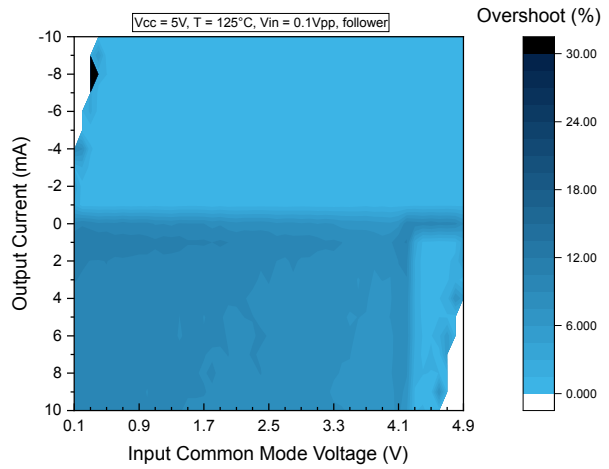


Figure 40. Slew rate vs. temperature

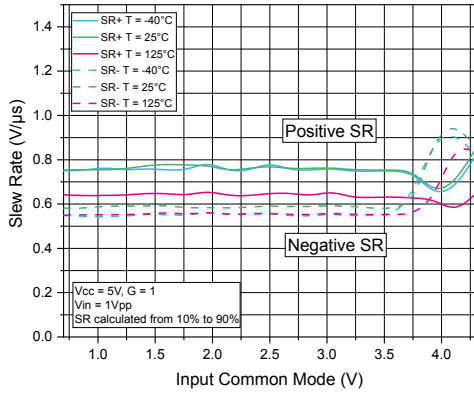


Figure 41. Noise vs. time at  $V_{CC} = 5V$

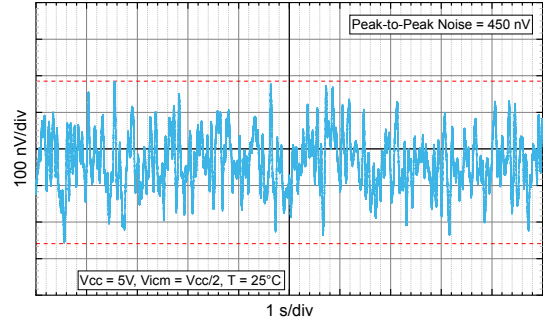


Figure 42. Voltage noise density vs. frequency at  $V_{CC} = 1.8V$

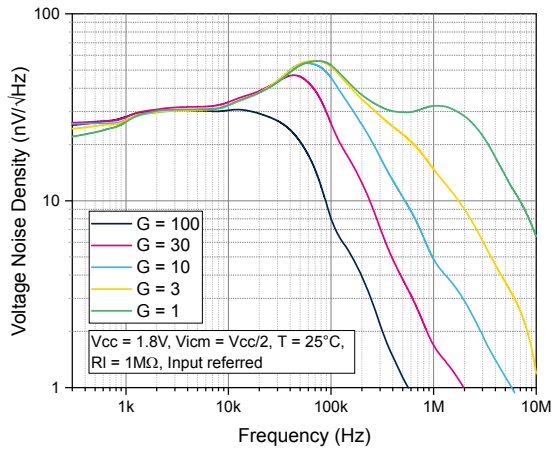


Figure 43. Voltage noise density vs. frequency at  $V_{CC} = 5V$

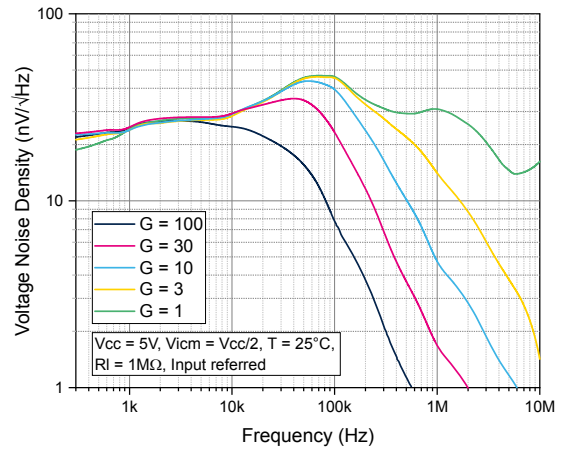


Figure 44. Small signal step response at  $V_{CC} = 1.8\text{ V}$

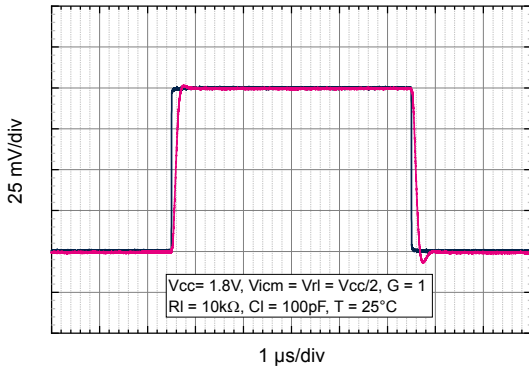


Figure 45. Large signal step response at  $V_{CC} = 1.8\text{ V}$

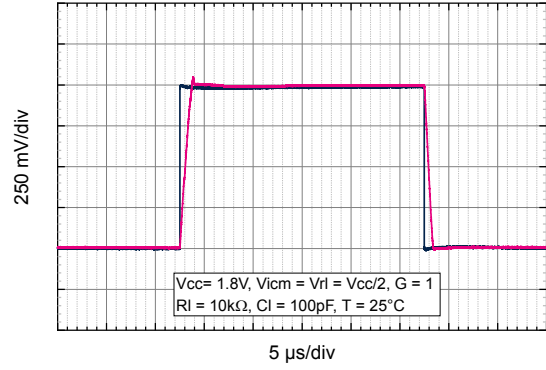


Figure 46. Small signal step response at  $V_{CC} = 5\text{ V}$

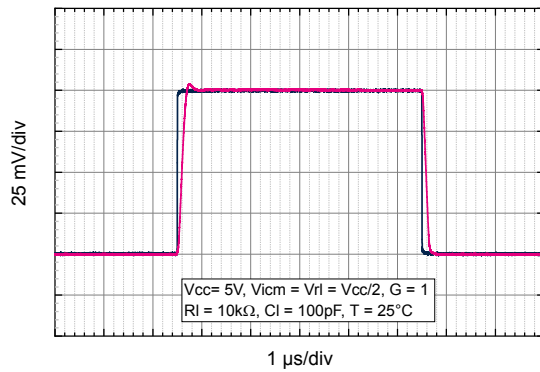


Figure 47. Large signal step response at  $V_{CC} = 5\text{ V}$

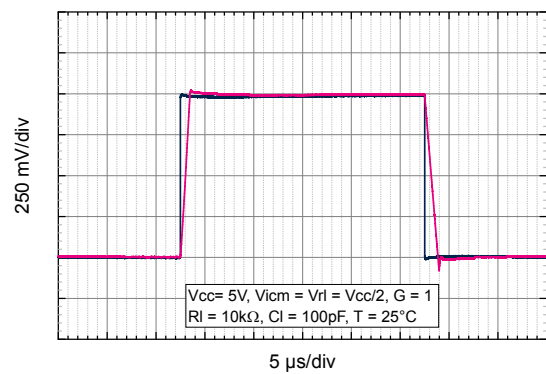


Figure 48. Very large signal step response at  $V_{CC} = 5\text{ V}$

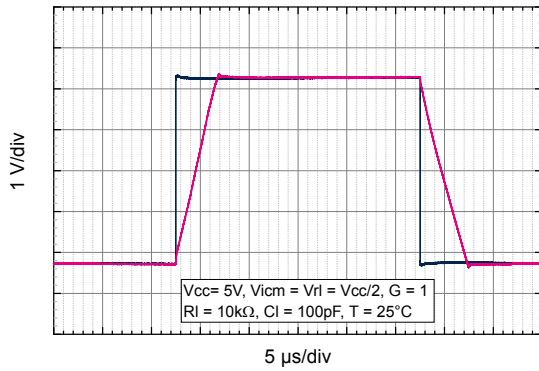


Figure 49. Output voltage saturation response at  $V_{CC} = 5\text{ V}$

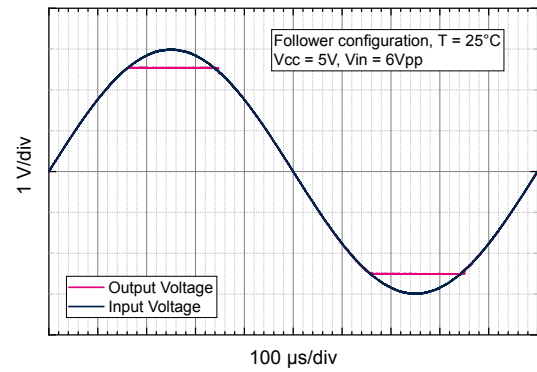


Figure 50. Positive overvoltage recovery at  $V_{CC} = 1.8\text{ V}$

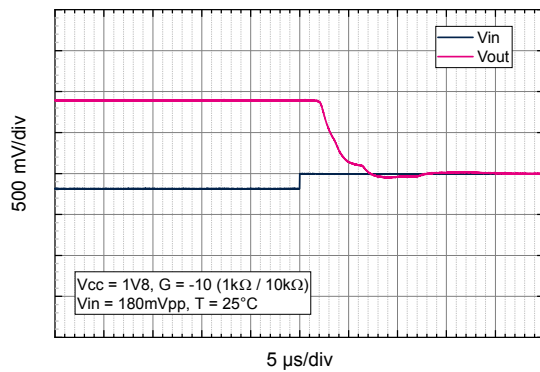


Figure 51. Positive overvoltage recovery at  $V_{CC} = 5\text{ V}$

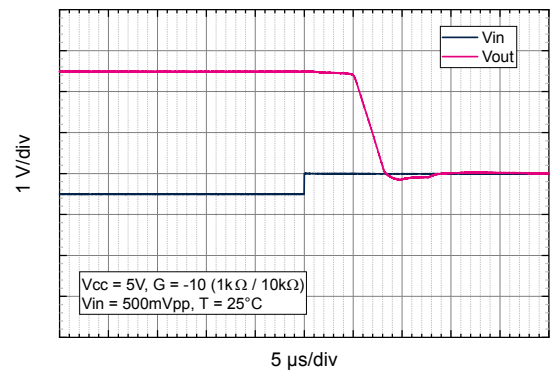


Figure 52. Negative overvoltage recovery at  $V_{CC} = 1.8\text{ V}$

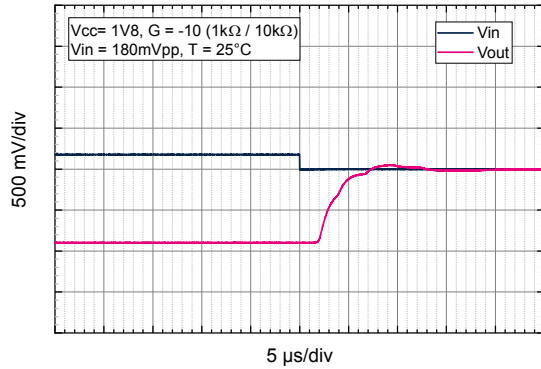


Figure 53. Negative overvoltage recovery at  $V_{CC} = 5\text{ V}$

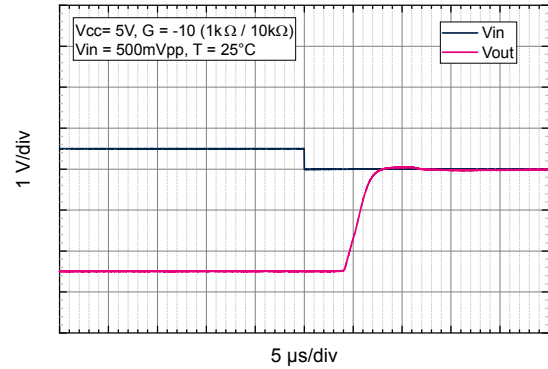


Figure 54. Settling time on negative input step

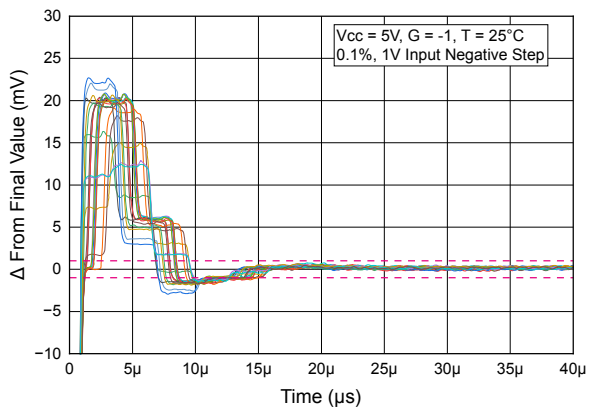


Figure 55. Settling time on positive input step

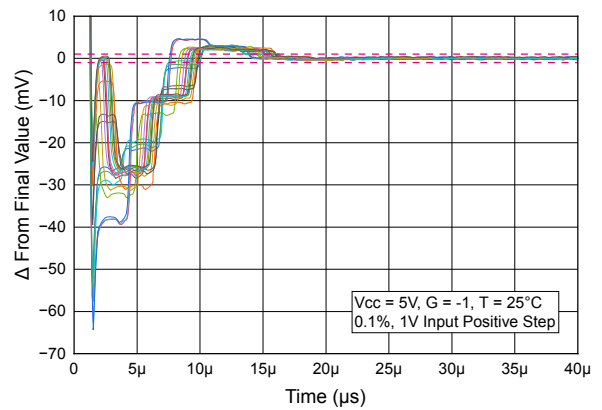


Figure 56. Output voltage vs. input voltage at  $V_{CC} = 5\text{ V}$

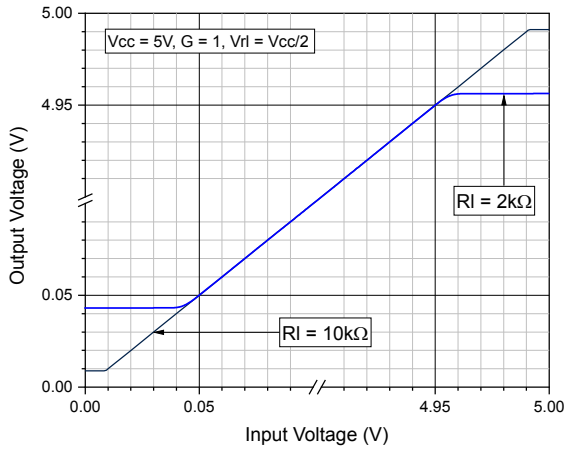


Figure 57. Startup behavior at  $V_{CC} = 5\text{ V}$

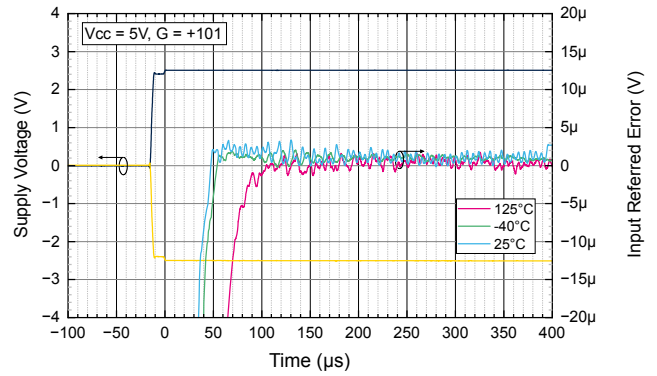


Figure 58. CMRR and PSRR vs. frequency at  $V_{CC} = 1.8\text{ V}$

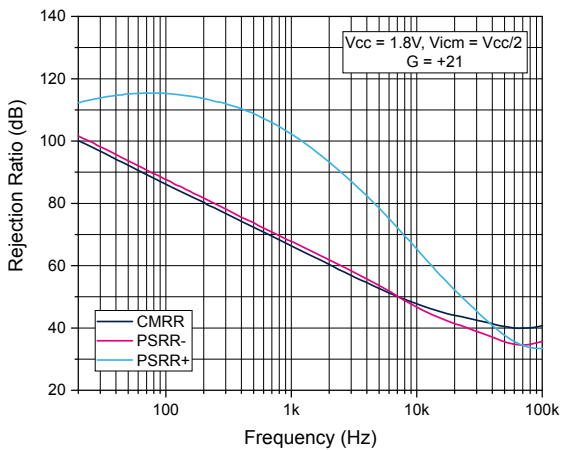


Figure 59. CMRR and PSRR vs. frequency at  $V_{CC} = 5\text{ V}$

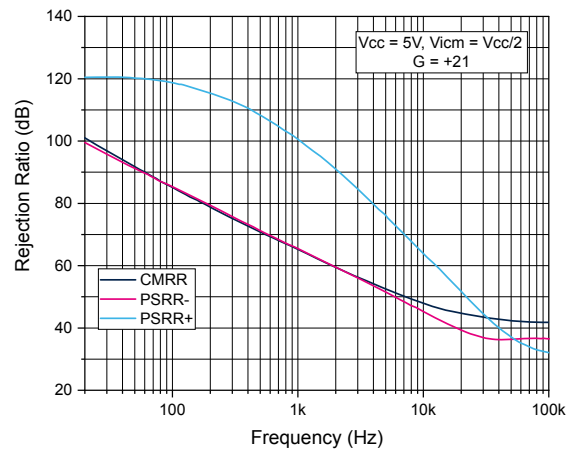




Figure 60. THD vs. frequency

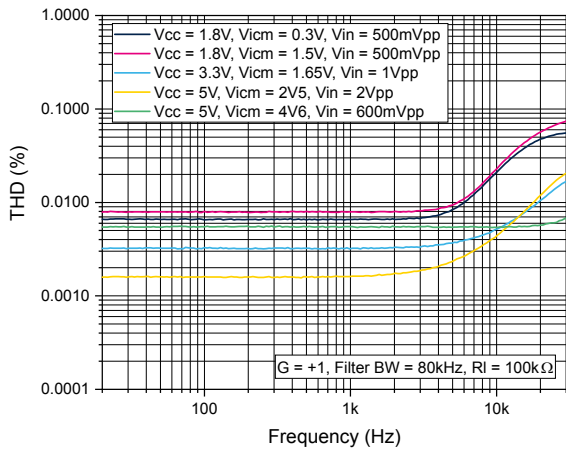
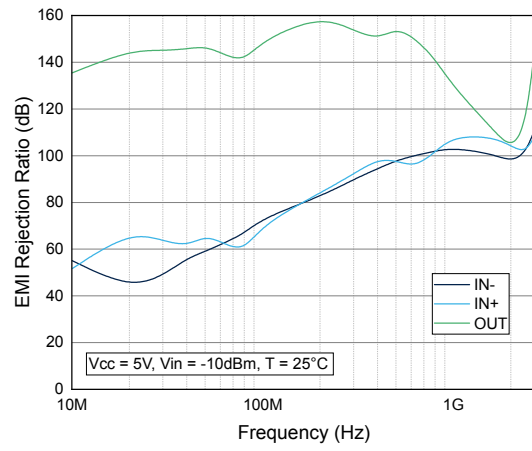


Figure 61. EMI rejection vs. frequency



## 5 Application information

### 5.1 Operating voltages

The TSZ151 and TSZ152 devices can operate from 1.8 to 5.5 V. The parameters are fully specified at 1.8 V, 3.3 V, and 5 V power supplies. However, the parameters are very stable over the full  $V_{CC}$  range and several characterization curves show the TSZ151 and TSZ152 devices characteristics over the full operating range. Additionally, the main specifications are guaranteed in an extended temperature range from -40 to 125 °C.

### 5.2 Input offset voltage drift vs. temperature

The maximum input voltage drift variation vs. temperature is defined as the offset variation related to the offset value measured at 25 °C. The operational amplifier is one of the main circuits of the signal conditioning chain, and the amplifier input offset ( $V_{io}$ ) is a major contributor to the chain accuracy. The signal chain accuracy at 25 °C can be compensated during production at application level. The maximum input voltage drift vs. temperature enables the system designer to anticipate the effect of temperature variations. The maximum input voltage drift vs. temperature is computed using equation 1.

$$\frac{\Delta V_{io}}{\Delta T} = \left( \frac{V_{io,T} - V_{io,25^{\circ}C}}{T - 25^{\circ}C} \right)_{T = -40^{\circ}C \text{ and } T = 125^{\circ}C} \quad (1)$$

The datasheet minimum and maximum values are guaranteed by a measurement on a representative sample size ensuring a  $C_{pk}$  (process capability index) greater than 1.3.

### 5.3 Maximum power dissipation

The usable output load current drive is limited by the maximum power dissipation allowed by the device package. The absolute maximum junction temperature for the TSZ151 is 150 °C. The junction temperature can be estimated as follows:

$$T_J = P_D \times \theta_{JA} + T_A \quad (2)$$

$T_J$  is the die junction temperature.

$P_D$  is the power dissipated in the package.

$\theta_{JA}$  is the junction to thermal resistance of the package.

$T_A$  is the ambient temperature.

The power dissipated in the package  $P_D$  is the sum of the quiescent power dissipated and the power dissipated by the output stage transistor. It is calculated as follows:

$P_D = (V_{CC} \times I_{CC}) + (V_{CC+} - V_{OUT}) \times I_{OUT}$  when the op amp is sourcing the current.

$P_D = (V_{CC} \times I_{CC}) + (V_{OUT} - V_{CC-}) \times I_{OUT}$  when the op amp is sinking the current.

Do not exceed the 150 °C maximum junction temperature for the device. Exceeding the junction temperature limit can cause degradation in the parametric performance or even destroy the device.

### 5.4 PCB layout recommendations

Particular attention must be paid to the layout of the PCB tracks connected to the amplifier, load, and power supply. The power and ground traces are critical as they must provide adequate energy and grounding for all circuits. The best practice is to use short and wide PCB traces to minimize voltage drops and parasitic inductance. In addition, to minimize parasitic impedance over the entire surface, use a multi-via technique that connects the bottom and top layer ground planes together in many locations. The copper traces that connect the output pins to the load and supply pins should be as wide as possible to minimize trace resistance.

### 5.5 Decoupling capacitor

In order to ensure op amp full functionality, it is mandatory to place a decoupling capacitor of at least 22 nF as close as possible to the op amp supply pin. A good decoupling helps to reduce an electromagnetic interference impact.

## 5.6 Macromodel

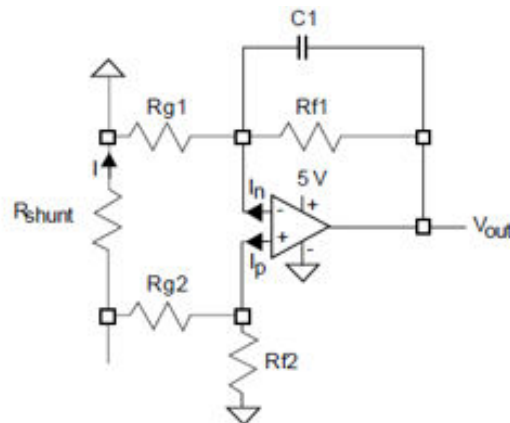
Accurate macromodels of the TSZ151 device are available on the STMicroelectronics website at: [www.st.com](http://www.st.com). These models are a trade-off between the accuracy and complexity (that is, time simulation) of the TSZ151 operational amplifier. They emulate the nominal performance of a typical device within the specified operating conditions mentioned in the datasheet. They also help to validate a design approach and to select the right operational amplifier, but they do not replace on-board measurements.

## 6 Typical applications

### 6.1 Low-side current sensing

Power management mechanisms are found in most electronic systems. Current sensing is useful for protecting applications. The low-side current sensing method consists of placing a sense resistor between the load and the circuit ground. The resulting voltage drop is amplified using the TSZ151 and TSZ152.

**Figure 62. Low-side current sensing schematic**



$V_{out}$  can be expressed as follows:

$$V_{out} = R_{shunt} \cdot I \left( 1 - \frac{R_{g2}}{R_{g2} + r_{f2}} \right) \cdot \left( 1 + \frac{R_{f1}}{R_{f2}} \right) + I_p \cdot \frac{R_{g2} \cdot R_{f2}}{R_{g2} + R_{f2}} \cdot \left( 1 + \frac{R_{f1}}{R_{g1}} \right) - I_n \cdot R_{f1} - V_{io} \cdot \left( 1 + \frac{R_{f1}}{r_{g1}} \right) \quad (3)$$

Assuming that  $R_{f2} = R_{f1} = R_f$  and  $R_{g2} = R_{g1} = R_g$ , equation 4 can be simplified as follows:

$$V_{out} = R_{shunt} \cdot I \cdot \frac{R_f}{R_g} - V_{io} \cdot \left( 1 + \frac{R_f}{R_g} \right) + R_f \cdot I_{io} \quad (4)$$

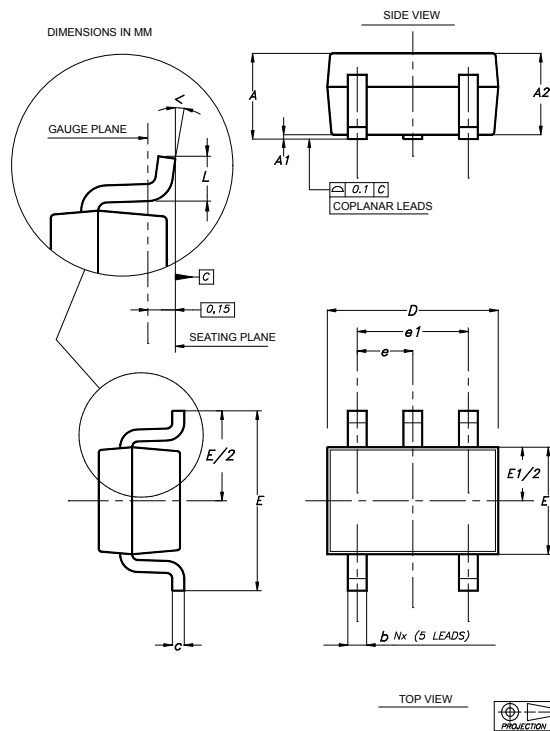
The main advantage of using the TSZ151 and TSZ152 for a low-side current sensing relies on its low  $V_{io}$ , compared to general purpose operational amplifiers. For the same current and targeted accuracy, the shunt resistor can be chosen with a lower value, resulting in lower power dissipation, lower drop in the ground path, and lower cost. Particular attention must be paid to the precision of  $R_{g1}$  and  $R_{f1}$ , to maximize the accuracy of the measurement.

Note that the open loop gain of TSZ151 and TSZ152 are defined close to the rail. It enables measurements over a wide range of currents.

## 7 Package information

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In order to meet environmental requirements, ST offers these devices in different grades of **ECOPACK** packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: [www.st.com](http://www.st.com). ECOPACK is an ST trademark.

**7.1 SC70-5 (or SOT323-5) package information**
**Figure 63. SC70-5 (or SOT323-5) package outline**

**Table 9. SC70-5 (or SOT323-5) package mechanical data**

Ref.	Dimensions					
	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A	0.80		1.10	0.032		0.043
A1			0.10			0.004
A2	0.80	0.90	1.00	0.032	0.035	0.039
b	0.15		0.30	0.006		0.012
c	0.10		0.22	0.004		0.009
D	1.80	2.00	2.20	0.071	0.079	0.087
E	1.80	2.10	2.40	0.071	0.083	0.094
E1	1.15	1.25	1.35	0.045	0.049	0.053
e		0.65			0.025	
e1		1.30			0.051	
L	0.26	0.36	0.46	0.010	0.014	0.018
<	0°		8°	0°		8°

## 7.2 SOT23-5 package information

Figure 64. SOT23-5 package outline

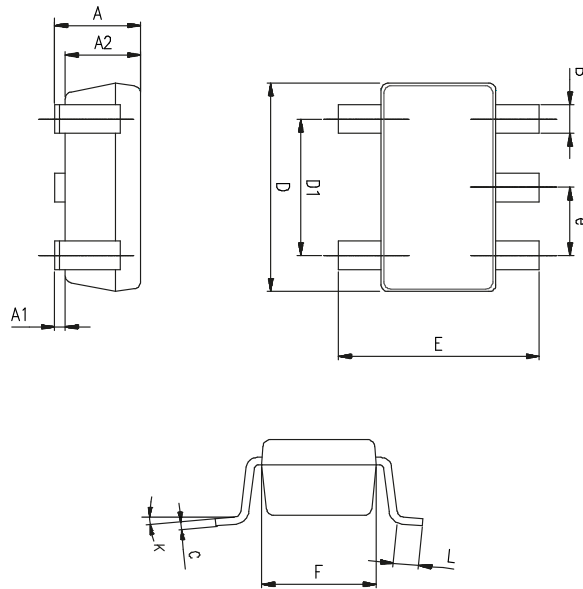
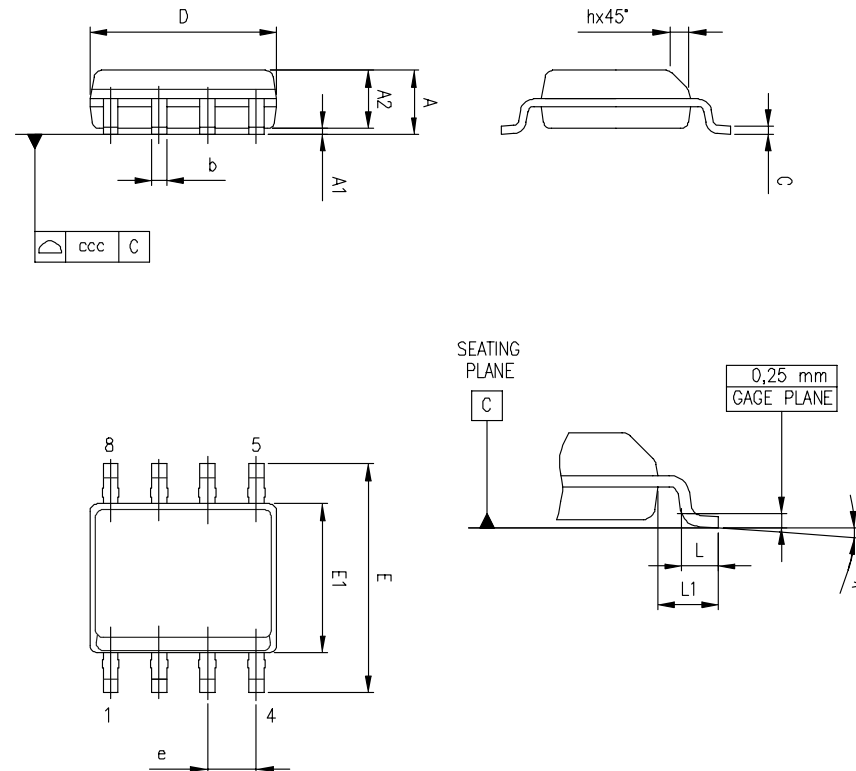


Table 10. SOT23-5 mechanical data

Symbol	Dimensions					
	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A	0.90	1.20	1.45	0.035	0.047	0.057
A1			0.15			0.006
A2	0.90	1.05	1.30	0.035	0.041	0.051
B	0.35	0.40	0.50	0.014	0.016	0.020
C	0.09	0.15	0.20	0.004	0.006	0.008
D	2.80	2.90	3.00	0.110	0.114	0.118
D1		1.90			0.075	
e		0.95			0.037	
E	2.60	2.80	3.00	0.102	0.110	0.118
F	1.50	1.60	1.75	0.059	0.063	0.069
L	0.10	0.35	0.60	0.004	0.014	0.024
K	0°		10°	0°		10°

### 7.3 SO8 package information

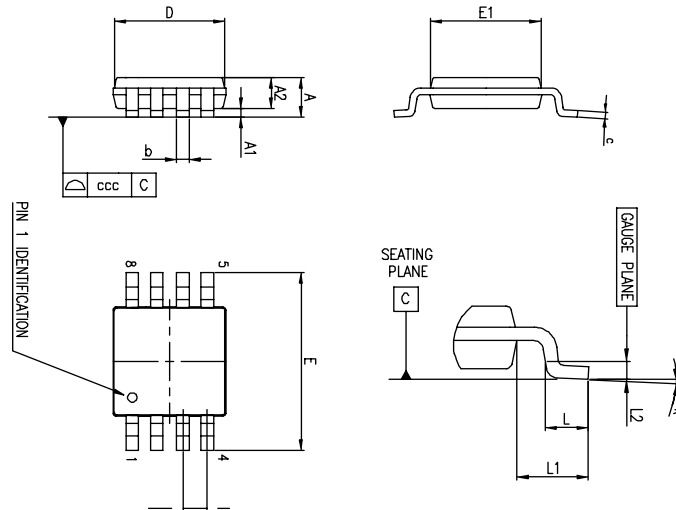
**Figure 65. SO8 package outline**

**Table 11. SO-8 mechanical data**

Symbol	Millimeters			Inches <sup>(1)</sup>		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A			1.75			0.069
A1	0.10		0.25	0.004		0.010
A2	1.25			0.049		
b	0.28		0.48	0.011		0.019
c	0.17		0.23	0.007		0.010
D	4.80	4.90	5.00	0.189	0.193	0.197
E	5.80	6.00	6.20	0.228	0.236	0.244
E1	3.80	3.90	4.00	0.150	0.154	0.157
e		1.27			0.050	
h	0.25		0.50	0.010		0.020
L	0.40		1.27	0.016		0.050
L1		1.04			0.040	
k	0°		8°	0°		8°
ccc			0.10			0.004

1. Values in inches are converted from mm and rounded to 4 decimal digits.



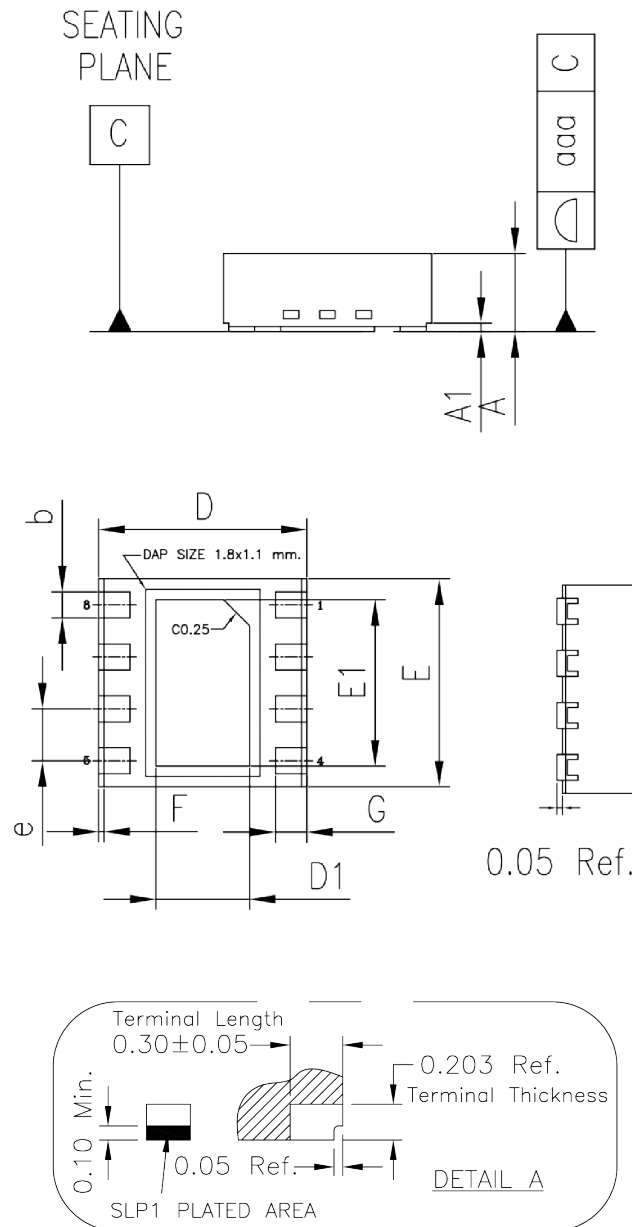
## 7.4 MiniSO8 package information

**Figure 66. MiniSO8 package outline**

**Table 12. MiniSO8 package mechanical data**

Ref.	Dimensions					
	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A			1.1			0.043
A1	0		0.15	0		0.0006
A2	0.75	0.85	0.95	0.030	0.033	0.037
b	0.22		0.40	0.009		0.016
c	0.08		0.23	0.003		0.009
D	2.80	3.00	3.20	0.11	0.118	0.126
E	4.65	4.90	5.15	0.183	0.193	0.203
E1	2.80	3.00	3.10	0.11	0.118	0.122
e		0.65			0.026	
L	0.40	0.60	0.80	0.016	0.024	0.031
L1		0.95			0.037	
L2		0.25			0.010	
k	0°		8°	0°		8°
ccc			0.10			0.004

## 7.5 DFN8 2 x 2 wettable flank package information

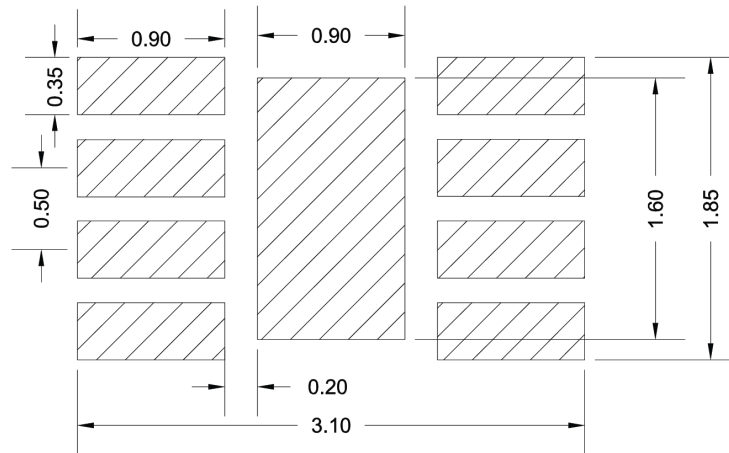
Figure 67. DFN8 2 x 2 wettable flank package outline



**Table 13. DFN8 2 x 2 wettable flank package mechanical data**

Ref.	Dimensions					
	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A	0.70	0.75	0.80	0.028	0.030	0.031
A1	0.10			0.004		
b	0.20	0.25	0.30	0.008	0.010	0.012
D	1.95	2.00	2.05	0.077	0.079	0.081
D1	0.80	0.90	1.00	0.031	0.035	0.039
E	1.95	2.00	2.05	0.077	0.079	0.081
E1	1.50	1.60	1.70	0.059	0.063	0.067
e		0.50			0.020	
F		0.05			0.002	
G	0.25	0.30	0.35	0.010	0.012	0.014
aaa		0.10			0.004	

**Figure 68. DFN8 2 x 2 wettable flank recommended footprint**



## 8 Ordering information

**Table 14. Order codes**

Order code	Temperature range	Package	Marking
TSZ151ICT	-40 °C to 125 °C	SC70-5	K4K
TSZ151ILT		SOT23-5	K234
TSZ151YCT	-40 °C to 125 °C automotive grade <sup>(1)</sup>	SC70-5	K4L
TSZ151YLT		SOT23-5	K235
TSZ152IDT	-40 °C to 125 °C	SO8	TSZ152I
TSZ152YDT	-40 °C to 125 °C automotive grade <sup>(1)</sup>		TSZ152IY
TSZ152IST	-40 °C to 125 °C	MiniSO8	K4M
TSZ152IYST	-40 °C to 125 °C automotive grade <sup>(1)</sup>		K4O
TSZ152IQ2T	-40 °C to 125 °C	DFN8	K4M
TSZ152IYQ2T	-40 °C to 125 °C automotive grade <sup>(1)</sup>		K4O

1. Qualification and characterization according to AEC Q100 and Q003 or the equivalent, advanced screening according to AEC Q001 & Q002 or the equivalent. For qualification status details, check the "Maturity Status" link on the first page ("Quality & Reliability" tab on [www.st.com](http://www.st.com)).

## Revision history

**Table 15. Document revision history**

Date	Revision	Changes
13-Oct-2023	1	Initial release.
20-Nov-2023	2	Added new Section 4.
22-Jan-2024	3	Updated figure and description on the cover page, <a href="#">Section 8</a> . Added new TSZ152 root part number, new <a href="#">Section 1.3</a> , <a href="#">Section 7.3</a> , <a href="#">Section 7.4</a> and <a href="#">Section 7.5</a> .

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