

## 2.5-GHz Double Balanced Mixer

### Description

U2795B is a 2.5-GHz mixer for WLAN and RF telecommunications equipment, e.g., DECT and PCN, built with TELEFUNKEN's advanced bipolar technology. A double balanced approach was chosen to assure good

isolation characteristics and a minimum of spurious products. The input and output are single ended, and their characteristics are programmable. No output transformer or balun is required.

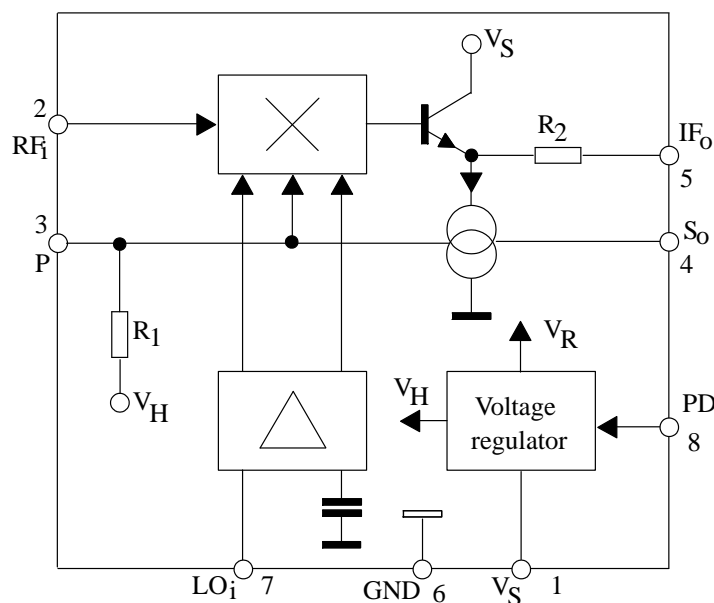
### Features

- Supply voltage range: 2.7 to 5.5 V
- Single-ended output, no balun required
- Single-ended input for RF and LO
- Excellent isolation characteristics
- Power down mode
- IP3 and compression point programmable
- 2.5-GHz operating frequency
- SO-8 package

### Benefits

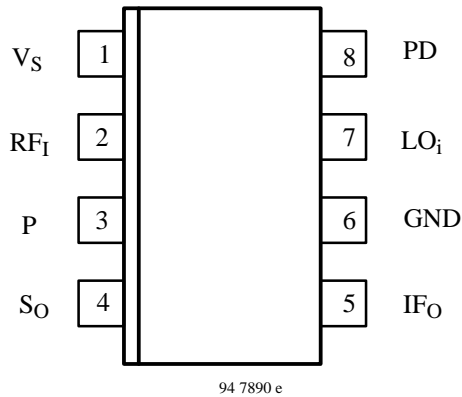
- Reduced system costs due to few external component (no balun) requirements
- Standard independent product
- 3-V operation reduces the battery count and saves space

### Block Diagram



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## Pin Description



Pin	Symbol	Function
1	$V_S$	Supply voltage
2	$RF_i$	RF input
3	P	Progammig port IP3, CP
4	$S_O$	Output symmetry
5	$IF_O$	IF output
6	GND	Ground
7	$LO_i$	LO input
8	PD	Power down

## Functional Description

### Supply Voltage

The IC is designed for a supply voltage of 2.7 to 5.5 V. As the IC is internally stabilized, the performance of the circuit is nearly independent of the supply voltage.

### Input Impedance

Input impedance,  $Z_{RF_i}$ , is about  $700 \Omega$  with an additional capacitive component. This condition provides the best noise figure in combination with a matching network.

### 3. Order Intercept Point (IP3)

Voltage divider,  $R_P / R_1$ , determinates both the input and output intercept point, IIP3 and OIP3. If  $R_P$  is infinity the IIP3 has the maximum of about  $-4 \text{ dBm}$ .

The IP3/ $R_P$  characteristics are shown in figure 1 and 2.

### Output Impedance and Intercept Point

Output impedance is shown in figure 9.

Both low output impedance and a high intercept point are with reference to a high value of  $R_P$ .

### Current Consumption, $I_S$

Depending on the chosen input and output conditions of the IC, the current consumption,  $I_S$ , is between 4 mA and 10 mA. The current consumption in dependence of  $R_P$  is shown in figure 4.

### Power Down

This feature provides an extension of battery life. If this function is not used, Pin 8 has to be connected to  $V_S$  (Pin 1).

### Output Symmetry

The symmetry of the load current can be matched and so be optimized for a given load impedance.

## Absolute Maximum Ratings

Parameters	Symbol	Value	Unit
Supply voltage Pin 1	$V_S$	6	V
Input voltage Pins 2, 3, 7 and 8	$V_I$	0 to $V_S$	V
Junction temperature	$T_j$	125	°C
Storage temperature range	$T_{stg}$	-40 to +125	°C

## Thermal Resistance

Parameters	Symbol	Value	Unit
Junction ambient SO 8	$R_{thja}$	175	K/W

## Operating Range

Parameters	Symbol	Value	Unit
Supply voltage range Pin 1	$V_S$	2.7 to 5.5	V
Ambient temperature range	$T_{amb}$	-40 to +85	°C

## Electrical Characteristics

$V_S = 3\text{ V}$ ,  $f_{LOi} = 1\text{ GHz}$ ,  $IF = 900\text{ MHz}$ ,  $RF = 100\text{ MHz}$ ,  $R_p = \infty$ , system impedance  $Z_o = 50\ \Omega$ ,  $T_{amb} = 25^\circ\text{C}$ ,  $R_T = 56\ \Omega$  reference point Pin 6, unless otherwise specified

Parameters	Test Conditions / Pin	Symbol	Min.	Typ.	Max.	Unit
Supply voltage range	Pin 1	$V_S$	2.7		5.5	V
Typical supply current range <sup>1</sup>	Pin 1	$I_S$	4		11	mA
Maximum supply current	Pin 1	$I_S$			13	mA
Conversion power gain	$R_L = 50\ \Omega$ , $R_T = \infty$ $R_L = 50\ \Omega$ , $R_T = 56\ \Omega$	$PG_C$		9		dB
		$PG_C$		4		dB
<b>Operating frequencies</b>						
$RF_i$ frequency	Pin 2	$RF_i$	10		2500	MHz
$LO_i$ frequency	Pin 7	$f_{LOi}$	50		2500	MHz
$IF_o$ frequency	Pin 5	$f_{IFo}$	50		2500	MHz
<b>Isolation</b>						
LO spurious at $RF_i$	Pin 7 to 2 $P_{iLO} = -10$ to $0\text{ dBm}$	$IS_{LO-RF}$		-30		dBm
$RF_i$ to $LO_i$	Pin 2 to 7 $P_{iRF} = -25\text{ dBm}$	$IS_{RF-LO}$		35		dB
LO spurious at $IF_o$	Pin 7 to 5, $P_{iLO} = -10$ to $0\text{ dBm}$	$IS_{LO-IF}$		-25		dBm
$IF_o$ to $LO_i$	Pin 5 to 7	$IS_{IF-LO}$		30		dB
<b>Output (IF)</b>						
Output compression point	Pin 5	$CP_o$		-10		dBm
<b>Input (RF)</b>						
Input impedance	Pin 2	$Z_{RFi}$		$700\ \Omega    0.8\text{pF}$		$\Omega$
Input compression point	Pin 2	$CP_i$		-14		dBm
Third order input intercept point	Pin 2	IIP3		-4		dBm
<b>Input (LO)</b>						
LO level	Pin 7	$P_{iLO}$		-6		dBm
<b>Voltage standing wave ratio (VSWR)</b>						
Input LO	Pin 7	$VSWR_{LOi}$		<2		
Output IF	Pin 4	$VSWR_{IFo}$		<2		
<b>Noise performance</b>						
Noise figure	$P_{iLO} = 0\text{ dBm}$ , $R_T = \infty$	NF		10		dB
<b>Power down mode</b>						
Supply current	Pin 1 $V_{PD} < 0.5\text{ V}$ Pin 1 $V_{PD} = 0\text{ V}$	$I_{SPD}$		<5	30	$\mu\text{A}$
<b>Power down voltage</b>						
“Power ON”	Pin 8 $V_S = 3.5$ to $5.5\text{ V}$ $V_S = 2.7$ to $3.5\text{ V}$	$V_{PON}$	$V_S - 0.5$ $V_S$		$V_S + 0.5$ $V_S + 0.5$	V V
“Power DOWN”	Pin 8	$V_{PDN}$			1	V
Power down current	Pin 8 Power ON Power DOWN	$I_{PON}$		0.15		mA
		$I_{PDN}$		<5		$\mu\text{A}$
Settling time	Pin 8 to 5	$t_{sPD}$		<30		$\mu\text{s}$

Note 1: Depending on  $R_p$

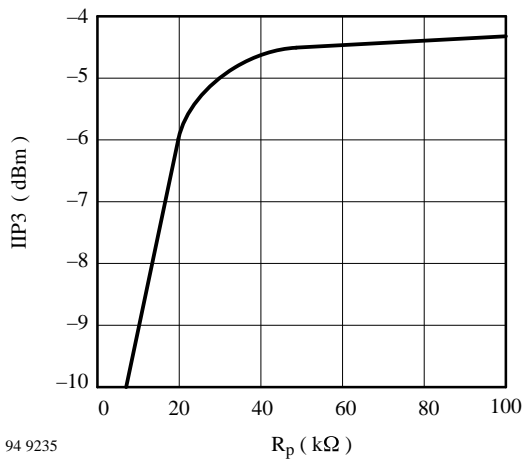


Figure 1. IIP3 versus resistor  $R_p$ , IF: 900 MHz

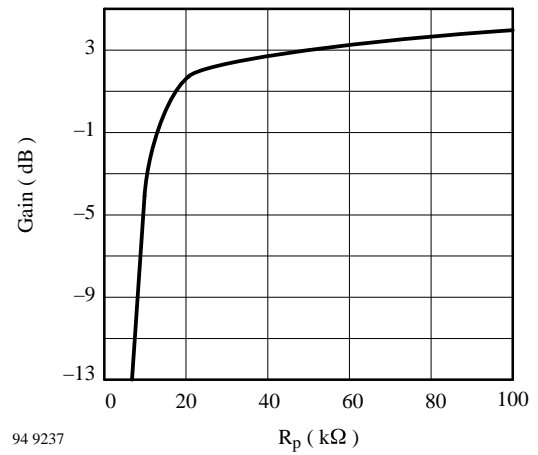


Figure 3. Gain versus resistor  $R_p$ , LO: 1030 MHz, level  $-10$  dBm; RF: 130 MHz,  $-30$  dBm,  $R_T = 56 \Omega$

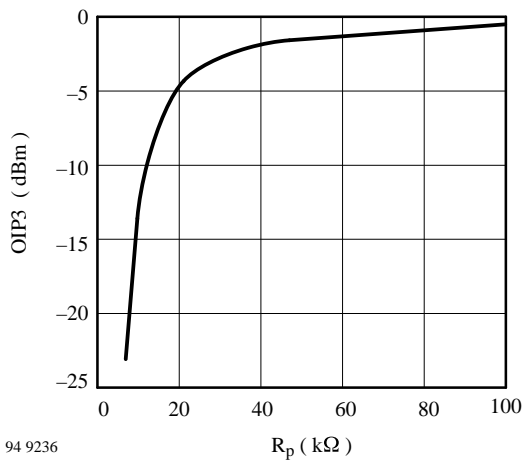


Figure 2. OIP3 versus resistor  $R_p$ , IF: 900 MHz

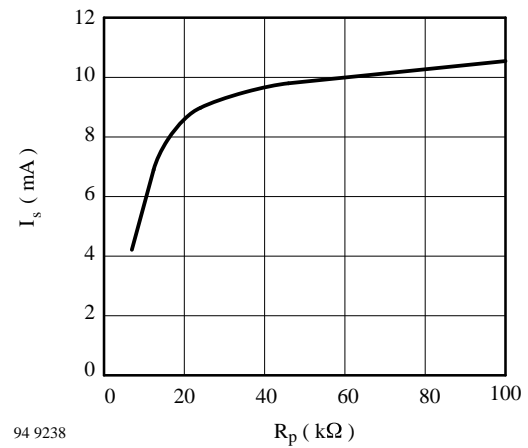


Figure 4. Supply current  $I_S$  versus resistor  $R_p$

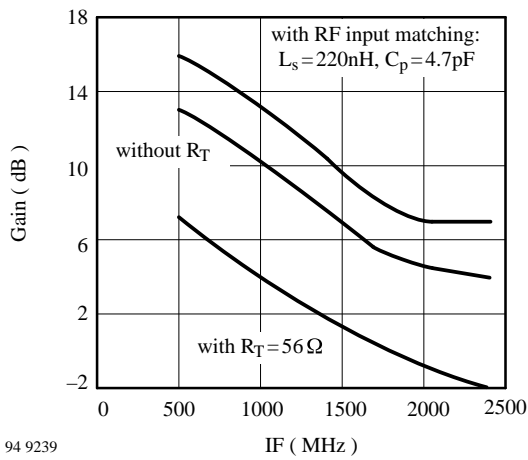


Figure 5. Gain versus IF output frequency, LO level: -6 dBm, RF: 130 MHz, -35 dBm; parameter: RF input termination

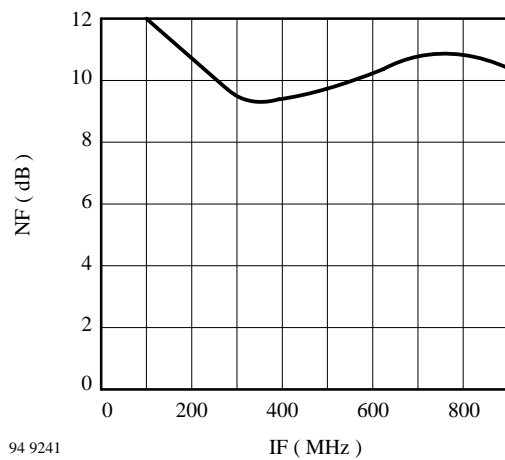


Figure 7. Double sideband noise figure versus IF output frequency; LO: 1000 MHz, level 0 dBm; no RF input matching,  $R_T$  left out

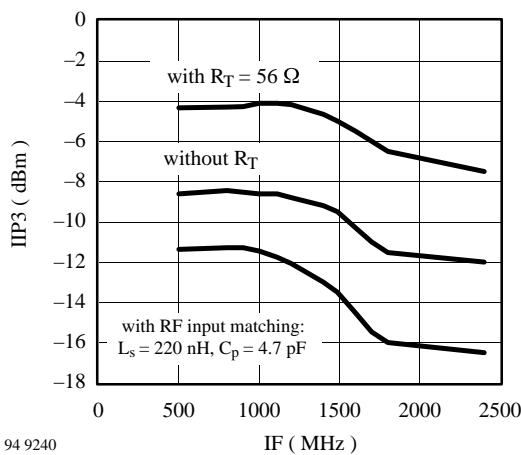


Figure 6. IIP3 versus IF output frequency, LO level: -6 dBm; RF: 130 MHz / 130.1 MHz, -35 dBm; parameter: RF input termination

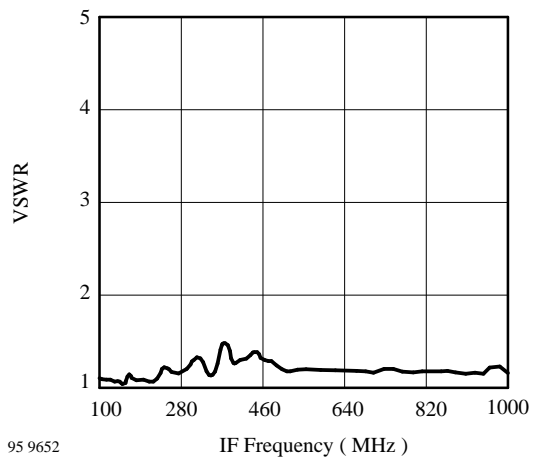


Figure 8. Typical VSWR frequency response of the IF output,  $R_p = \infty$

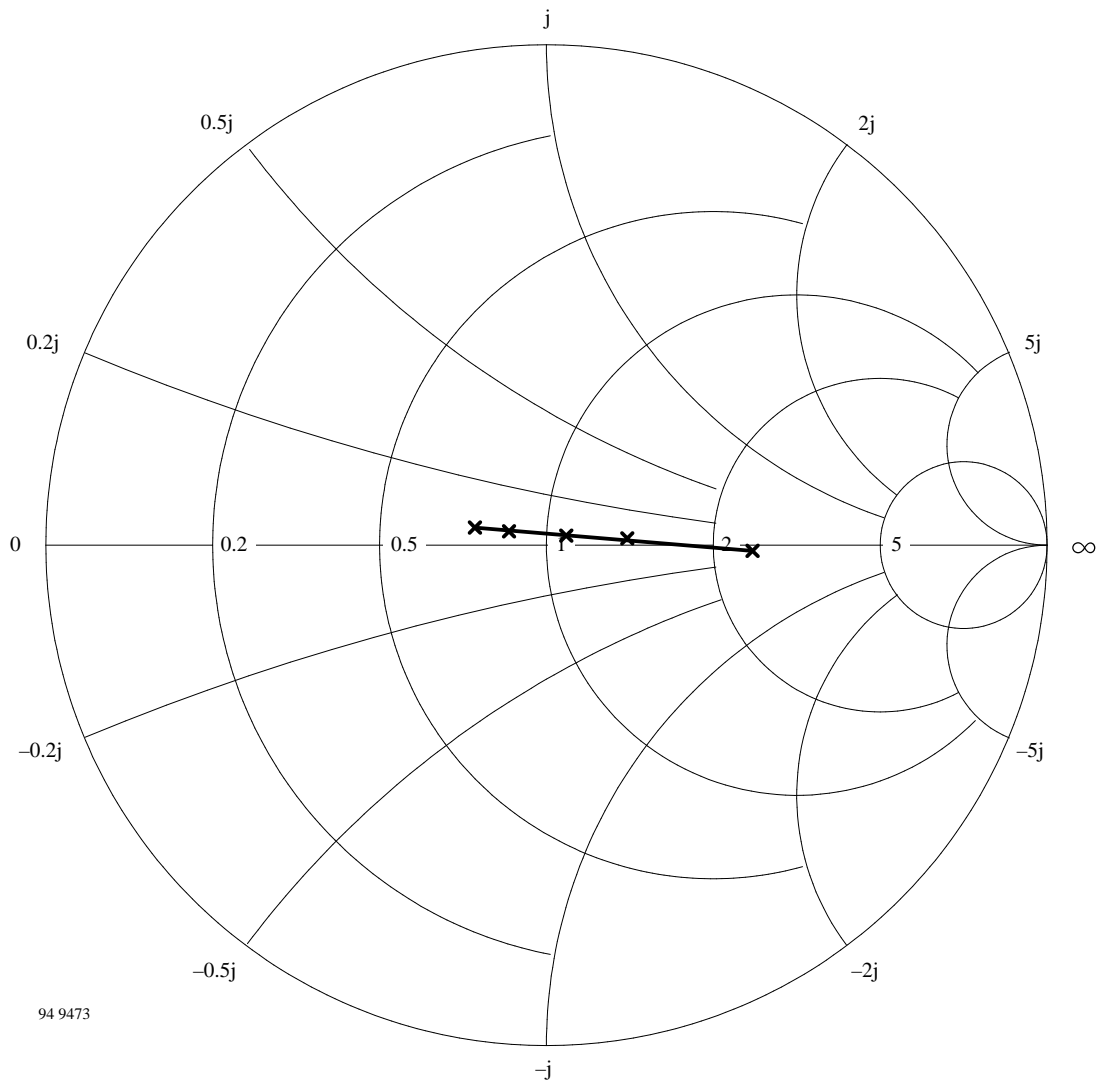


Figure 9. Typical Impedance of the output versus  $R_P$  at frequency  $f_{IF0} = 900$  MHz  
 markers (from left to right):  $R_P = \infty / 22 \text{ k}\Omega / 10 \text{ k}\Omega / 8.2 \text{ k}\Omega / 5.6 \text{ k}\Omega$

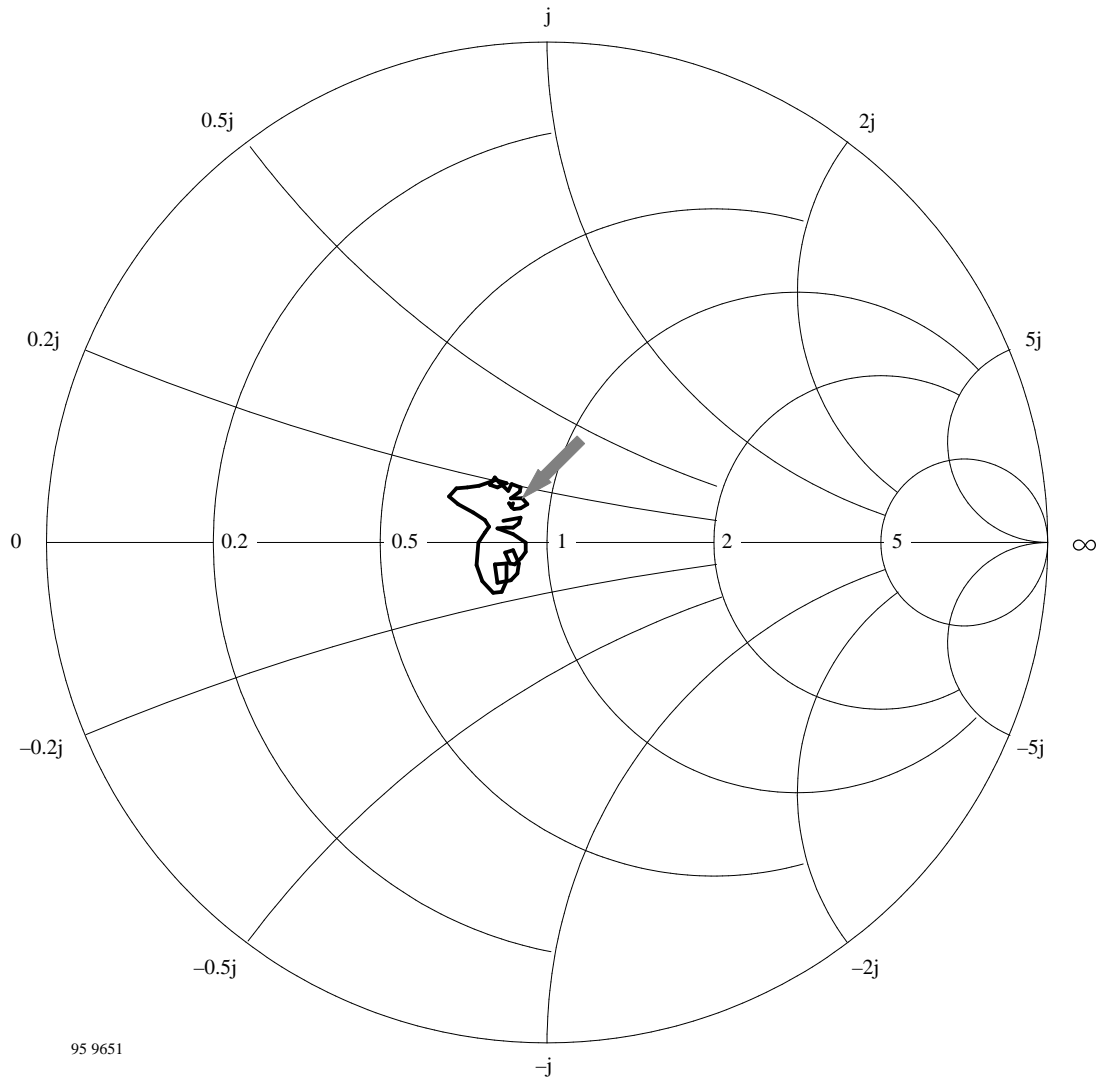


Figure 10. Typical S11 frequency response of the IF output,  $R_p = \infty$ ,  
IF frequency from 100 MHz to 1000 MHz, marker: 900 MHz



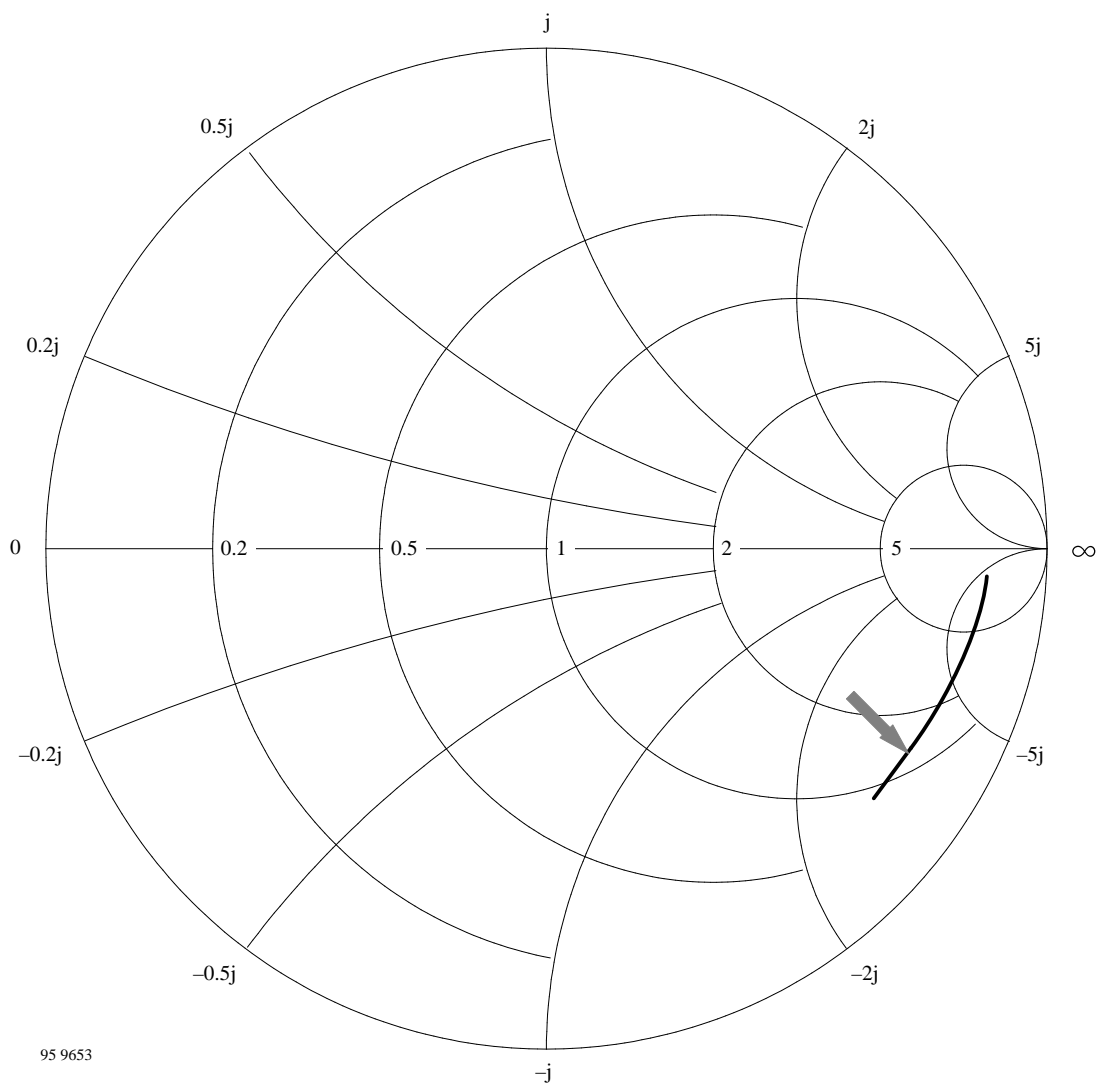


Figure 11. Typical S11 frequency response of the RF input,  $R_P = \infty$ ,  $R_T = \infty$   
RF frequency from 100 MHz to 1000 MHz, marker: 900 MHz

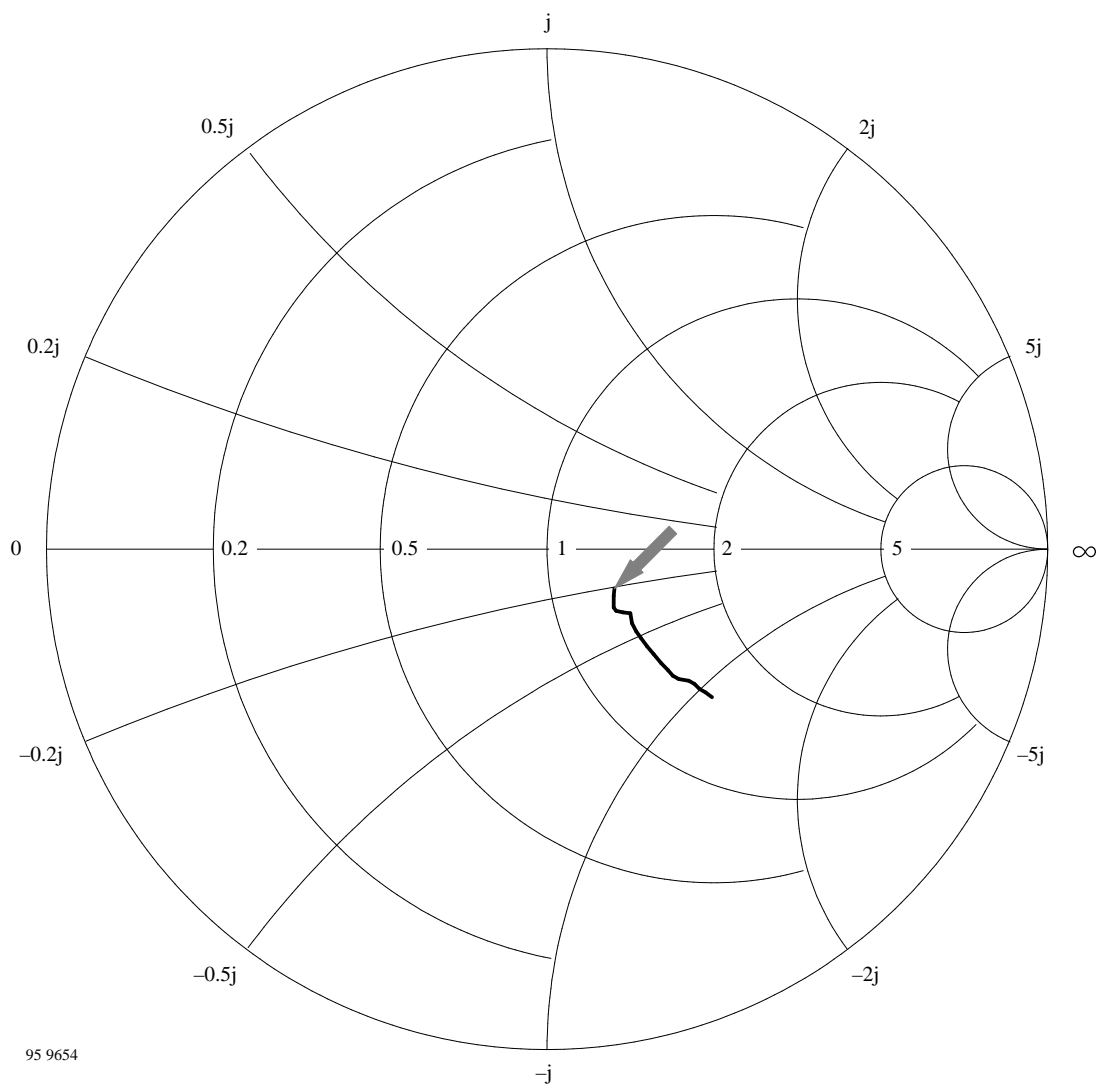
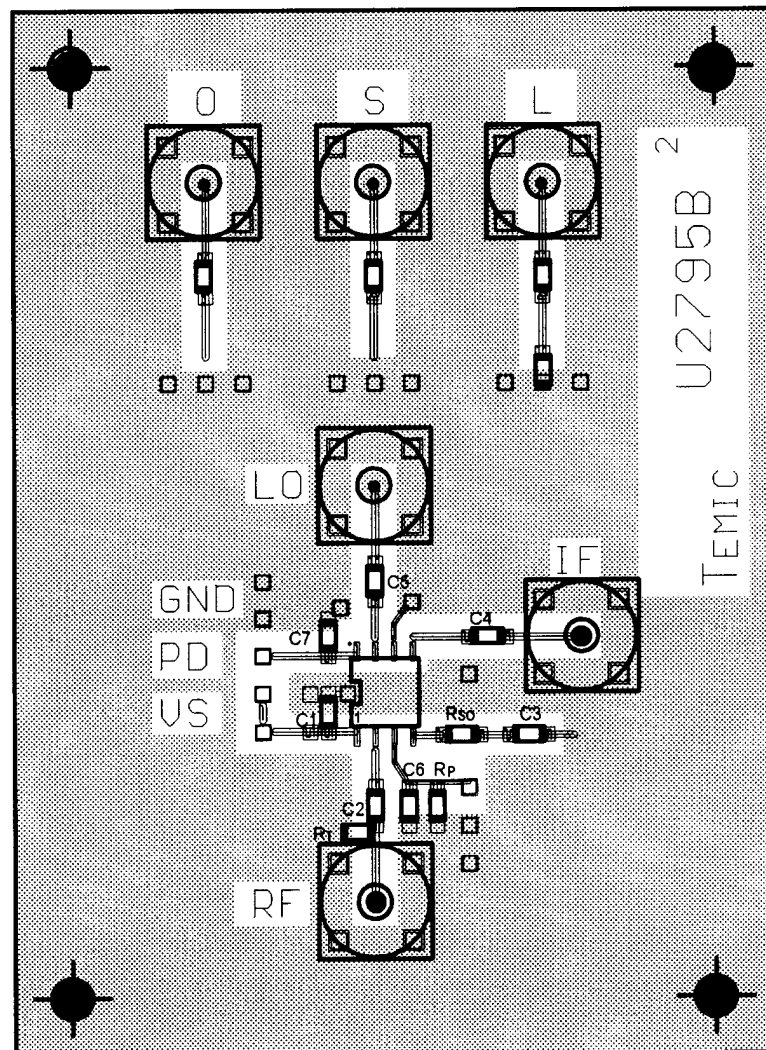


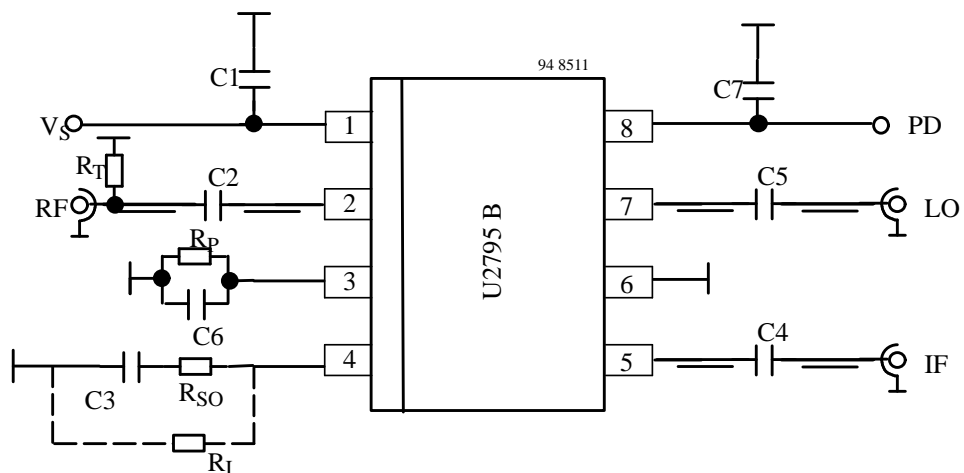
Figure 12. Typical S11 frequency response of the LO input,  $R_p = \infty$ ,  
LO frequency from 100 MHz to 1000 MHz, marker: 900 MHz

### Application Circuit (Evaluation Board)



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### Application



## U2795B

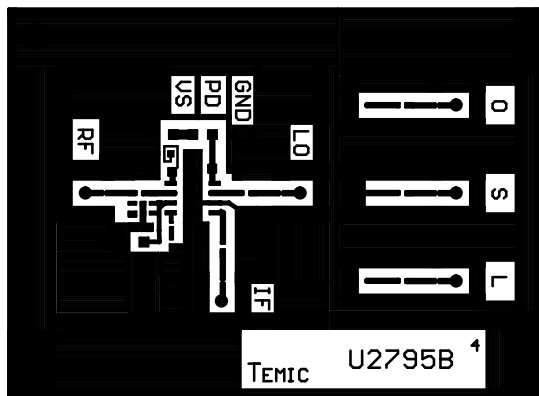
Part List	
C 1	10 nF
C2, C3, C4, C5, C6, C7	100 pF
*R <sub>P</sub>	
=====	50-Ω Microstrip
*R <sub>SO</sub>	68 Ω
— — — —	optional
R <sub>T</sub>	56 Ω

With the part list values, the PD settling time is  $< 20 \mu\text{s}$ . Using other values, time requirements in burst-mode applications have to be considered.

Values of R<sub>SO</sub> and R<sub>P</sub> depending on the input and output condition requirements. For R<sub>SO</sub> 68 Ω is recommended.

With the optional R<sub>I</sub> the intercept and compression point can be slightly increased; values between 500 Ω and 1 kΩ are suitable. Please note that such modification will also increase the supply current.

### Application Board



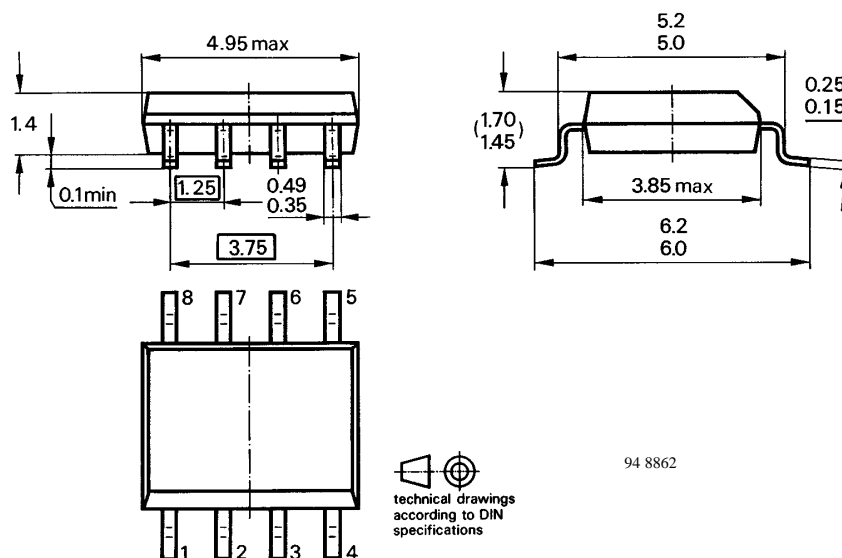
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### Ordering Information

Extended Type Number	Package
U2795B-FP	SO 8

### Dimensions in mm

Package: SO 8



technical drawings  
according to DIN  
specifications

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## Ozone Depleting Substances Policy Statement

It is the policy of **TEMIC TELEFUNKEN microelectronic GmbH** to

1. Meet all present and future national and international statutory requirements.
2. Regularly and continuously improve the performance of our products, processes, distribution and operating systems with respect to their impact on the health and safety of our employees and the public, as well as their impact on the environment.

It is particular concern to control or eliminate releases of those substances into the atmosphere which are known as ozone depleting substances (ODSs).

The Montreal Protocol (1987) and its London Amendments (1990) intend to severely restrict the use of ODSs and forbid their use within the next ten years. Various national and international initiatives are pressing for an earlier ban on these substances.

**TEMIC TELEFUNKEN microelectronic GmbH** semiconductor division has been able to use its policy of continuous improvements to eliminate the use of ODSs listed in the following documents.

1. Annex A, B and list of transitional substances of the Montreal Protocol and the London Amendments respectively
2. Class I and II ozone depleting substances in the Clean Air Act Amendments of 1990 by the Environmental Protection Agency (EPA) in the USA
3. Council Decision 88/540/EEC and 91/690/EEC Annex A, B and C (transitional substances) respectively.

**TEMIC** can certify that our semiconductors are not manufactured with ozone depleting substances and do not contain such substances.

**We reserve the right to make changes to improve technical design and may do so without further notice.**

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