**U3750BM** 

## **One Chip Telephone Circuit**

#### **Description**

TELEFUNKEN microelectronic's one chip telephone circuit, U 3750 BM, is BICMOS integrated circuit that performs all the speech and line interface functions required in an electronic telephone set, the tone ringer, the

pulse and DTMF dialling with redial, a keyboard interface with the possibility to interface with an external microcontroller using the internal serial bus, and a power supply for peripherals.

#### **Features**

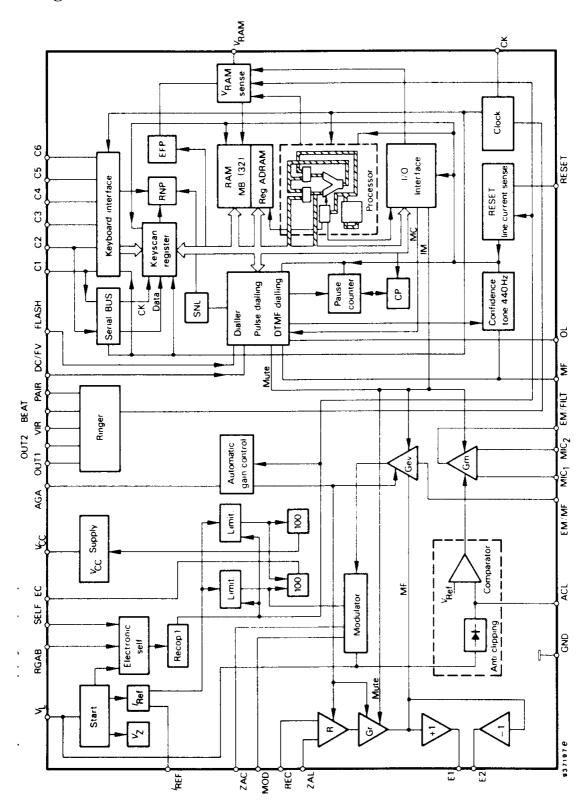
- Adjustable de slope characteristic
- Adjustable automatic line length receiving and sending gain control (not used in DTMF) with the possibility of fixed gain (PABX)
- Adjustable dynamic impedance
- Stabilized power supply for peripherals
- Confidence level during pulse and DTMF dialling
- Receiving amplifier for dynamic or piezo-electric ear pieces
- High-impedance microphone inputs (80 k $\Omega$  in symmetrical and 40 k $\Omega$  in asymmetrical) suitable for dynamic, magnetic, piezo-electric or electret microphone
- Dynamic limiting in sending (anticlipping) prevents distortion of line signal and sidetone
- Ringing balanced output in HVMOS for higher power capability

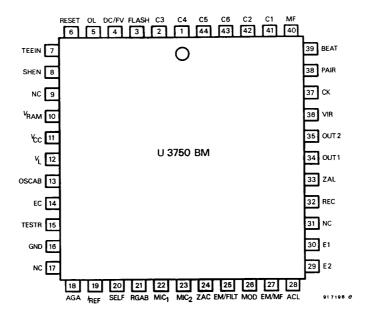
- Four ringing tones adjustable without external components
- Internal speed up circuit permits a faster charge of V<sub>CC</sub> and V<sub>RAM</sub> capacitors
- Logic bounce elimination
- Pulse dialling 66/33 or 60/40 or DTMF dialling selectable by programming pin
- Adjustable flashing duration
- Pause function
- Confidence tone (440 Hz)
- Last number redial up to 23 digits
- Standard low-cost ceramic 455 kHz
- Binary data input in serial mode
- Test-mode capability

#### **Benefits**

- Low number of external components
- High quality through one chip solution

## **Block Diagram**



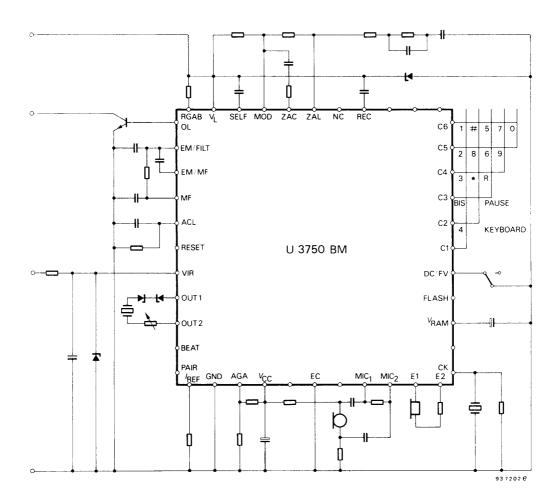


## **Pin Description**

Pin	Symbol	Function
1	C4	Keyboard input
2	C3	Keyboard input
3	FLASH	Flashing selection
4	DC/FV	Dialling selection (33/66 pulse, 40/60 pulse or DTMF)
5	OL	Open line output
6	RESET	Output reset
7	TEEIN	Test pins
8	SHEN	Test pins
9	NC	Not connected
10	$V_{RAM}$	RAM and internal logic supply
11	$V_{CC}$	Power supply for peripherals
12	$V_{\rm L}$	Line voltage
13	OSCAB	Test pin
14	EC	Extra current for peripherals or can be used to dissipate power for high line current applications.
15	TESTR	Test pin
16	GND	Ground
17	NC	Not connected
18	AGA	Line length AGC adjustment
19	I <sub>REF</sub>	Bias adjustment
20	SELF	Electronic self input
21	RGAB	DC characteristic slope adjustment
22	MIC <sub>1</sub>	Microphone input
	MICI	microphone input

Pin	Symbol	Function
23	MIC <sub>2</sub>	Microphone input
24	ZAC	Dynamic impedance adjustment
25	EM/FILT	First sending stage output
26	MOD	Modulator output
27	EM/MF	Second sending stage input and DTMF input
28	ACL	Anticlipping time constant adjustment
29	E2	Receiver output
30	E1	Receiver output
31	NC	Not connected
32	REC	Receiver input
33	ZAL	Sidetone network
34	OUT1	Buzzer output
35	OUT2	Buzzer output
36	VIR	Ringing supply
37	CK	Ceramic input (455 KHz)
38	PAIR	Adjustment between two pairs of ringing frequencies
39	BEAT	Beat adjustment of each pair of ringing frequencies
40	MF	DTMF output
41	C1	Keyboard inputs
42	C2	Keyboard inputs
43	C6	Keyboard inputs
44	C5	Keyboard inputs

## **Application Circuit**



## **Absolute Maximum Ratings**

See application circuit

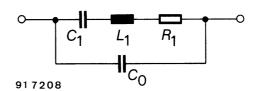
Parameters	Symbol	Value	Unit
DC line voltage Pin 36	VIR	35	V
DC line current Pin 36	IR	30	mA
Conversation line voltage Pin 12	$V_{ m L}$	15	V
Pulse duration, $t = 20 \text{ ms}$	$V_{ m L}$	17	V
Conversation line current Pin 12	$I_{L}$	150	mA
Power dissipation, $T_{amb} = 55^{\circ}C$	$P_{tot}$	1	W
Junction temperature	$T_{j}$	125	°C
Ambient temperature range	$T_{amb}$	-25 to +55	°C
Storage temperature range	$T_{stg}$	-55 to +155	°C

#### **Thermal Resistance**

Parameters	Symbol	Value	Unit
Junction ambient	$R_{thJA}$	70	K/W

## **Electrical Characteristics**

 $f=1~kHz,~f_{clock}=455~kHz,~R_E=20~k\Omega,~T_{amb}=25^{\circ}C,~$  unless otherwise specified, Q (Resonance factor) = 3100,  $L_1=6.1~mH,~C_1=21~pF,~C_O=268.5~pF,~R_1=5.5~\Omega.$  All resistances are specified at 1%, all capacitance at 2%.



Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Line voltage normal Line voltage operation	$\begin{split} I_L &= 8 \text{ mA} \\ I_L &= 15 \text{ mA} \\ I_L &= 28 \text{ mA} \\ I_L &= 60 \text{ mA} \\ \text{figure 3} \end{split}$	$V_L$	2.9 4.1 6.4 12.3	4.5	3.6 4.9 7.3 13.7	V
Stabilized voltage	$I_{L} = 8 \text{ mA}, I_{CC} = 0.6 \text{ mA}$ $I_{L} \ge 28 \text{ mA}, I_{CC} = 2.1 \text{ mA}$ figure 3	V <sub>CC</sub>	2.0 3.3	2.5 3.5	3.7	V
Transmission						•
Sending gain  AGC	$\begin{aligned} V_{MI} &= 2 \text{ mV}_{RMS} \text{ (note 1)} \\ I_L &= 28 \text{ mA (G}_S \text{ max)} \\ I_L &= 60 \text{ mA (G}_S \text{ min)} \\ I_L &= 28 \text{ to } 60 \text{ mA} \\ &\qquad \qquad \text{figure 3} \end{aligned}$	$G_{ m S}$ $\Delta G_{ m S}$	47 39.5 –6	48 41 -7	49 42.5 –8	dB
Psophometric sending noise	$V_{MI} = 0$ , $I_L = 28$ mA figure 3			-73	-68	dBmp
Attenuation gain during dialing	$V_{MI} = 2 \text{ mV}_{RMS},$ $I_L = 28 \text{ mA}$ figure 3	As	63			dB
Microphone input impedance (Pins 22–23)	figure 3		70	120		kΩ
Common mode rejection ratio	I <sub>L</sub> = 28 mA figure 3	CMRR		80		dB
From transmission to dialing mode	I <sub>L</sub> = 28 to 60 mA figure 3, Pin 25	Step	-100		+100	mV
Dynamic limiter (anticlipping) Output voltage swing (peak-to-peak value)	$\begin{split} C_{ACL} &= 470 \text{ nF}, \\ R_{ACL} &= 6.8 \text{ M}\Omega \\ I_L &= 20 \text{ mA} \\ I_L &\geq 28 \text{ mA}, \\ V_{MI} &= 8 \text{ mV}_{RMS} \\ \text{figure 3} \end{split}$		3.0	2.5 3.6	4.2	V <sub>pp</sub>
Overdrive dynamic range	$I_L \ge 28 \text{ mA}$ figure 3				5	dB
Line distortion (on $600 \Omega$ )	$I_{L} \ge 28 \text{ mA}$ $V_{MI} = 4.6 \text{ mV}_{RMS}$ $V_{MI} = 8 \text{ mV}_{RMS}$ $V_{MI} = 80 \text{ mV}_{RMS}$ $\text{figure 3}$				3 5 5	%

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## **TELEFUNKEN Semiconductors**

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Available current	$\begin{array}{c} \text{Close switch S4} \\ I_L = 28 \text{ mA} \\ I_L = 60 \text{ mA} \\ \text{figure 3} \end{array}  \text{Pin 14} \\$		7.0 35	8 40		mA
Reception	inguie 5					
Receiving gain $G_R = V_R/V_L$ AGC	$\begin{aligned} V_L &= 0.3 \ V_{RMS} \\ I_L &= 28 \ \text{mA (} G_R \ \text{max)} \\ I_L &= 60 \ \text{mA (} G_R \ \text{min)} \\ I_L &= 28 \ \text{to } 60 \ \text{mA} \\ &\qquad \qquad \text{figure } 4 \end{aligned}$	$G_R$ $\Delta G_R$	10 2.5 -6	11 4 -7	12 5.5 -8	dB
Psophometric receiving noise	$V_L = 0V$ , $I_L = 28 \text{ mA}$ figure 4 Pin 29–30				-65	dBmp
Receiving distortion	$\begin{split} I_L &= 15 \text{ mA},  V_R = 2.8  V_{pp} \\ I_L &= 28 \text{ mA},  V_R = 5.5  V_{pp} \\ I_L &= 60  \text{mA},  V_R = 5.0  V_{pp} \\ \text{figure 4} \end{split}$				3 3 3	%
Receiver output impedance	$V_R = 50 \text{ mV}_{RMS},$ $I_L = 28 \text{ mA}$ figure 4 $Pin 29-30$		45	65	85	Ω
Receiver output offset	$I_L = 28 \text{ mA}$ figure 4 Pin 29–30		-650		+650	mV
Sidestone (V <sub>R</sub> /V <sub>M</sub> )	I <sub>L</sub> = 28 mA figure 3			36	40	dB
Z line matching impedance	$\begin{aligned} V_L &= 0.3 \ V_{RMS} \\ I_L &= 28 \ \text{and} \ 60 \ \text{mA} \\ & \text{figure} \ 4 \end{aligned}$		580	660	750	Ω
Ringer		•				
Turn on voltage	Measured at pin VIR figure 5	V <sub>ON</sub>		16	18.0	V
Turn off voltage	figure 5	V <sub>OFF</sub>	9.5	10.5		V
Current consumption without load	VIR = 18 V			1.2	1.5	mA
Output voltage swing	Load = $10 \text{ k}\Omega$ figure 5	V <sub>OUT</sub>	VIR-2			Vp
Output tone frequencies	Pin 38 grounded Pin 38 open			1458 1166 547 438		Hz
Sweep frequencies	Pin 39 grounded Pin 39 open			4.0 9.1		Hz
Leakage current	$VIR = 30 V$ $II_L \text{ at } VI_L = 0 V$ $Pins 38 \text{ and } 39$ $figure 5$				5	μΑ

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
DTMF generation Pin 4 gr	rounded					
Tone frequency accuracy (confidence tone included)	$V_{RAM} = 3.5 \text{ V}$		-0.4		+0.25	%
Low group tone level (depends on external components)	(Note 2) Measured on 600 $\Omega$ $I_L = 28 \text{ mA}$ figure 3		-10	-8	-6	dBm
High group tone level (depends on external components)	(Note 2) Measured on 600 $\Omega$ I <sub>L</sub> = 28 mA figure 3		-8	-6	-4	dBm
Preemphasis (depends on external components)	(Note 2) Measured on 600 $\Omega$ $I_L = 28 \text{ mA}$ figure 3		1	2	3	dB
DTMF distortion (depends on external components)	(Note 2 and note 3) Measured on 600 $\Omega$ I <sub>L</sub> = 28 mA, 300 < f < 3400 Hz figure 3				3.5	%
DTMF transmission time DTMF interdigit time		t <sub>MF</sub> t <sub>IMF</sub>	80.2 89.2		82.4 89.2	ms ms
Transmission mute tm <sub>MF</sub>			169.4		171.6	ms
Confidence tone	Only by serial bus					
FCT frequency				440.9		Hz
Tone level (depends on external components)	$\begin{tabular}{ll} Measured on 600 $\Omega$ \\ I_L = 25 mA \\ figure 3 \end{tabular}$			-9		dBm

Note 1: sending gain:  $G_S = V_L/V_{MI}$  with the values of  $R_{AG1}$  and  $R_{AG2}$  (figure 3) so the maximum gain is at 28 mA and the minimum gain is at 60 mA.

**Note 2**: For DTMF measurements, close switches S1 and S3 and select each group of frequencies on the keyboard.

**Note 3**: The level of each harmonic on line is under the limited curve given below with the filter components value chosen for the test.

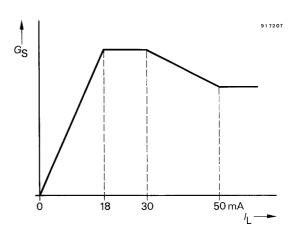


Figure 1

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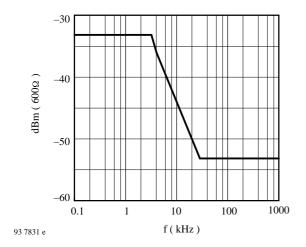


Figure 2 DTMF distortion

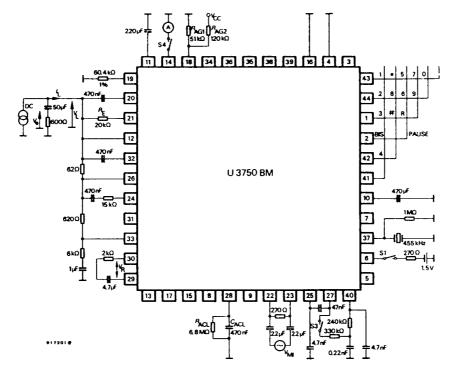


Figure 3 Test circuit

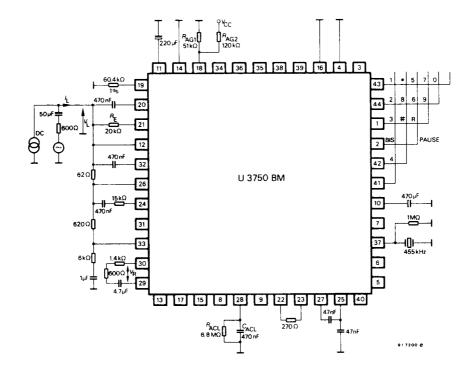


Figure 4 Test circuit

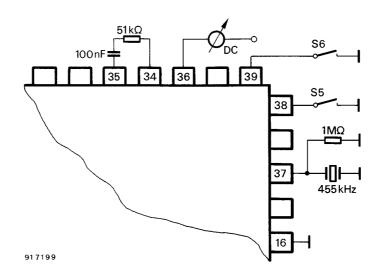


Figure 5 Test circuit

S <sub>5</sub> (Pin 38)	S <sub>6</sub> (Pin 39)	Pair of Frequencies (Hz)	Sweep Frequency (Hz)
ON	ON	1458/1166	4
ON	OFF	1458/1166	9.1
OFF	ON	547/438	4
OFF	OFF	547/438	9.1

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## **Electrical Characteristics of Logical Part**

 $f_{clock} = 455 \ kHz$  (other specifications as under electrical characteristics)

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
V <sub>RAM</sub> Speed-up off threshold Speed-up on threshold Logic operating voltage in normal mode	$I_L = 8 \text{ mA}$ $I_L = 15 \text{ to } 70 \text{ mA}$	V <sub>SOFF</sub> V <sub>SON</sub>	2.4 1.9	2.75 2.2 2.5 3.5	2.9 2.3	V V V
I <sub>RAM</sub> Oscillator on Leakage	$V_{RAM} = 3.5 \text{ V}$				800 300	μ <b>Α</b> μ <b>Α</b>
Inputs: C1, C2, C3, C4, C5, C6, DC/FV, FLASH	Input voltage low, V <sub>IL</sub> Input voltage high, V <sub>IH</sub>		0.8 V <sub>RAM</sub>		$V_{\rm RAM}$	V V
Keyboard pins: C1, C2, C3, C4, C5, C6 Internal pull down Output current	$V_{IL} = 3.5 \text{ V}$ $V_{IH} = 0 \text{ V}$		15 0.8		50 2.5	μ <b>Α</b> μ <b>Α</b>
FLASH, DC/FV Internal pull-up current IPV Leakage current	$\begin{vmatrix} V_{IL} = 0 \text{ V} \\ V_{IH} = 3.5 \text{ V} \end{vmatrix}$		0.5		5 1	μ <b>Α</b> μΑ
Timing and frequency Reset time t <sub>r</sub> (see figure 6 and 7) Clock start-up time t <sub>on</sub> Time line break generating	In mode 60/40 In mode 66/33 and DTMF mode			30 33 5		ms ms ms
a reset: t <sub>lb</sub> Debounce time, t <sub>e</sub>	In mode 60/40 In mode 66/33 and DTMF mode In mode 60/40 In mode 66/33 and DTMF mode		290 319 14 15.4	24 26.4	300 330 34 37.4	ms ms ms
RESET output (with 390 Ω series) Output low current Output high current	V <sub>OL</sub> = 2.5 V V <sub>OL</sub> = 0.5 V	I <sub>OL</sub> I <sub>OH</sub>	0.25 0.25		1.2 1.2	mA mA
OL output Output low current Output high current	$V_{OL} = 0.5 \text{ V}$	I <sub>OL</sub>	2		20 4	μA mA
MF output High impedance	$V_{OHI} = 1.4 \text{ V} \\ FL = L, FH = H, V_{OH} = 3.5 \text{ V} \\ FB = L, FH = H, V_{OH} = 0 \text{ V} \\ FB = H, FH = L, V_{OH} = 3.5 \text{ V} \\ FB = H, FH = L, V_{OH} = 0 \text{ V} \\ \label{eq:VOH}$		150 200 200 200 150		0.5 350 550 550 350	μΑ μΑ μΑ μΑ μΑ
CK Input Low input leakage High input leakage	$V_{IL} = 0.5 \text{ V} $ $V_{IL} = 3.0 \text{ V}$				1 1	μ <b>Α</b> μ <b>Α</b>

## **TELEFUNKEN Semiconductors**

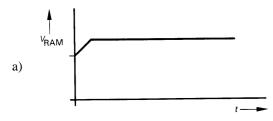
## **U3750BM**

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Serial bus (see figure 12)						
Pulse width clock		t <sub>wl</sub> , t <sub>wh</sub>	2			μS
Pulse width enable signal		t <sub>el</sub> , t <sub>eh</sub>	2			μS
Set-up time data to clock		t <sub>set up</sub>	0			μS
Hold time data from clock			100			μS
Enable time		t <sub>e</sub>	0			μS
Time between two trans-		t <sub>RRN</sub>	900			μS
missions Pulse dialing (OL)	In mode 60/40 (Pin 4 tied to			10		Hz
Dialing pulse frequency	RESET)			10		ПZ
Diamig pulse frequency	In mode 66/33 (Pin 4 not			10.11		Hz
	connected)			10.11		HZ
Dialing pulse period		T <sub>OL</sub>		100		ms
		OL OL		98.9		ms
Break time		t <sub>b</sub>		60		ms
				66		ms
Make time		t <sub>m</sub>		40		ms
				33		ms
Interdigit time		t <sub>IDOL</sub>	830		833	ms
			813		816.5	ms
Transmission mute:			[n·100		(n · 100)	ms
$t_{mol} = (t_m + t_b) \cdot n + t_{m1}$			+ 30]		+ 32	
n pulses dialling			[n·98.8		(n·98.9)	ms
			+ 22]		+ 24.2	
Flash pulse	D' 2 CMD D' 4 DEGET		00.5		0.2	
Flash pulse duration	Pin 3 to GND Pin 4 to RESET	$t_{\rm fl}$	89.5		92	
	Pin 3 to GND Pin 4 to NC		98.5		101	
	Pin 3 to GND Pin 4 to GND		102.5		105	
	Pin 3 to NC Pin 4 to RESET Pin 3 to NC Pin 4 to NC		239.5 263.5		242 266	****
	Pin 3 to NC Pin 4 to NC		274.0		277	ms
	Pin 3 to RESET Pin 4 to RE-		109.5		112	
	SET		120.5		123	
	Pin 3 to RESET Pin 4 to NC		120.3		123	
Transmission mute	Pin 3 to RESET Pin 4 to NC	tmfl	830		832	
Transmission muc	In mode 60/40	unil	813		815.5	
	In mode 66/33		846		848.5	
	In mode DTMF				0.0.5	
Pause time	In mode 60/40		3116		3118	
	In mode 66/33	Tp	3075		3077	ms
	In DTMF mode	1	3012		3110	

#### **Power-on Reset and Pin RESET**

To avoid undefined states of the device when it is powered on, an internal reset clears the control logic. When the power supply rises above the internal reference level, the pin RESET goes to high during  $t_{rt}$ . After a line break longer than  $t_{lb}$ , a reset is generated. A short line break (<  $t_{lb}$ ) does not affect the reset. Power on reset timings (t >  $t_{lb}$ ).

1) 
$$V_{RAM} > V_{SON}$$
 at  $t = 0 \dots a$ 



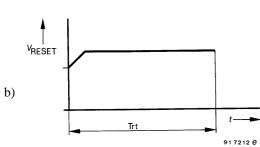
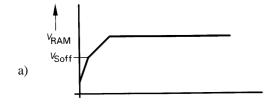


Figure 6

$$V_{RAM} < V_{SON}$$
 at  $t = 0 \dots a$ 

$$Trt = T (V_{RAM} \text{ to } V_{S \text{ off}}) + Tr + T \text{ on } \dots b)$$



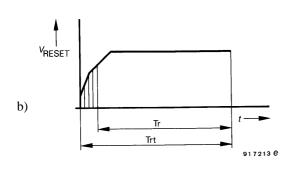


Figure 7

#### Pin RESET

It is the power on reset output.

In test mode, it permits to force the IC U 3750 BM in permanent DTMF dialing by applying a negative voltage (test schematic figure 3).

## **Data Acquisition**

Input data is derived from any standard matrix keyboard (15 keys) or from a remote microcontroller.

## **Keyboard**

The keyboard is connected to the IC by six pins, (see figure 8). Its matrix is triangular.

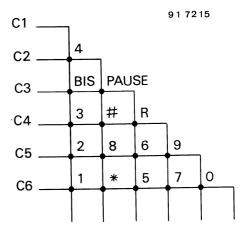


Figure 8 Keyboard inputs to the U 3750 BM

Internal pull down resistors (typical value: 120  $k\Omega$  ) are connected to the inputs Ci.

A push button is made by a short-circuit of two pins among the six.

Entries are scanned every 12 ms or 13.2 ms and go to the logical state 1 during this scanning.

The scanning is inhibited as soon as a calibrated linebreak is produced at the OL output.

The scanning-cycle has eight phases: six of them are reserved for the scanning of the six pins, the two others are kept for the reset of the logic keyboard (T0) and the acquisition (T7).

A push button is valid, if it is unique and if it's pressed long enough (see table of pressed and released push buttons).

Every acquisition time, T7, the input code is decoded into 5 bits (with or without a pressed push button). The microprocessor can read it as soon as the logic keyboard has set a flag, tested about every 800 ms and which indicates that a pushbutton has been correctly detected.

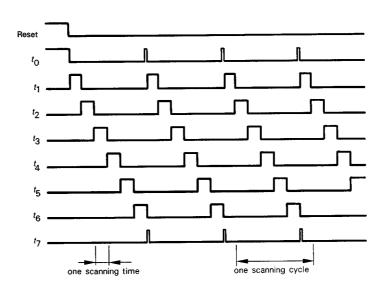
Table 2 The scanning principle

Keyboard clock	2 ms	2.2 ms
Ti except T0, T1, T7	2 ms	2.2 ms
T1	1.956 ms	2.156 ms
T0 or T7	22 ms	22 ms
Scanning cycle	12 ms	13.2 ms

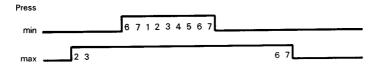
T0: reset of the logic keyboard

T7: acquisition of the code present at the keyboard





#### Push button



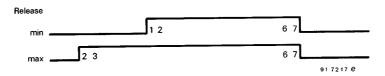


Figure 9 The scanning principle

#### **Timing of a Push Button**

The information from a pressed push button or released push button is taken into account if it is still present during at least two sampling times, T7.

Table 3 Timing of a push button

Clock Keyboard 2 ms, 2.2 ms	Min.	Тур.	Max.	Unit
Minimum time – push button on	14	24	34	ms
	15.4	26.4	37.4	ms
Minimum time – push button off	24	24	34	ms
-	26.4	26.4	37.4	ms

The entries are debounced on both the leading and trailing edges for 34 ms or 37.4 ms according to the value of the keyboard clock, and so the time remains less than 40 ms. At this time the information can be processed. If the information is still present after more than 40 ms, it is only taken one time.

#### **Serial Bus**

The remote microcontroller is connected to the IC by 4 pins: C2, C3, C4, C5 (see figure 10).

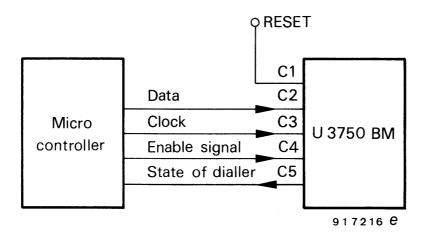


Figure 10 Connection of the microcontroller to the U3750BM

C2 transmits the data, C3 the clock, C4 the enable signal, and C5 indicates the state of the dialer:

C5 = 0, dialer is busy

C5 = 1, dialer is free

Data is serially shifted in a 5-bit register during the positive going transition of the clock pulse. The positive going transition of the enable signal validates the transmission.

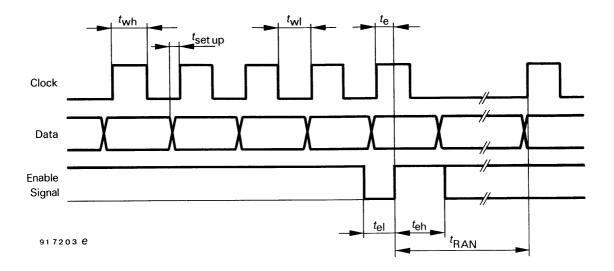


Figure 11 Timing of the serial bus

## **Code Entries Table**

0	0	0	0	0	*
0	0	0	0	1	1
0	0	0	1	0	2
0	0	0	1	1	3
0	0	1	0	0	4
0	0	1	0	1	5
0	0	1	1	0	6
0	0	1	1	1	7
0	1	0	0	0	8
0	1	0	0	1	9
0	1	0	1	0	0
0	1	0	1	1	A
0	1	1	0	0	В
0	1	1	0	1	С
0	1	1	1	0	D
0	1	1	1	1	#
1	0	0	0	0	16
1	0	0	0	1	R flash
1	0	0	1	0	Redial
1	0	1	0	0	Confidence tone
1	0	1	0	1	Micro inhibition
1	0	1	1	0	Pause
1	0	1	1	1	23

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#### **Dialer**

The IC includes a dialing circuit for either pulse dialing or dual tone multifrequency dialing. The dialer transmits the codes decoded by the logic keyboard on the outputs OL and MF.

#### **Mode Selection**

The choice of dialing is made by the tri-state-level on the DC/FV Pin

FV DC = Z pulse dialing in 66/33 ms FV DC

tied to pin RESET pulse dialing in 60/40 ms FV DC = 0 DTMF dialing calibrated

When the circuit is in pulse mode, it is possible to change over to DTMF dialing with the "\*" key. The code "\*" is sent in line. The circuit returns in pulse mode after a reset condition or after a flash pulse (see figure 14).

## **Dialing Codes**

The dialing codes are the numeric keys 0 to 9, and the non numeric keys A, B, C, D, \*, #. All are stored in RAM. The codes A, B, C and D can be only transmitted by the serial bus. In pulse dialing, the code #, B, C and D have no effect

on the dialing. The code A is filtered and corresponds to eleven pulses.

## **Dialing**

As soon as the code is detected by the logic keyboard and written in RAM, it can only be loaded in the dialer if the dialer is not occupied and a pause is not generated.

## **Pulse Dialing**

The output which provides control signals for proper timing in pulse dialing is pin, OL. The dialling starts with a make time (see figure 12).

## **Dual tone Multifrequency Dialing**

The output pin, MF, provides the multifrequency signal to transmit in line. This signal results from the sum of two frequency pulses modulated and requires a filter to compose a dual sine wave. The frequencies are chosen in a low group and a high group. Table 3 shows the frequency tolerance of the output tones for DTMF signalling. In manual dialing or in redial, output tone is timed with a fixed duration.

Table 4 Frequency tolerance of the output tones for DTMF signaling

Standard Frequency	Tone Output Frequency	Frequency Deviation		
Hz	Hz	%	Hz	
Low Group				
697	697.8	+0.12	+0.85	
770	768.6	-0.18	-1.42	
852	848.9	-0.37	-3.12	
941	940.1	-0.10	-0.92	
High Group				
1209	1210.1	+0.09	+1.11	
1336	1338.2	+0.17	+2.23	
1477	1477.3	+0.02	+0.27	
1633	1636.7	+0.22	+3.69	

Tone output frequency when using a 455 kHz ceramic.

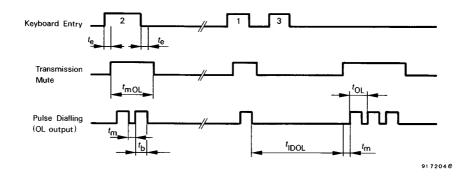


Figure 12 Timing diagram for pulse dialing

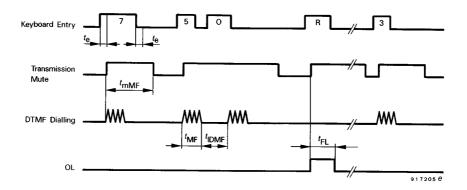


Figure 13 Timing diagram for DTMF dialing

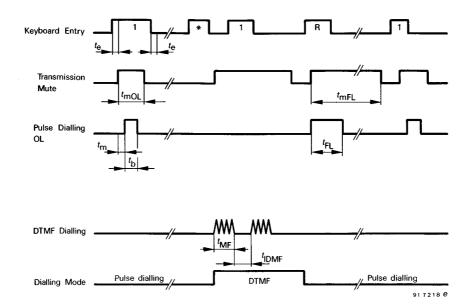


Figure 14 Timing diagram for mixed mode dialing

#### Flash Control

Detecting a "R" code produces either a short timed line break (< 200 ms) or long timed line break (> 200 ms) at the OL output.

For the duration of the flash, it is not possible to take information from the keyboard.

Flash signifies that the circuit executes a particular work as a dialing, a redial, or a pause function, and the code "R" is lost and not used.

The flash pulse resets the read address counter and does not erase the data storage, so later redial is possible.

The flash duration is programmed by the FLASH pin (Pin 3) and depends on the selection of the tri-state-level pin (Pin 4).

Mutes (transmission mute and dialing mute) become active high from the beginning of the line break. Timings explains this.

According to these timings and what has previously been said, a second pulse flash could only follow the first one 810 ms or 850 ms later.

Consequently the "R" entry remains inhibited during a time less than 1 second.

#### **Pause Function**

A pause separates the dial sequence. It is used for waiting for a dial tone.

A pause code takes one position in the RAM like a digit. However, if the circuit executes a pause and if another pause code is entered, the storage of the second one does not occur. Furthermore, the pause running is aborted.

Duration of the pause is given in electrical characteristics for the following configuration: digit, pause, digit, and consequently takes into account the interdigit.

#### **Particular Functions**

After the reset, the particular functions are cleared. The state of the circuit is no confidence tone, no microphone inhibition.

## **Confidence Tone Output**

When the data entries are derived from the serial bus, a pulse frequency modulation corresponding to a 440 Hz sine wave can be generated on the output MF by transmitting the confidence tone code which is 20 (in decimal). The function confidence tone is a flip-flop function.

## **Microphone Inhibition**

Like the confidence tone, it is a flip flop function activated through the serial bus by the code 21 (in decimal).

#### **RAM**

#### **Organization**

The RAM is 32 words of 5 bits and is organized in two parts: one for the data storage and the other for the working RAM.

#### Safeguard

The safeguard is guaranteed by an external capacitor. If  $V_{RAM}$  decreases under the data retention supply voltage, the redial function is forbidden. After the reset of the circuit, a test is executed on  $V_{RAM}$  in order to ensure the redial validity.

#### Data storage

Storage, overflow and erasing are realized through three address counters. The written address counter (P1) points out the location where the code will be stored. At each storage, P1 is incremented by one. As each code is recalled from the RAM for line dialing, the read address counter (M1) is incremented by one to select the RAM location of the next code to be recalled.

Consequently, the difference between the contents of P1 and of M1 represents the number of codes that have been written into the RAM but not yet converted into line dialing.

The third counter (P2) gives the real capacity of the redial register.

#### **Redial features**

#### Capacity

If more than 23 codes are entered into the RAM memory, overflow results and the excess codes replace the data in the lower numbered RAM locations. In this event, automatic redial is no longer possible.

#### Storage

Storage pertains to the dialing codes 0 to 9, \*, #, pause, A, B, C and D. It is independent of the dialing mode (pulse dialing or DTMF dialing). The storage generally contains the last digits transmitted.

#### Use of redial

The use of redial is always possible except if the content of the RAM is empty (P2=0). This happens when the RAM supply is not high enough , when an overflow occurred, or when previously an erroneous use of the redial occurred (start of manual dialing not equal to the content of the RAM).

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If the redial is ordered after a manual dialing the redial is executed according to the digits already transmitted in line.

The redial is effective if the comparison digit-by-digit of all digits is correct. It is not produced if the comparison is incorrect or if the number of digits exceeded the last capacity of the redial. During redial, entry codes are accepted.

## **Procedure with \***

Storage does not pertain to the "\*" code and the codes entered after it when the manual dialing starts with a numeric code or A, B, C, D, #, or pause codes. If the dialing starts with the "\*" code, all codes can be stored.

Note: the "#" code is treated for storage like a number.

#### Procedure with flash

The flash pulse does not reset the content of the RAM.

Example: 1 2 3 R Redial  $\$   $\Rightarrow$  1 2 3 R 1 2 3 Redial  $\$   $\Rightarrow$  1 2 3

## **Erasing redial**

The erasing is possible through the serial bus with the decimal codes 16 and 23. There is not much difference

between these two codes: code 16 always erases the redial, code 23 inhibits a later redial if it is transmitted after or before dialing codes. The following examples illustrates this.

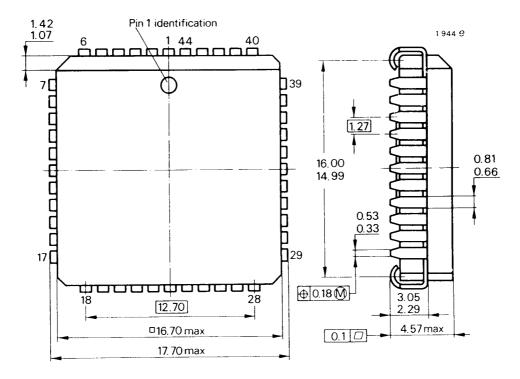
Transmi	On Line		
1 2 3 Redial \$	16		1 2 3 –nothing
1 2 3 Redial \$	23		123
1 2 3 Redial \$	16	4 \$	1234
1 2 3 Redial \$	23	4 \$	1234
1 2 3 Redial \$	R	16\$	1 2 3 R -
1 2 3 Redial \$	R	23\$	1 2 3 R 1 2 3

## **Special Case**

When the overflow flag is set and when the written address counter becomes equal to the read address counter, the codes are not stored.

#### **Dimensions in mm**

Package: PLCC 44



## **U3750BM**

**TELEFUNKEN Semiconductors** 

## **Ozone Depleting Substances Policy Statement**

It is the policy of TEMIC TELEFUNKEN microelectronic GmbH to

- 1. Meet all present and future national and international statutory requirements.
- Regularly and continuously improve the performance of our products, processes, distribution and operating systems with respect to their impact on the health and safety of our employees and the public, as well as their impact on the environment.

It is particular concern to control or eliminate releases of those substances into the atmosphere which are known as ozone depleting substances (ODSs).

The Montreal Protocol (1987) and its London Amendments (1990) intend to severely restrict the use of ODSs and forbid their use within the next ten years. Various national and international initiatives are pressing for an earlier ban on these substances.

**TEMIC TELEFUNKEN microelectronic GmbH** semiconductor division has been able to use its policy of continuous improvements to eliminate the use of ODSs listed in the following documents.

- 1. Annex A, B and list of transitional substances of the Montreal Protocol and the London Amendments respectively
- 2. Class I and II ozone depleting substances in the Clean Air Act Amendments of 1990 by the Environmental Protection Agency (EPA) in the USA
- 3. Council Decision 88/540/EEC and 91/690/EEC Annex A, B and C (transitional substances) respectively.

**TEMIC** can certify that our semiconductors are not manufactured with ozone depleting substances and do not contain such substances.

We reserve the right to make changes to improve technical design and may do so without further notice.

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