µA108A · µA208A · µA308A
µA108 · µA208 · µA308
SUPER BETA OPERATIONAL AMPLIFIERS
FAIRCHILD LINEAR INTEGRATED CIRCUITS

GENERAL DESCRIPTION — The µA108 Super Beta Operational Amplifier series is constructed using the Fairchild Planar* epitaxial process. High input impedance, low noise, low input offsets, and temperature drift are made possible through use of super beta processing, making the device suitable for applications requiring high accuracy and low drift performance. The µA108A series is specially selected for extremely low offset voltage and drift, and high common mode rejection, giving superior performance in applications where offset nulling is undesirable. Increased slew rate without performance compromise is available through use of feedforward compensation techniques, maximizing performance in high speed sample-and-hold circuits and precision high speed summing amplifiers. The wide supply range and excellent supply voltage rejection assure maximum flexibility in voltage follower, summing, and general feedback applications.

- **GUARANTEED LOW INPUT OFFSET CHARACTERISTICS**
- **HIGH INPUT IMPEDANCE**
- **LOW OFFSET CURRENT**
- **LOW BIAS CURRENT**
- **OPERATION OVER WIDE SUPPLY RANGE**

ABSOLUTE MAXIMUM RATINGS

Supply Voltage	
µA108A, µA108, µA208A, µA208	±20 V
µA308A, µA308	±18 V
Internal Power Dissipation (Note 1)	
Metal Can	500 mW
DIP	600 mW
Flatpak	570 mW
Mini DIP	310 mW
Differential Input Current (Note 2)	±10 mA
Input Voltage (Note 3)	±15 V
Storage Temperature Range	-65°C to +150°C
Operating Temperature Range	
Military (µA108A, µA108)	-55°C to +125°C
Industrial (µA208A, µA208)	-25°C to +85°C
Commercial (µA308A, µA308)	0°C to +70°C
Pin Temperature (Soldering, 60 s)	300°C
Output Short-Circuit Duration (Note 4)	Indefinite

CONNECTION DIAGRAMS
**8-PIN METAL CAN**
(TOP VIEW)
PACKAGE OUTLINE 5S
PACKAGE CODE H
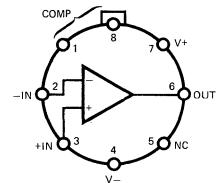
ORDER INFORMATION

TYPE	PART NO.
µA108A	µA108AHM
µA108	µA108HM
µA208A	µA208AHM
µA208	µA208HM
µA308A	µA308AHC
µA308	µA308HC

CONNECTION DIAGRAMS
8-PIN MINI DIP
(TOP VIEW)
PACKAGE OUTLINE 9T
PACKAGE CODE T
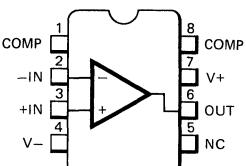
ORDER INFORMATION

TYPE	PART NO.
µA308A	µA308ATC
µA308	µA308TC

14-PIN DIP *
(TOP VIEW)
PACKAGE OUTLINE 6A
PACKAGE CODE D

ORDER INFORMATION

TYPE	PART NO.
µA108A	µA108ADM
µA108	µA108DM
µA208A	µA208ADM
µA208	µ208DM
µA308A	µA308ADC
µA308	µA308DC

10-PIN FLATPAK*
(TOP VIEW)
PACKAGE OUTLINE 3F
PACKAGE CODE F

ORDER INFORMATION

TYPE	PART NO.
µA108A	µA108AFM
µA108	µA108FM
µA208A	µA208AFM
µA208	µA208FM

** Available on special order

See notes on following pages.

*Planar is a patented Fairchild process.

EQUIVALENT CIRCUIT
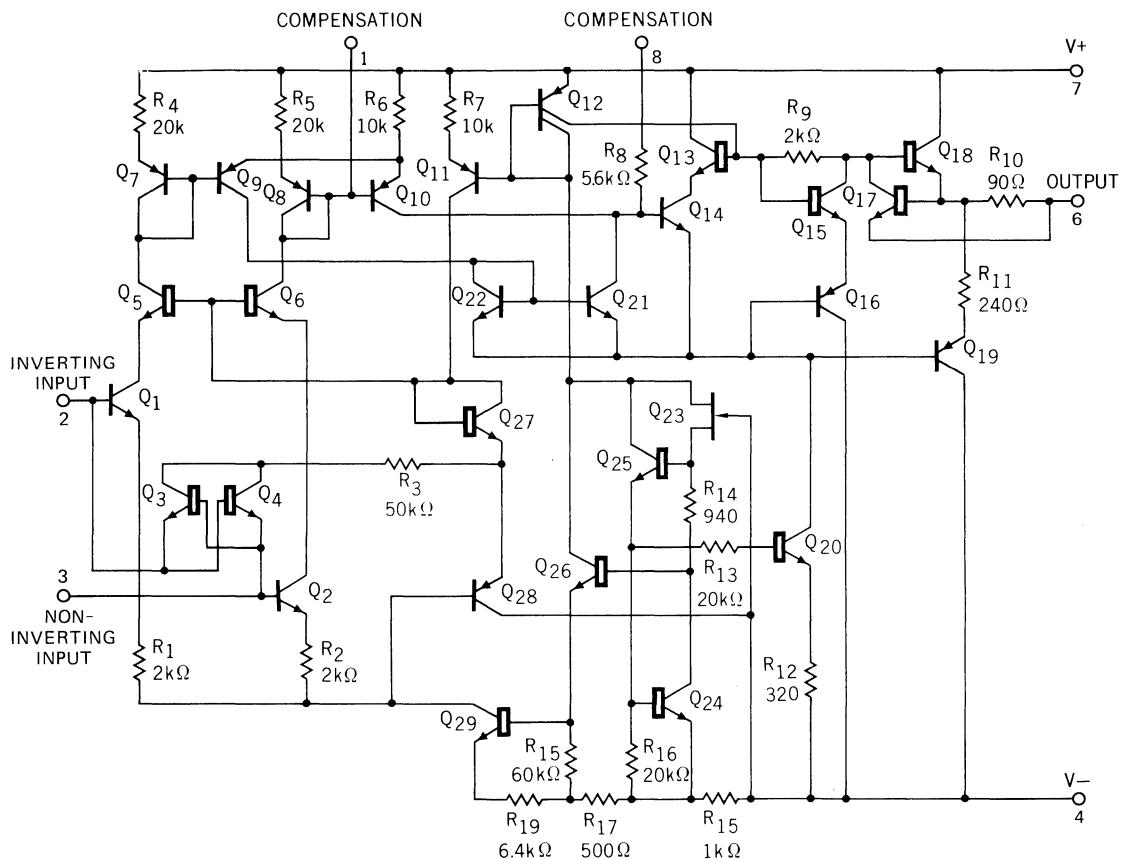
Pin numbers are for metal can only

μA108A and μA208A

ELECTRICAL CHARACTERISTICS: ±5.0 V ≤ V$_S$ ≤ ±20 V, T$_A$ = 25°C, C$_C$ = 30 pF unless otherwise specified.

CHARACTERISTICS	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage			0.3	0.5	mV
Input Offset Current			0.05	0.2	nA
Input Bias Current			0.8	2.0	nA
Input Resistance		30	70		MΩ
Supply Current	V$_S$ = ±15 V		0.3	0.6	mA
Large Signal Voltage Gain	V$_S$ = ±15 V, R$_L$ ⩾ 10 kΩ, V$_{OUT}$ = ±10 V	80,000	300,000		V/V

The following specifications apply for −55°C ≤ T$_A$ ≤ +125°C (Note 5)

Input Offset Voltage				1.0	mV
Average Input Offset Voltage Drift			1.0	5.0	μV/°C
Input Offset Current				0.4	nA
Average Input Offset Current Drift			0.5	2.5	pA/°C
Input Bias Current			0.8	3.0	nA
Supply Current	T$_A$ = +125°C		0.15	0.4	mA
Input Voltage Range	V$_S$ = ±15 V	±13.5			V
Common Mode Rejection Ratio		96	110		dB
Supply Voltage Rejection Ratio		96	110		dB
Large Signal Voltage Gain	V$_S$ = ±15 V, R$_L$ ⩾ 10 kΩ V$_{OUT}$ = ±10 V	40,000			V/V
Output Voltage Swing	V$_S$ = ±15 V, R$_L$ ⩾ 10 kΩ	±13	±14		V

μA308A

ELECTRICAL CHARACTERISTICS: ±5.0 V ≤ V$_S$ ≤ ±15 V, T$_A$ = 25°C, C$_C$ = 30 pF unless otherwise specified.

CHARACTERISTICS	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage			0.3	0.5	mV
Input Offset Current			0.2	1.0	nA
Input Bias Current			1.5	7.0	nA
Input Resistance		10	40		MΩ
Supply Current	V$_S$ = ±15 V		0.3	0.8	mA
Large Signal Voltage Gain	V$_S$ = ±15 V, R$_L$ ⩾ 10 kΩ, V$_{OUT}$ = ±10 V	80,000	300,000		V/V

The following specifications apply for 0°C ≤ T$_A$ ≤ +70°C

Input Offset Voltage				0.73	mV
Average Input Offset Voltage Drift			1.0	5.0	μV/°C
Input Offset Current				1.5	nA
Average Input Offset Current Drift			2.0	10	pA/°C
Input Bias Current				10	nA
Input Voltage Range	V$_S$ = ±15 V	±13.5			V
Common Mode Rejection Ratio		96	110		dB
Supply Voltage Rejection Ratio		96	110		dB
Large Signal Voltage Gain	V$_S$ = ±15 V, R$_L$ ⩾ 10 kΩ, V$_{OUT}$ = ±10 V	60,000			V/V
Output Voltage Swing	V$_S$ = ±15 V, R$_L$ ⩾ 10 kΩ	±13	±14		V

1. Rating applies to ambient temperatures up to 70°C. Above 70°C ambient derate linearly at 6.3 mW/°C for metal can, 8.3 mW/°C for the DIP, 5.6 mW/°C for the mini DIP and 7.1 mW/°C for the flatpak.
2. The inputs are shunted with back-to-back diodes for overvoltage protection. Therefore, excessive current will flow if a differential input voltage in excess of 1 V is applied between the inputs unless adequate limiting resistance is used.
3. For supply voltages less than ±15 V, the absolute maximum input voltage is equal to the supply voltage.
4. Short circuit may be to either supply or ground. Rating applies to operation up to the maximum operating temperature range.
5. For the μA208A/208, all temperature specifications apply over −25°C ≤ T$_A$ ≤ 85°C.

μA108 and μA208

ELECTRICAL CHARACTERISTICS: $\pm 5.0\,V \leqslant V_S \leqslant \pm 20\,V$, $T_A = 25°C$, $C_C = 30\,pF$ unless otherwise specified.

CHARACTERISTICS	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage			0.7	2.0	mV
Input Offset Current			0.05	0.2	nA
Input Bias Current			0.8	2.0	nA
Input Resistance		30	70		MΩ
Supply Current	$V_S = \pm 15\,V$		0.3	0.6	mA
Large Signal Voltage Gain	$R_L \geqslant 10\,k\Omega$, $V_{OUT} = \pm 10\,V$ $V_S = \pm 15\,V$	50,000	300,000		V/V

The following specifications apply for $-55°C \leqslant T_A \leqslant 125°C$ (Note 5)

Input Offset Voltage				3.0	mV
Average Input Offset Voltage Drift			3.0	15	μV/°C
Input Offset Current				0.4	nA
Average Input Offset Current Drift			0.5	2.5	pA/°C
Input Bias Current				3.0	nA
Supply Current	$T_A = +125°C$		0.15	0.4	mA
Input Voltage Range	$V_S = \pm 15\,V$	±13.5			V
Common Mode Rejection Ratio		85	100		dB
Supply Voltage Rejection Ratio		80	96		dB
Large Signal Voltage Gain	$V_S = \pm 15\,V$, $R_L \geqslant 10\,k\Omega$, $V_{OUT} = \pm 10\,V$	25,000			V/V
Output Voltage Swing	$V_S = \pm 15\,V$, $R_L = 10\,k\Omega$	±13	±14		V

μA308

ELECTRICAL CHARACTERISTICS: $\pm 5.0\,V \leqslant V_S \leqslant \pm 15\,V$, $T_A = 25°C$, $C_C = 30\,pF$ unless otherwise specified.

CHARACTERISTICS	CONDITIONS	MIN	TYP	MAX	UNITS
Input Offset Voltage			2.0	7.5	mV
Input Offset Current			0.2	1.0	nA
Input Bias Current			1.5	7.0	nA
Input Resistance		10	40		MΩ
Supply Current	$V_S = \pm 15\,V$		0.3	0.8	mA
Large Signal Voltage Gain	$V_S = \pm 15\,V$, $R_L \geqslant 10\,k\Omega$, $V_{OUT} = \pm 10\,V$	25,000	300,000		V/V

The following specifications apply for $0°C \leqslant T_A \leqslant +70°C$

Input Offset Voltage				10	mV
Average Input Offset Voltage Drift			6.0	30	μV/°C
Input Offset Current				1.5	nA
Average Input Offset Current Drift			2.0	10	pA/°C
Input Bias Current				10	nA
Input Voltage Range	$V_S = \pm 15\,V$	±13.5			V
Common Mode Rejection Ratio		80	100		dB
Supply Voltage Rejection Ratio		80	96		dB
Large Signal Voltage Gain	$V_S = \pm 15\,V$, $R_L \geqslant 10\,k\Omega$, $V_{OUT} = \pm 10\,V$	15,000			V/V
Output Voltage Swing	$V_S = \pm 15\,V$, $R_L = 10\,k\Omega$	±13	±14		V

5

TYPICAL PERFORMANCE CURVES FOR μA108 SERIES

INPUT NOISE VOLTAGE AS A FUNCTION OF FREQUENCY
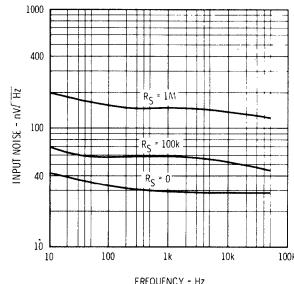
POWER SUPPLY REJECTION AS A FUNCTION OF FREQUENCY
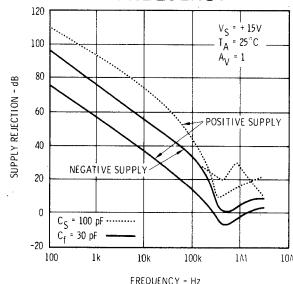
OPEN LOOP FREQUENCY RESPONSE
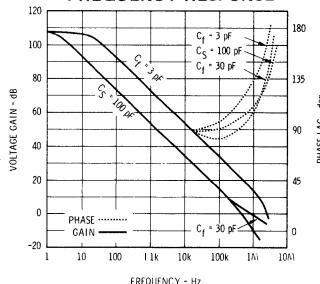
CLOSED LOOP OUTPUT IMPEDANCE
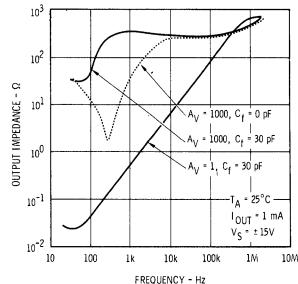
LARGE SIGNAL FREQUENCY RESPONSE
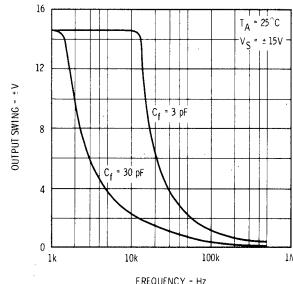
VOLTAGE FOLLOWER PULSE RESPONSE
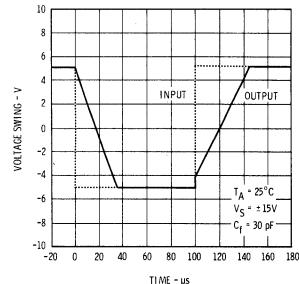
TYPICAL PERFORMANCE CURVES FOR μA108A • μA208A • μA108 • μA208 (Unless Otherwise Specified)

INPUT CURRENTS AS A FUNCTION OF AMBIENT TEMPERATURE
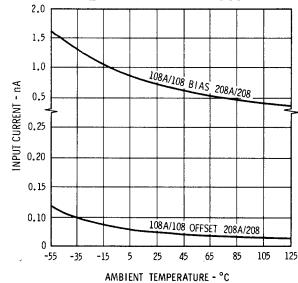
MAXIMUM OFFSET ERROR
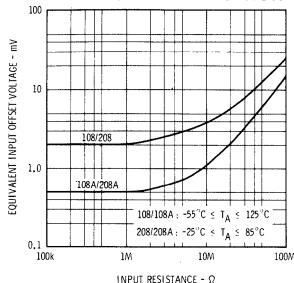
MAXIMUM DRIFT ERROR
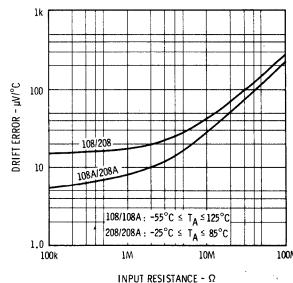
VOLTAGE GAIN AS A FUNCTION OF SUPPLY VOLTAGE μA108
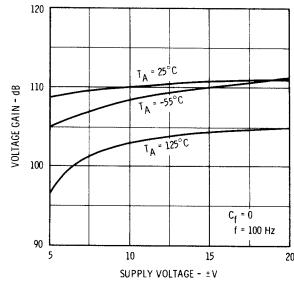
OUTPUT SWING AS A FUNCTION OF OUTPUT CURRENT μA108
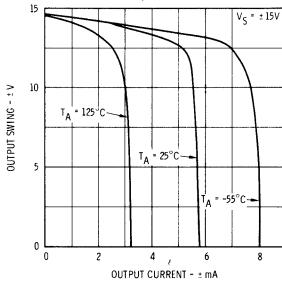
SUPPLY CURRENT AS A FUNCTION OF SUPPLY VOLTAGE μA108
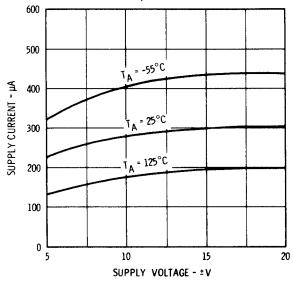
TYPICAL PERFORMANCE CURVES FOR μA308A AND μA308 (Unless Otherwise Specified)

**INPUT CURRENT AS A FUNCTION OF AMBIENT TEMPERATURE**
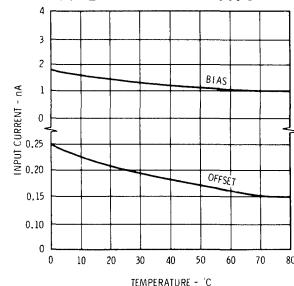
MAXIMUM OFFSET ERROR μA308
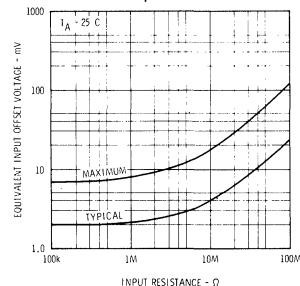
MAXIMUM DRIFT ERROR μA308
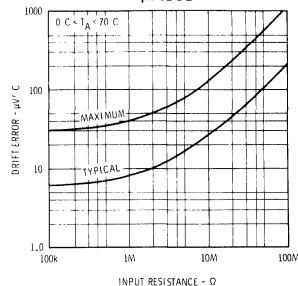
**VOLTAGE GAIN AS A FUNCTION OF SUPPLY VOLTAGE**
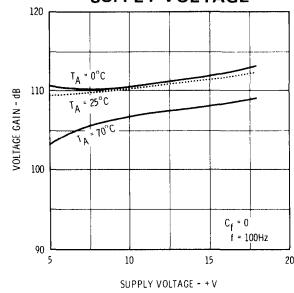
OUTPUT SWING AS A FUNCTION OF OUTPUT CURRENT
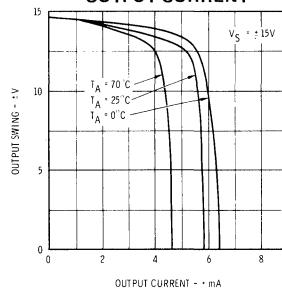
SUPPLY CURRENT AS A FUNCTION OF SUPPLY VOLTAGE
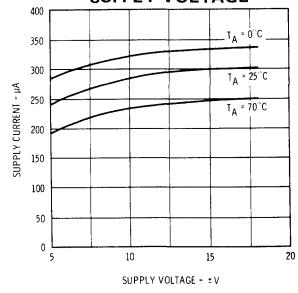
STANDARD COMPENSATION CIRCUITS

$$C_f \geq 30 \left(\frac{1}{1 + \dfrac{R_2}{R_1}} \right)$$

FEEDFORWARD COMPENSATION
### HIGHER SLEW RATES AND WIDER BANDWIDTH
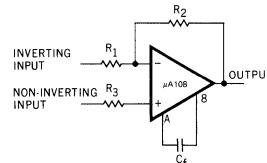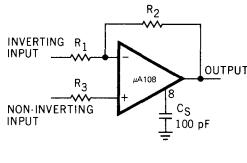
STANDARD FEEDFORWARD
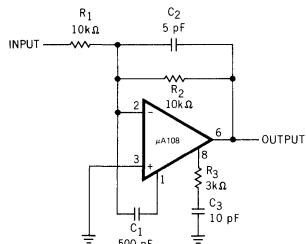
OPEN LOOP VOLTAGE GAIN
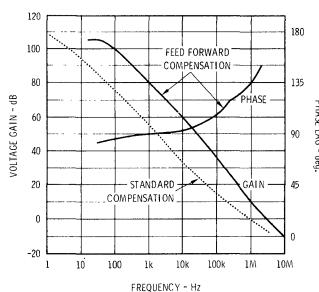
FEEDFORWARD COMPENSATION FOR DECOUPLING LOAD CAPACITANCE
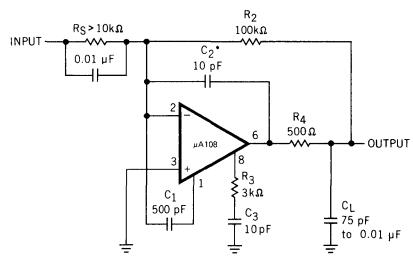
$$^*C_2 > \frac{5 \times 10^5}{R_2} \, pF$$

GUARDING

Extra care must be taken in the assembly of printed circuit boards to take full advantage of the low input currents of the μA108 amplifier. Boards must be thoroughly cleaned with TCE or alcohol and blown dry with compressed air. After cleaning, the boards should be coated with epoxy or silicone rubber to prevent contamination.

Even with properly cleaned and coated boards, leakage currents may cause trouble at 125°C, particularly since the input pins are adjacent to pins that are at supply potentials. This leakage can be significantly reduced by using guarding to lower the voltage difference between the inputs and adjacent metal runs. Input guarding of the 8-pin TO-99 package is accomplished by using a 10-pin pin circle, with the leads of the device formed so that the holes adjacent to the inputs are empty when it is inserted in the board. The guard, which is a conductive ring surrounding the inputs, is connected to a low impedance point that is at approximately the same voltage as the inputs. Leakage currents from high voltage pins are then absorbed by the guard.

The pin configuration of the dual in-line package is designed to facilitate guarding, since the pins adjacent to the inputs are not used (this is different from the standard μA741 and μA101A pin configuration).

CONNECTION OF INPUT GUARDS

**INVERTING AMPLIFIER**　　　　**FOLLOWER**　　　**NON-INVERTING AMPLIFIER**　　**BOARD LAYOUT FOR INPUT GUARDING WITH TO-99 PACKAGE (BOTTOM VIEW)**
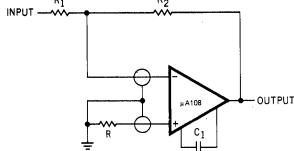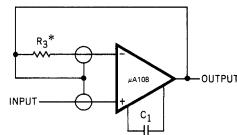
$$R = \frac{R_1 R_2}{R_1 + R_2}$$

*Use to compensate for large source resistances.
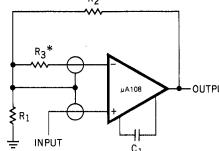
$$NOTE: \frac{R_1 R_2}{R_1 + R_2} \quad \text{Must be low impedance}$$
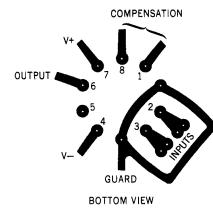
TYPICAL APPLICATIONS

**FAST† SUMMING AMPLIFIER WITH LOW INPUT CURRENT**　　　　**SAMPLE AND HOLD**
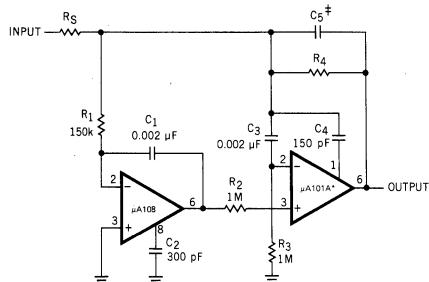
*In addition to increasing speed, the μA101A raises high and low frequency gain, increases output drive capability and eliminates thermal feedback.

† Power Bandwidth: 250 kHz
Small Signal Bandwidth: 3.5 MHz
Slew Rate: 10 V/μs

$$‡ C_5 = \frac{6 \times 10^{-8}}{R_1}$$
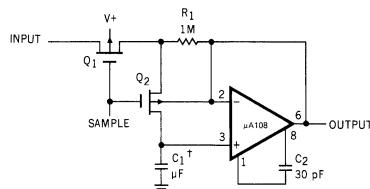
* Worst case drift less than 2.5 mV/s
† Teflon, Polyethylene or Polycarbonate Dielectric Capacitor