

DATASHEET

UF4C120070B7S



Part Number	Package	Marking
UF4C120070B7S	D ² PAK-7L	UF4C120070B7S



Silicon Carbide (SiC) Cascode JFET -EliteSiC, Power N-Channel, D2PAK-7L, 1200 V, 72 mohm

Rev B, January 2025

Description

The UF4C120070B7S is a 1200V, $72m\Omega$ G4 SiC FET. It is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows use of off-the-shelf gate drivers hence requiring minimal re-design when replacing Si IGBTs, Si superjunction devices or SiC MOSFETs. Available in the space-saving D²PAK-7L package which enables automated assembly, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

Features

- On-resistance R_{DS(on)}: 72mΩ (typ)
- Operating temperature: 175°C (max)
- Excellent reverse recovery: Q_{rr} = 101nC
- Low body diode V_{FSD}: 1.43V
- Low gate charge: Q_G = 37.8nC
- Threshold voltage V_{G(th)}: 4.8V (typ) allowing 0 to 15V drive
- Low intrinsic capacitance
- ESD protected: HBM class 2 and CDM class C3
- D²PAK-7L package for faster switching, clean gate waveforms

Typical applications

- EV charging
- PV inverters
- Switch mode power supplies
- Power factor correction modules
- Induction heating





Maximum Ratings

Parameter	Symbol	Test Conditions	Value	Units
Drain-source voltage	V _{DS}		1200	V
Cato source voltage	V	DC	-20 to +20	V
Gale-source voltage	V GS	AC (f > 1Hz)	-25 to +25	V
Continuous drain surrant ¹	I	T _C = 25°C	25.7	А
Continuous drain current	D	T _C = 100°C	19.2	А
Pulsed drain current ²	I _{DM}	T _c = 25°C	76	А
Single pulsed avalanche energy ³	E _{AS}	L=15mH, I _{AS} =2.2A	36	mJ
SiC FET dv/dt ruggedness	dv/dt	$V_{DS} \le 800V$	200	V/ns
Power dissipation	P _{tot}	T _C = 25°C	183	W
Maximum junction temperature	T _{J,max}		175	°C
Operating and storage temperature	TJ, T _{STG}		-55 to 175	°C
Reflow soldering Temperature	T _{solder}	reflow MSL 1	245	°C

1. Limited by $T_{J,max}$

2. Pulse width t_p limited by $T_{J,max}$

3. Starting $T_J = 25^{\circ}C$

Thermal Characteristics

Parameter	Symbol	Test Conditions		Units		
Parameter	Symbol		Min	Тур	Max	Units
Thermal resistance, junction-to-case	$R_{ ext{ hetaJC}}$			0.63	0.82	°C/W

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Electrical Characteristics (T_J = +25°C unless otherwise specified) Typical Performance - Static

Demonstern	Gundhal	Test Canditiana	Value			Linite
Parameter	Symbol	Test Conditions	Min	Тур	Max	Units
Drain-source breakdown voltage	BV _{DS}	V _{GS} =0V, I _D =1mA	1200			V
		V _{DS} =1200V,		0.4	18	
Total drain lookage current		V _{GS} =0V, T _J =25°C		0.4	10	
Total drain leakage current	DSS	V _{DS} =1200V,		10		μΑ
		V _{GS} =0V, T _J =175°C	10			
Table about the last of the second	I _{GSS}	V _{DS} =0V, T _J =25°C,		1	20	μΑ
Total gate leakage current		V _{GS} =-20V / +20V		0	20	
		V_{GS} =12V, I _D =20A,		70	01	
		TJ=22°C		12	71	
Drain-source on-registance	D	V_{GS} =12V, I _D =20A,	140			mΩ
	NDS(on)	T _J =125°C				
		V_{GS} =12V, I_{D} =20A,		107		
		T_=175°C		177		
Gate threshold voltage	V _{G(th)}	V _{DS} =5V, I _D =10mA	4	4.8	6	V
Gate resistance	R _G	f=1MHz, open drain		4.5		Ω

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Typical Performance - Reverse Diode

Darameter	Symbol	Test Conditions		Lipite			
Palameter	Symbol	Test Conditions	Min	Тур	Max	Units	
Diode continuous forward current ¹	I _S	T _c =25°C			25.7	А	
Diode pulse current ²	I _{S,pulse}	T _C =25°C			76	А	
Forward valtage	N/	V _{GS} =0V, I _S =10A, T _J =25°C		1.43	1.64	M	
Forward voltage	V _{FSD}	V _{GS} =0V, I _S =10A, T _J =175°C	2.38		v		
Reverse recovery charge	Q _{rr}	V_{DS} =800V, I _S =20A, V_{GS} =-5V, R _G =20 Ω ,		101		nC	
Reverse recovery time	t _{rr}	di/dt=1800A/µs, T _J =25°C		11		ns	
Reverse recovery charge	Q _{rr}	V _{DS} =800V, I _S =20A, V _{GS} =-5V, R _G =20Ω,		116		nC	
Reverse recovery time	t _{rr}	di/dt=1800A/µs, TJ=150°C		11		ns	





Typical Performance - Dynamic

Deveneter	Cumphiel	Test Conditions	Value			Lipite	
Parameter	Symbol	lest Conditions	Min	Тур	Max	Units	
Input capacitance	C _{iss}			1370			
Output capacitance	C _{oss}	f=100kHz		35		pF	
Reverse transfer capacitance	C _{rss}	1-100012		2			
Effective output capacitance, energy	C	V _{DS} =0V to 800V,		42		nF	
related	Coss(er)	V _{GS} =0V		72		p	
Effective output capacitance, time	C	V_{DS} =0V to 800V,		71		ьE	
related	C _{oss} (tr)	V _{GS} =0V		/1		рг	
C _{OSS} stored energy	E _{oss}	V _{DS} =800V, V _{GS} =0V		13.4		μJ	
Total gate charge	Q_{G}	V800V_I_=20A		37.8		_	
Gate-drain charge	Q_{GD}	$V_{DS} = 000 V$, $T_D = 20 A$,		9.5		nC	
Gate-source charge	Q _{GS}	V _{GS} =0V to 15V		10			
Turn-on delay time	t _{d(on)}			20			
Rise time	t _r	V_{DS} =800V. I_{D} =20A. Gate		32		nc	
Turn-off delay time	$t_{d(off)}$	$\begin{array}{c} \text{Driver=-5V to +15V,} \\ \text{R}_{\text{G}_{\text{ON}}}=10\Omega, \text{R}_{\text{G}_{\text{OFF}}}=50\Omega, \end{array}$		57		115	
Fall time	t _f			12			
Turn-on energy including R _s energy	E _{ON}	FWD: same device with		352			
Turn-off energy including R _s energy	E _{OFF}	V_{GS} =-5V and R_{G_OFF} =50 Ω ,		62			
Total switching energy	E _{TOTAL}	Device Snubber:		414		μJ	
Snubber R _s energy during turn-on	E _{RS_ON}	Cs=100pF,Rs=20Ω T.=25°C		10			
Snubber R_s energy during turn-off	E_{RS_OFF}	.,		8			
Turn-on delay time	t _{d(on)}	Notes 4 and 5.		24			
Rise time	t _r	V _{DS} =800V, I _D =20A, Gate		33		ns	
Turn-off delay time	t _{d(off)}	Driver=-5V to +15V,		63		115	
Fall time	t _f	R _{G_ON} =10Ω, R _{G_OFF} =50Ω, inductive Load, FWD: same device with		13			
Turn-on energy including R _S energy	E _{ON}			396			
Turn-off energy including R_S energy	E _{OFF}	$V_{GS}\text{=-}5\text{V}$ and $R_{G_OFF}\text{=-}50\Omega\text{,}$		81			
Total switching energy	E _{TOTAL}	Device Snubber:		477		μJ	
Snubber R_s energy during turn-on	E _{RS_ON}	Cs=100pF,Rs=20Ω T.=150°C		5			
Snubber R_{S} energy during turn-off	E_{RS_OFF}	1 100 0		12			

4. Measured with the half-bridge mode switching test circuit in Figure 26.

5. In this datasheet, all the switching energies (turn-on energy, turn-off energy and total energy) presented in the tables and Figures include the device RC snubber energy losses.

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Typical Performance Diagrams



Figure 1. Typical output characteristics at $T_J = -55^{\circ}C$, tp < 250 μ s



Figure 3. Typical output characteristics at T_J = 175°C, tp < 250 μ s



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Figure 2. Typical output characteristics at $T_J = 25$ °C, tp < 250 μ s



Figure 4. Normalized on-resistance vs. temperature at V_{GS} = 12V and I_{D} = 20A



Figure 5. Typical drain-source on-resistances at V_{GS} = 12V



Figure 7. Threshold voltage vs. junction temperature at V_{DS} = 5V and I_{D} = 10mA



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Figure 6. Typical transfer characteristics at V_{DS} = 5V



Figure 8. Typical gate charge at I_D = 20A and V_{DS} =800V

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Figure 9. 3rd quadrant characteristics at $T_J = -55^{\circ}C$

Figure 10. 3rd quadrant characteristics at $T_J = 25^{\circ}C$



Figure 11. 3rd quadrant characteristics at $T_J = 175^{\circ}C$



Figure 12. Typical stored energy in C_{OSS} at V_{GS} = 0V

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Figure 13. Typical capacitances at f = 100kHz and V_{GS} = 0V





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Figure 14. DC drain current derating



Figure 15. Total power dissipation

Figure 16. Maximum transient thermal impedance







Figure 19. Clamped inductive switching energy vs. Drain Current at V_{DS} = 800V and T_J = 25°C



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Figure 18. Reverse recovery charge Qrr vs. junction temperature at V_{DS} = 800V



Figure 20. RC snubber energy loss vs. Drain Current at V_{DS} = 800V, I_D = 20A, and T_J = 25°C





Figure 21. Clamped inductive switching energies vs. $R_{G,EXT}$ at V_{DS} = 800V, I_D = 20A, and T_J = 25°C



Figure 23. Clamped inductive switching energies vs. snubber capacitance C_S at V_{DS} = 800V, I_D = 20A, and T_J = 25°C



Figure 22. RC snubber energy loss vs. $R_{G,EXT}$ at V_{DS} = 800V, I_D = 20A, and T_J = 25°C



Figure 24. RC snubber energy losses vs. snubber capacitance C_s at V_{DS} = 800V, I_D = 20A, and T_J = 25°C

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Figure 25. Clamped inductive switching energy vs. junction temperature at V_{DS} = 800V and I_D =20A

Figure 26. Schematic of the half-bridge mode switching test circuit with device RC snubbers (R_s =20 Ω , C_s = 100pF) and a bus RC snubber (R_{ss} = 2.5 Ω , C_{ss} =100nF).

Applications Information

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ($R_{DS(on)}$), output capacitance (C_{oss}), gate charge (Q_G), and reverse recovery charge (Qrr) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

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Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see https://www.qorvo.com/design-hub.

A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small $R_{(G)}$ will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the Qorvo website at https://www.qorvo.com/design-hub.





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PACKAGE OUTLINE



7L-D2PAK						
0.04	M	М	IN	СН		
SIM	Min	Max	Min	Max		
A	4.30	4.56	.169	.180		
A1	0.00	0.25	.000	.010		
A2	2.45	2.75	.096	.108		
b	0.50	0.70	.020	.028		
b1	0.50		.020	-		
с	0.40	0.60	.016	.024		
c1	0.40		.016			
c2	1.20	1.40	.047	.055		
D	8.93	9.23	.352	.363		
D1	4.65	4.95	.183	.195		
D2	7.90	8.10	.311	.319		
e	1.27	BSC	.050	BSC		
E	10.08	10.28	.397	.405		
E1	6.82	7.62	.269	.300		
E2	6.50	8.60	.256	.339		
E3	3.50	3.70	.138	.146		
н	15.00	16.00	.591	.630		
H1	6.68	6.88	.263	.271		
H3	7.3	REF.	.287	REF		
L	1.90	2.50	.075	.098		
L1	0.98	1.42	.039	.056		
L3	0.25	BSC	.0098	3 BSC		
L4	5.22	REF	.205	REF		
ØP1	0.65	0.85	.026	.033		
ØP2	1.40	1.60	.055	.063		

Notes:

- 1. GENERAL TOLERANCE: ±0.1 unless otherwise specified
- 2. CONTROLLING DIMENSION: MILLIMETERS
- 3. PACKAGE BODY SIDES EXCLUDE MOLD FLASH AND GATE BURRS.
- 4. DIMENSION L IS MEASURED IN GAUGE LINE.
- 5. CONVERTED INCH DIMENSIONS ARE NOT NECESSARILY EXACT.
- 6. DIMENSION c1 AND b1 APPLIES TO BASE METAL ONLY



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PART MARKING





PACKING TYPE

Carrier Tape



UNIT: MM

01111												
PACKAGE	AO	BO	KO	DO	D1	E	E1	E2	P0	P1	P2	Т
D2PAK (24 mm)	10.80 ±0.10	16.30 ±0.10	4.70 ±0.10	1.50 ±0.10	1.50 +0.1 -0	24.00 ±0.30	1.75 ±0.10	11.50 ±0.10	16.00 ±0.10	4.00 ±0.10	2.00 ±0.10	0.35 ±0.10

Exte	erior	size	
	W1	16.9±0.1	
Spec 1	W2	1.3±0.1	
	W3	1.0±0.1	
~	W1	17.2±0.1	0
Spec 2	W2	1.8±0.1	Ф
2	W3	0.85±0.1	\bigcirc

<u>Reel</u>



All dimensions in millimeters Anti-Static Tape and Rell (T&R) Quantity per Reel: 800 units



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REVISION HISTORY

Revision	Create Date (mm/dd/yyyy)	Description of Change	Initiator of Change
С	11/06/2023	Updated to Qorvo template Updated Package outline drawing based latest drawing revision	Glenn Galang
D	05/21/2024	Added illustration of device orientation on carrier tape (page 3)	Glenn Galang

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