

High-Efficiency Boost DC/DC Controller

General Description

The VP3478 is a versatile controller designed for use in boost power converter and topologies that needs an external low-side N-MOSFET acting as primary switch. Besides cycle-by-cycle current limiting, current mode control scheme also makes it wide bandwidth and good transient response. The current limit can be programmed simply with an external resistor.

The switching frequency can be set in any value between 100kHz and 1MHz with a resistor or any external clock source. The VP3478 can be operated at high switching frequency to save the solution board size. While entering shutdown mode, the VP3478 only sinks 10µA and it allows power supply sequencing. It has built-in protection circuits such as thermal shutdown, short circuit protection, and overvoltage protection. Internal softstart circuitry reduces the inrush current at start- Applications up.

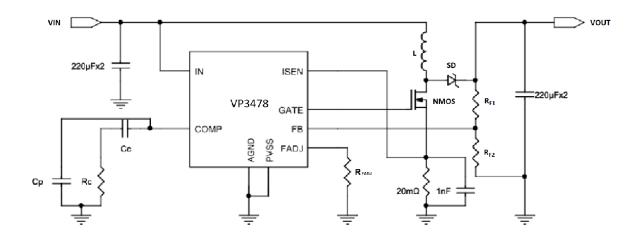
VP3478 is available in small MSOP-8 green package.

Features

- VP3478MSG8-A/VP3478SPG8-A in MSOP8 package is AEC-Q100 Qualified and Manufactured on an Automotive Grade Flow
- Wide Input Voltage from 2.97V to 40V
- Reference Voltage with ±2.0% Accuracy
- Adjustable 100kHz~1MHz Clock Frequency
- 0.8V FB Version: VP3478MSG8
- 1.26V FB Version : VP3478SPG8
- 10µA Shutdown Current
- 1A Peak Current Limit Using Internal Driver
- Current Mode Operation
- Internal 12/4 Ω MOSFET
- External RC Compensation
- Internal Soft-Start
- High Efficiency at Light Loads
- Current Limit and Over Temperature Protection

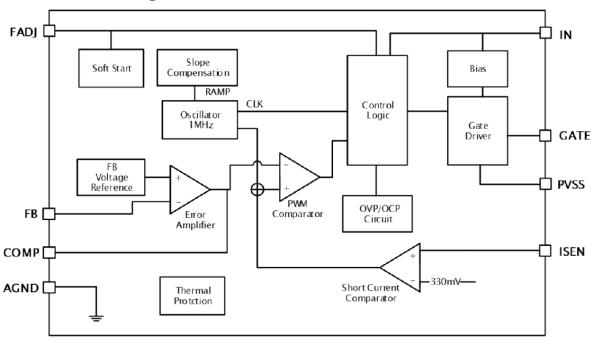
- Portable Speakers
- Offline Power Supply
- **Battery Powered Device**
- Set-Top Box
- Photovoltaic Inverters

Typical Application



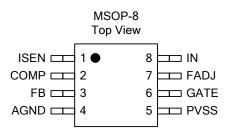


Functional Block Diagram



FB=0.8V; VP3478MSG8 FB=1.26V; VP3478SPG8

Pin Assignments And Descriptions



Pin No.	Pin	I/O/P	Function Description
1	ISEN	Р	Current Sense. Use an external resistor in series with ground to measure the voltage drop.
2	COMP	I	Compensation. Use a RC/C network to do proper loop compensation.
3	FB	1	Output Feedback. Connect the external resistor divider network from output to this pin to sense output voltage. The FB pin voltage is regulated to internal 0.8V or 1.26V reference voltage. (Reference Item Version)
4	AGND	Р	Analog Ground. Connect to exposed pad.
5	PVSS	I	Power Ground. Connect to exposed pad.
6	GATE	0	Gate Drive. Connect this terminal to the gate pin of the external MOSFET.
7	FADJ	О	Frequency Adjust/Synchronization/Shutdown. A resistor connected from this pin to ground simply sets the oscillator frequency. An external clock signal at this pin will synchronize the controller to the clock. Pull on this pin for \geq 30 μ s will turn the device off and the device will then very few current about 5μ A from the supply.
8	IN	I	Power Supply Input.



Absolutely Maximum Ratings

Over operating free-air temperature range, unless otherwise specified (* 1)

Symbol	Parameter	Limit	Unit
V _{IN}	Supply voltage range	-0.3 to 42	V
V _{LV} (COMP/FB/FADJ/GATE)	Low voltage range	-0.3 to 6	V
V _{ISEN}	Current sense pin range	-0.4 to 0.6	V
T _{JMAX}	Junction temperature MAX.	150	۰C
T _{STG}	Storage temperature range	-65 to 150	۰C
Electrostatic discharge	Human body model	2	kV
Electrostatic discharge	Machine model	200	V
$\theta_{ m JC(top)}$	Thermal resistance (Junction to Case (top))	47	°C/W
θ_{JA}	Thermal resistance (Junction to Air)	146	°C/W
$\theta_{ m JB}$	Thermal resistance (Junction to Board)	84	°C/W
Ψյт	Junction-to-top characterization parame- ter	4	°C/W
Ψյв	Junction-to-board characterization pa- rameter	80	°C/W

^{(*1):} Stress beyond those listed at "absolute maximum rating" table may cause permanent damage to the device. These are stress rating ONLY. For functional operation are strongly recommend follow up "recommended operation conditions" table.

Recommended Operating Conditions

Symbol	Parameter	Specifi	Unit		
Symbol	i arameter	Min	Max	Oilit	
V _{IN}	Supply voltage	2.97	40	V	
f _{osc}	Switching Frequency range	0.1	1	MHz	
T _J	Operating Junction range	-40	125	°C	



Electrical Characteristics

 V_{IN} =12V, R_{FADJ} =40k Ω , T_J =25°C, unless otherwise specified (* 1)

Comple al	Dawawatan	Test Condition		Specification			Unit	
Symbol	Parameter			Min	Тур	Max	Unit	
W	Foodback voltage	$V_{COMP}=1.4V$,	VP3478MSG8	0.788	0.8	0.816	V	
V_{FB}	Feedback voltage	$3V < V_{IN} < 40V$	VP3478SPG8	1.2416	1.26	1.2843	V	
			V _{IN} =12V		10			
		V _{FADJ} =3V	V _{IN} =12V,			15	μΑ	
I_{Q}	Quiescent current in shutdown		-40°C <t<sub>J<125°C</t<sub>			13		
iQ	mode	V PADJ 3 V	$V_{IN} = 5V$		5			
			V _{IN} =5V, -40°C <t<sub>J<125°C</t<sub>			10		
V_{UVLO}	Under voltage lockout				2.5		V	
V _{UV(HYS)}	Input Under voltage Lock-out Hysteresis				160		mV	
D	High–side switch R _{DS(ON)} (*1)	V _{IN} =5\	/, I _{GATE} =0.2A		12		0	
$R_{DS(ON)}$	Low-side switch R _{DS(ON)} (*1)	V _{IN} =5\	/, I _{GATE} =0.2A		4		Ω	
A _{VOL}	Error amplifier voltage gain	V _{COMP} =1.4	V _{COMP} =1.4V, I _{EAO} =100μA		60		V/V	
9 м	Error amplifier trans- conductance	V _{COMP} =1.4V			430		μ℧	
		V _{IN} <5.8V			V _{IN}		V	
V_{GATE}	Maximum GATE driving swing	V _{IN} ≥5.8V			5.7			
r	On all lating fragman and	$R_{FADJ}=40k\Omega$			0.44		MHz	
f_{OSC}	Oscillation frequency	R_{FADJ} =40k Ω , -40°C $<$ T $_{J}$ $<$ 125°C		0.34		0.48		
D _{MAX}	Maximum duty cycle	$R_{FADJ}=40k\Omega$			85		%	
ΔV_{LINE}	Voltage line regulation	$3V < V_{IN} < 40V$			0.05		%/V	
ΔV_{LOAD}	Voltage load regulation	I _{EAO} Source/Sink			±0.5		%/A	
t _{MIN(ON)}	Minimum on-time				410		nS	
lausau.	Supply Current	$\begin{array}{c} R_{FADJ}{=}40k\Omega \\ \hline R_{FADJ}{=}40k\Omega \text{ , } -40^{\circ}\text{C}{<}\text{T}_{J}{<}125^{\circ}\text{C} \end{array}$			2.7		mA	
SUPPLY	Зирріу Сипенс					4.5		
V _{SENSE}	Current sense threshold volt-	١	$I_{IN}=5V$	120	160	180	mV	
▼ SENSE	age	V _{IN} =5V , -	40°C <t<sub>J<125°C</t<sub>	120		200	111 V	
V_{SC}	Overload current limit sense		$I_{IN}=5V$		290		mV	
	voltage	V _{IN} =5V , -	40℃ <t<sub>J<125℃</t<sub>	240		415		
V_{SL}	Internal compensation ramp	\	/ _{IN} =5V		90		— mV	
	internal compensation ramp	V _{IN} =5V , -	40°C <t<sub>J<125°C</t<sub>	52		132		
V_{OVP}	Output overvoltage protection	V _{COMP} =1.4V			85		mV	
V _{OVP(HYS)}	Output overvoltage protection hysteresis	V _{COMP} =1.4V			70		mV	



Electrical Characteristics (cont.)

 $V_{\text{IN}}{=}\,1\,2\text{V},\,R_{\text{FADJ}}{=}\,40k\Omega,\,T_{\text{J}}{=}\,25^{\circ}\!\text{C},\,\text{unless otherwise specified}$ (* 1)

Symbol	Parameter	Test Condition	Specification			Unit	
Syllibol	rarameter	rest Condition	Min	Тур	Max	Oilit	
		$Source, V_{COMP} = 1.4V, V_{FB} = 0V$		127			
	Error amplifier output current	Source, $V_{COMP} = 1.4V$, $V_{FB} = 0V$ $-40^{\circ}C < T_J < 125^{\circ}C$	65		180		
I _{EAO}	(Source/Sink)	Sink, $V_{COMP} = 1.4V$, $V_{FB} = 1.4V$	180			μΑ	
		Sink, $V_{COMP} = 1.4V$, $V_{FB} = 1.4V$ $-40^{\circ}C < T_{J} < 125^{\circ}C$	100		230		
		V _{FB} =0V, COMP pin floating		2.38			
	Error amplifier output voltage	V_{FB} =0V, COMP pin floating -40°C <t<sub>J<125°C</t<sub>	2		2.6		
V_{EAO}		V _{FB} =1.4V		0.67		V	
		V _{FB} =1.4V -40°C <t<sub>J<125°C</t<sub>	0.2	0.2 1			
	Shutdown signal threshold on FADJ pin (*2)	Output=High Level		1.29	1.29		
V_{SD}		Output=High Level, −40°C <t<sub>J<125°C</t<sub>					
VSD		Output=Low Level	0.63			- V -	
		Output=low Level, -40°C <t<sub>J<125°C</t<sub>			0.4		
t _{SS}	Soft start delay	$V_{FB} = 1.2V$, COMP pin floating		9		mS	
t _R	GATE pin rising time	$Cgs = 3000pF, V_{GATE} = 0V \; to \; 3V$		75		nS	
t _F	GATE pin falling time	$Cgs = 3000pF, V_{GATE} = 3V to 0V$		20		nS	
I _{SD}	Shutdown pin current FADJ pin	$V_{SD} = 0V$		20		μΑ	
T _{SD}	Thermal shutdown			170		$^{\circ}$	
T _{SD(HYS)}	Thermal shutdown hysteresis			10		$^{\circ}$	

^{(*1):} Stress beyond those listed at "absolute maximum rating" table may cause permanent damage to the device. These are stress rating ONLY. For functional operation are strongly recommend follow up "recommended operation conditions" table.

^(*2) The FADJ pin should be pulled to VIN through a resistor to turn the regulator off. The voltage on the FADJ pin must be above the maximum limit for Output = High Level to keep the regulator off and must be below the limit for Output = Low Level to keep the regulator on.



Functional Descriptions

The VP3478 employs the current-mode, adjustable frequency pulse-width modulation (PWM) architecture. It operates at adjustable switching frequency under medium to high load current conditions.

Overvoltage Protection

The VP3478 uses FB pin to detect overvoltage occurrence. The overvoltage protection should be triggered at the voltage rises to $V_{FB} + V_{OVP}$. When OVP occurs only the MOSFET will be turned off, the output voltage will drop. VP3478 will switch when the voltage on FB pin is less then $V_{FB} + (V_{OVP} - V_{OVP(HYS)})$ for limit on $V_{OVP(HYS)}$.

Frequency Adjust

The switching frequency can be adjusted from 100kHz to 1MHz by a external resistor in series with FADJ terminal and ground. The following equation is used to calculate resistor value.

a. When Fs < 300KHz the calculate as below,

$$R_{FADJ} \cong \frac{17.5 \times 10^3}{f_c} + 8.2$$

b. When Fs > 300KHz the calculate as below,

$$R_{FADJ} \cong \frac{21 \times 10^3}{f_s} - 7.3$$

Where f_S is in kHz and R_{FA} is in k Ω .

Clock Synchronization

VP3478 is able to be synchronized to an external clock by connecting to the FADJ terminal with R_{FADJ} in series with ground as shown in figure 2.

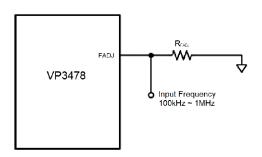


Figure 2. Clock Synchronization

Shutdown

The FADJ pin can be used as a shutdown pin. If the high signal pulls up this pin, VP3478 will stop the switching and then enter the shutdown state. In this state, VP3478 consumes only 5μ A typically.

The use of shutdown control in frequency adjustment mode is quite simple. Connects a resistor between the FADJ pin and ground will force the VP3478 runs at specified frequency and pulls this pin high will shutdown the IC. In both frequency and synchronization mode, pulls FADJ pin high lasting then 30µs will also force the VP3478 enter the shutdown state.

Slope Compensation

VP3478 employs current mode control scheme. It has many advantages such as cycle-by-cycle current limit for the switch and easier to parallel power stages because automatic current sharing. The compensation ramp is already added in VP3478 and the slope of the default compensation ramp could satisfy most applications.

Short Circuit Protection

The ISEN pin is used to sense the over-current occurrence. If the difference between ISEN pin and ground is greater then 330mV, the current limit will be activated. The comparator will decrease the switching frequency by the factor of 8 and maintains this condition until the over-current (short) event is removed.



Application Information

<u>Programming the Output Voltage and</u> Output Current

The output voltage can be programmed using a resistor divider between the output and the feedback pins, as shown in Figure 3. The resistors are selected such that the voltage at the feedback pin is 0.8V (1.26V). RF1 and RF2 can be selected using the equation,

$$V_{OUT} = V_{FB} \times (1 + \frac{R_{F1}}{R_{F2}})$$

A 100pF capacitor may be connected between the feedback and ground pins to reduce noise.

The maximum amount of current that can be delivered at the output can be controlled by the sense resistor, Rsen. Current limit occurs when the voltage that is generated across the sense resistor equals the current sense threshold voltage, Vsense. Limits for Vsense have been specified in the Electrical Characteristics section. This can be expressed as:

$$I_{\mathit{SW(pexk)}} \times R_{\mathit{SEN}} = V_{\mathit{SENSE}} - D \times V_{\mathit{SL}}$$

The peak current through the switch is equal to the peak inductor current.

$$I_{SW(pexk)} = I_L(\max) + \Delta i_L$$

Therefore for a boost converter,

$$I_{SW(pexk)} = \frac{I_{OUT(max)}}{(1-D)} + \frac{(D \times V_{IN})}{(2 \times f_s \times L)}$$

Combining the two equations yields an expression for $\ensuremath{\mathsf{RSEN}}$.

$$R_{SEN} = \frac{V_{SENSE} - (D \times V_{SL})}{\left[\frac{I_{OUT(\text{max})}}{(1-D)} + \frac{(D \times V_{IN})}{(2 \times f_S \times L)}\right]}$$

Evaluate RSEN at the maximum and minimum VIN values and choose the smallest RSEN calculated.

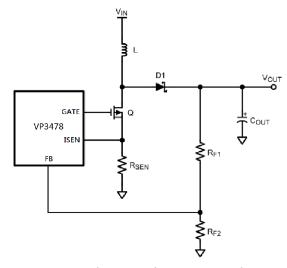


Figure 3. Adjusting the Output Voltage

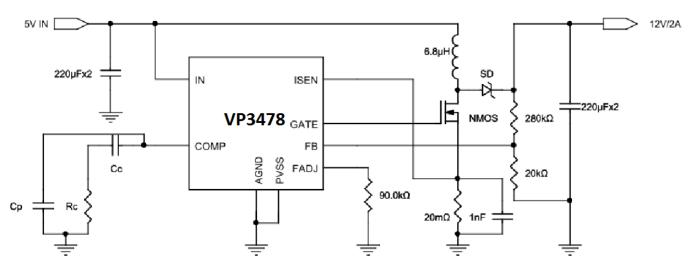
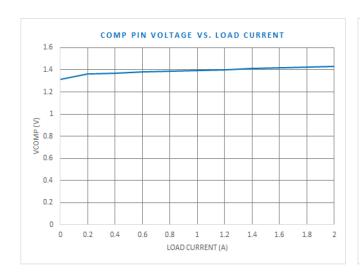
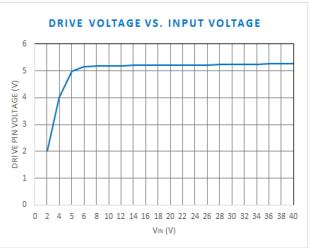


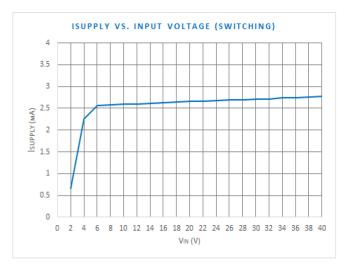
Figure 4. VP3478 Typical Boost Application

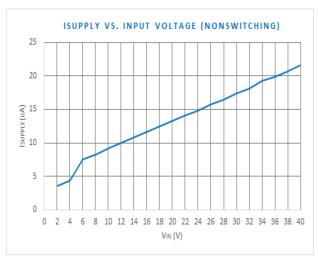


Typical Characteristic

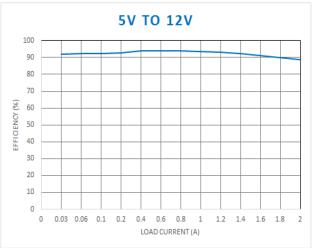








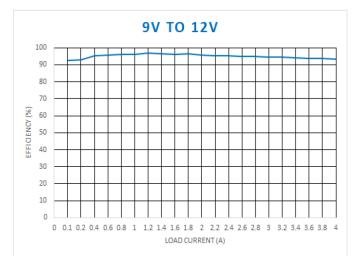


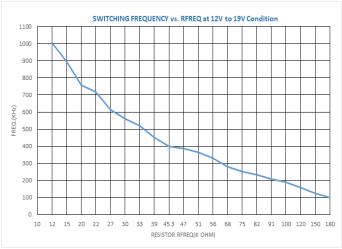


VP3478



Typical Characteristic (cont.)

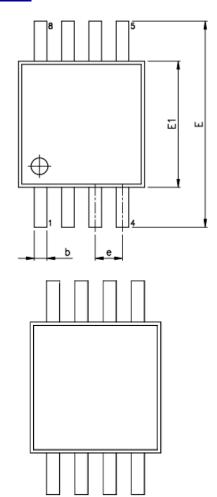






Package Information

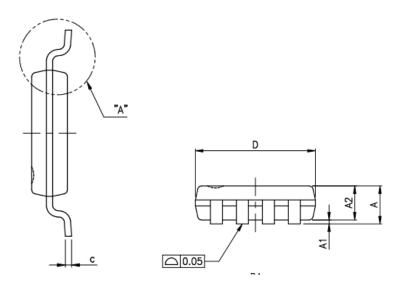
MSOP8

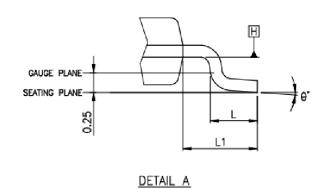


THERMALLY ENHANCED VARIATIONS ONLY

	SYMBOLS	MIN.	NOM.	MAX.	
	Α	-	-	1.10	
	A1	0.00	_	0.15	
	A2	0.75	0.85	0.95	
◬	Ь	0.22	_	0.38	
҈	С	0.08	_	0.23	
	D		3.00 BSC		
	E		4.90 BSC		
	E1	3.00 BSC			
<u></u>	е		0.65 BSC		
	L	0.40	0.60	0.80	
	L1		0.95 REF		
	θ.	0	_	8	
				LINUT . MANA	







NOTES:

1.JEDEC OUTLINE :

STANDARD : MO-187 AA.

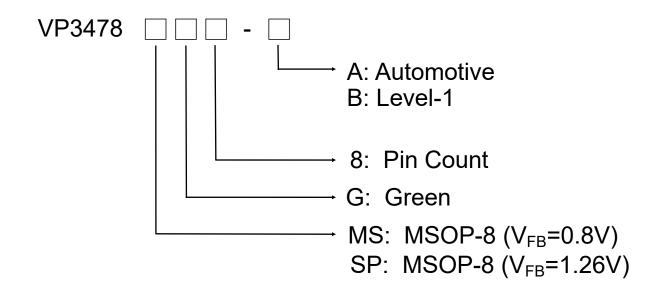
THERMALLY ENHANCED : MO-187 AA-T.
2.DIMENSION D DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.15 mm PER END. DIMENSION E1 DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.15 mm PER SIDE.

3.DIMENSION 'b' DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08 MM
TOTAL IN EXCESS OF THE 'b' DIMENSION AT MAXIMUM
MATERIAL CONDITION. THE DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OF THE FOOT. MINIMUM SPACE BETWEEN PROTRUSION AND AN ADJACENT LEAD SHALL NOT BE LESS THAN 0.07 mm.

4.D AND E1 DIMENSIONS ARE DETERMINED AT DATUM H .

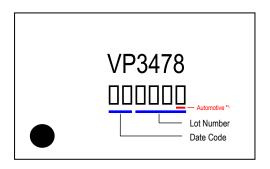


Ordering Information

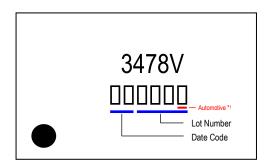


Part No.	FB voltage version	Q`ty/ Reel	MSL Peak Temp.	Op Temp.(°C)
VP3478MSG8-A	0.8V	2,500	Level 1(G1)	-40 to +125
VP3478MSG8-B	0.8V	2,500	Level 1	-40 to +125
VP3478SPG8-A	1.26V	2,500	Level 1(G1)	-40 to +125
VP3478SPG8-B	1.26V	2,500	Level 1	-40 to +125

Marking Information



FB: 0.8V Version Marking



FB: 1.26V Version Marking

^{*1} Automotive Specifications Marking will show " - "



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